

608E/F VHF SIGNAL GENERATOR

HP PART NUMBER 00608-90046

SERIALS PREFIXED:

This Operating and Service manual applies to HP 608E/F instruments with serial numbers prefixed 833- above 02720 (608E) and 832- above 01500 (608F).

SERIAL PREFIXES NOT LISTED:

For instruments with lower serial number prefixes a "Backdating" Appendix is supplied in the back of this manual.



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OPERATING AND SERVICE MANUAL PART NO. 00608-90046 Operating and Service Manual Microfiche Part No. 00608-90047

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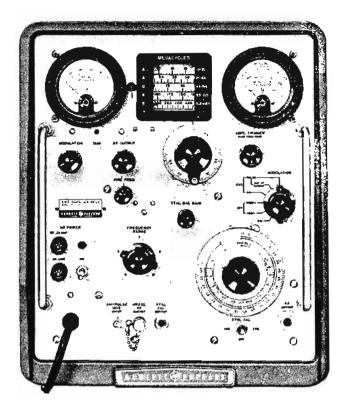
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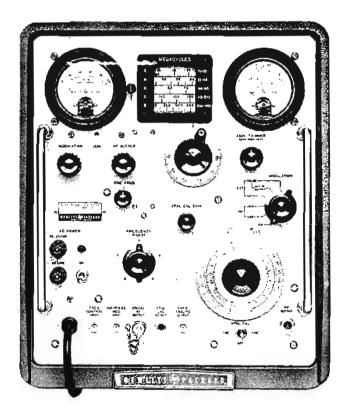
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Model 608E



Model 608F

Figure 1-1. Model 608E and 608F VHF Signal Generators

SECTION I GENERAL INFORMATION

1-1. DESCRIPTION.

1-2. The Hewlett-Packard Model 608E and 608F Signal Generators (see Figure 1-1) are designed to meet the requirements of precision laboratory work. and yet to be equally useful for general applications in the VHF frequency range. The Model 608E furnishes RF signals from 0.1 microvolt to 1 volt from 10 MHz to 480 MHz, while the Model 608F supplies RF signals from 0.1 microvolt to 0.5 volt in a frequency range from 10 MHz to 455 MHz. The RF carrier can be amplitude modulated by internally generated sinewave signals, or by externally supplied sine-wave or pulse signals. The Model 608F contains frequency control circuitry which permits the instrument to be used in phase-lock operation with the Model 8708A Synchronizer. This provision provides a stabilized output across most of the range of the instrument, with a drift factor of $< 2 \times 10^{-7}$ per 10 minutes. The instruments can be used for troubleshooting, testing, calibrating, measuring standing-wave ratios, and checking antenna and transmission line characteristics. To preserve accuracy, equipment design holds spurious modulation to a low value under all operating conditions. Detailed specifications for both instruments are listed in Table 1-1.

1-3. FREQUENCY.

1-4. The frequency of the cutput signal is indicated on a direct reading dial, the calibration of which is accurate to better than $\pm 0.5\%$ for the 608E and $\pm 1\%$ for the 608F. when the cursor and FINE FREQ are properly aligned. Calibration accuracy may be improved further by use of a built-in crystal-controlled heterodyne calibrator which furnishes 1 MHz checkpoints from 10 MHz to 270 MHz, or 5 MHz checkpoints over the entire frequency range of the instrument. At any checkpoint the calibration can be set very close to the calibrator accuracy of 0.01%, bringing overall accuracy to at least ±0.05%. Frequency check point signals are obtained when a headphone set (not furnished) is plugged into the XTAL CAL OUTPUT jack. The control for the output attenuator is calibrated in both decibels and volts. When the instrument is connected to a 50-ohm resistive load and power into the output attenuator is exactly at the ATTENUATOR CALIBRATED mark (+7 dB) on the RF OUTPUT meter, the level of power or voltage applied to the RF OUTPUT connector may be read directly on the ATTENUATOR dial with an accuracy of ±1 dB. When connected to a 50-ohm resistive load, the VSWR at the RF OUTPUT connector will not be greater than 1.2 (SWR of 1.6 dB). The signal generators feature automatic output leveling, maintaining the RF output within ± 1 dB of the adjusted output.

1-5. MODULATION.

1-6. The RF output signal can be amplitude modulated by internally generated 400 or 1000-cycle sine waves,

externally applied sine waves above 1.0 volt rms over the frequency range from 20 Hz to 20 kHz, or externally applied pulses of 10 volts peak-to-peak. When pulse modulated, the signal generators are capable of producing pulses of RF energy as short as 4 microseconds at signal frequencies above 40 MHz, and pulses as short as 2 microsecond above 220 MHz. The degree of sine-wave modulation is continuously variable from 0 to 90% by a front-panel control. All sine-wave modulation of the output signal is continuously monitored and indicated in percentage on a direct reading modulation meter.

1-7. RFI.

1-8. RF leakage is held to a minimum and is such that when the output signal is adjusted for 0.1 microvolt, the conducted signal leakage at any other front panel connector and the radiated leakage two inches from the instrument are each less than 1.0 microvolt.

1-9. AUXILIARY EQUIPMENT.

1-10. MODEL 11509A FUSEHOLDER.

1-11. To protect the output attenuator from damage, for some applications (such as transceiver testing) it is desirable to insert a fuse between the signal generators and external equipment. The 11509A is a special coaxial fuseholder which houses a type 8AG, 1/16 amp fuse which protects the output attenuator from damage in the event that an external voltage is accidently applied to the RF OUTPUT connector. The fuseholder has an insertion loss of 0.50 dB at 200 MHz, 0.56 dB at 300 MHz, and 0.65 dB at 400 MHz; its SWR is not greater than 1.35 when connected to a 50ohm resistive load.

1-12. MODEL 11508A TERMINATED OUTPUT CABLE.

1-13. This cable assembly provides a 50-ohm termination and standard binding posts at the end of a 24inch length of cable. The 11508A allows direct connection of the instruments to a high-impedance circuit.

1-14. INSTRUMENT IDENTIFICATION.

1-15. Hewlett-Packard instruments carry a two-section, eight-digit serial number. The first three digits are a prefix. The contents of this manual apply to those instruments having the serial number prefix shown on the title page. If the serial prefix on your instrument is not mentioned on the title page, in the Appendix to this manual, or in a Manual Change Sheet enclosed with the manual, the correct information may be obtained from any Hewlett-Packard Sales and Service Office (see rear of manual for addresses).

608E	608F
Range: 10 to 480 MHz in five bands (10-21, 21- 43, 43-95, 95-215, and 215-480 MHz).	Range: 10 to 455 MHz in five bands (10-21, 21- 44, 44-95, 95-210, and 210-455 MHz). 10- 430 MHz when used with 8708A Synchro- nizer.
Accuracy: ±0.5% with cur- sor adjustment.	Accuracy: ±1% with cur- sor adjustment.
Drift: Less than 50 parts in 10 ⁶ per 10 min. period after 1 hr. warmup.	Drift: Less than 50 parts in 10 ⁶ per 10 min. pe- riod after 1 hr. warm- up. Stability used with 8708A Synchronizer.
	$5 \ge 10^{-8}$ /min.
	$2 \ge 10^{-7}$ /min.
	(0°-55°C)
	$2 \times 10^{-7}/10\%$ line voltage change.
Tuning Control: Freque	ency control mechanism

- Tuning Control: Frequency control mechanism provides a main dial calibrated in MHz and a vernier dial for interpolation purposes. Total scale length approximately 45 inches. Calibration: every other MHz from 130 to 270 MHz; every 5 MHz above 270 MHz.
- Resettability: Main frequency control resettability better than ±0.1% after initial warmup. FNE FREQUENCY ADJUST provides approximately ±25 kHz settability at 480 MHz (proportionately finer adjustment at lower frequencies).
- Crystal Calibrator: Provides frequency check points every 1 MHz up to 270 MHz or every 5 MHz over the range of the instrument. Headphone jack provided for audio frequency output (headphones not included). Crystal frequency accuracy better than 0.01% at normal room temperatures. Cursor on frequency dial adjustable over small range to aid in interpolation adjustment. Calibrator may be turned off when not in use.
- Residual FM: $<\pm 5$ parts in 10^7 in a 10 kHz postdetection bandwidth.
- Harmonic Output: At least 35 dB below the carrier for harmonic frequencies below 500 MHz.
- Frequency Control Input: (608F only). The 608F FREQ CONTROL INPUT normally used with the 8708A Synchronizer can also be used for external frequency control by varying a dc voltage

input. A voltage change from -2 volts to -32 volts will change the output frequency more than 0.2% at the low end of each band and more than 1% at the high end of each band. Nominal 4 K Ω input impedance, direct coupled. Voltage limits: $0V \leq applied voltage \leq -50V$.

OUTPUT CHARACTERISTICS

608E

608F

- Output Level: Continuously adjustable from 0.1 μ V to 1V into a 50 Ω load. Output attenuator calibrated in V and dBm (0 dBm = 1 mW in 50 Ω). Output Level: Continuously adjustable from 0.1 μ V to 0.5V into a 50 Ω load. Output attenuator calibrated in V and dBm (0 dBm = 1 mW in 50 Ω).
- Accuracy: Within ±1 dB of attenuator dial reading at any frequency when RF output meter indicates "ATTENUATOR CALIBRATED."
- Leveling: Internal feedback circuit retains "AT-TENUATOR CALIBRATED" reference on RF output meter over wide frequency ranges (typically octave bands). Adjustment of front panel AMP. TRIMMER control (only) for maximum RF output indication automatically restores initial carrier level for greater frequency changes.
- Impedance: 50Ω; reflection coefficient <0.091
 (1.2 SWR, 20.8 dB return loss) for attenuator
 setting below -7 dBm.</pre>
- RFI: Meets all conditions specified in MIL-I-6181D; permits receiver sensitivity measurements down to at least 0.1 μ V.

Auxiliary RF Output:

608E

Fixed level CW signal from RF oscillator (minimum amplitude 180 mV rms into 50Ω) provided at front panel BNC female connector for use with external equipment (e.g., frequency counter).

608F

CW signal from RF Oscillator provided at front panel ENC female connector. Power levels into 50Ω are as follows:

10 to 215 MHz: -1.8 to +7 dBm 215 to 400 MHz: +2.0 to +6 dBm 400 to 430 MHz: +1.0 to +5 dBm

Signal for use with HP 8708A Synchronizer or other external equipment (e.g., frequency counter).

MODULATION CHARACTERISTICS

(Front-panel AMP, TRIMMER control adjusted for maximum indication of RF Output Meter and RF Output Meter set to Attenuator Calibrated.)

Internal AM

Frequency: 400 and 1000 Hz, +10%. Modulalation signal available at front panel BNC female connector for synchronization of external equipment.

608E

608F

Modulation Level: 0 to Modulation Level: 0 to 95% modulation at 95% modulation with carrier levels 0.5 V Output Attenuator at and below; continu-0.224 V (1 mW) or beously adjustable with low; continuously adfront-panel MOD justable with front-LEVEL control. panel MOD LEVEL control.

Carrier Envelope Distortion: Less than 2% at 30% AM and less than 5% at 70% AM.

External AM

Frequency: 20 Hz to 20 kHz.

608E

608F

- Modulation Level: 0 to Modulation Level: 0 to 95% modulation with 95% modulation at carrier levels 0.5 V Output Attenuator at and -below; continu-0.224 V (1 mW) or beously adjustable with low; continuously adfront-panel MOD justable with front-LEVEL control. Inpanel MOD LEVEL control. Input reput required: 1 to 10 quired: 1 to 10 V $V rms (1 k \Omega input)$ rms (1 k Ω impedance). impedance).
 - Carrier Envelope Distortion: Less than 2% at 30% AM, less than 5% at 70% AM (modulation source distortion less than 0.5%).
 - External control of carrier level can be achieved through application of dc voltage in EXT AM mode.

Modulation Meter

- Accuracy: $\pm 5\%$ of full scale 0 to 80%, $\pm 10\%$ from 80% to 95% (for INT AM or 20 Hz to 20 kHz EXT AM).
- Incidental Frequency Modulation (at 400 and 1000 Hz modulation)
 - Less than 1000 Hz peak at 50% AM for frequencies above 100 MHz. For frequencies below 100 MHz, less than 0.001% at 30% AM.

External Pulse Modulation

Rise and Decay Time: From 40 MHz to 220 MHz, combined rise and decay time less than $4 \mu s$.

Above 220 MHz, combined rise and decay time less than 2.5 μ s.

- On-Off Ratio: At least 20 dB for pulsed carrier levels of 0.5 V and above.
- Input Required: Positive pulse, 10-50 V peak; input impedance $2 \ k \ \Omega$.

GENERAL

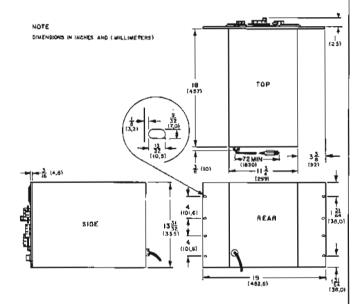
608E and 608F

Power: 115 or 230 V ±10%, 50 to 400 Hz. Approximately 220 W.

Dimensions

Cabinet Mount: 13-1/4 in. (337 mm) wide, 16-3/4in. (416 mm) high, 21 in. (533 mm) deep.

Rack Mount:



Weight

input

- Cabinet Mount: Net, 62 lb (28 kg). Shipping, 74 1b (33, 4 kg).
- Rack Mount: Net, 62 lb (28 kg). Shipping, 83 lb (37, 4 kg).

Accessories Available

- 11508A Output Cable provides 50- Ω termination and standard binding posts at the end of a 24inch (610 mm) length of cable. Allows direct connection of the signal generator to high impedance circuits.
- 11509A Fuse Holder provides protection for the output attenuator when the 608 is used for transceiver tests.
- 10514A Mixer for use as nanosecond pulse modulator or balanced modulator.

10515A Doubler for extending the frequency range.

SECTION II

2-1. INCOMING INSPECTION.

2-2. This instrument was inspected mechanically and electrically prior to shipment. Inspect for mechanical damage received in transit and test electrical performance using procedures given in Section V. If there is damage or deficiency, or if electrical performance is not within specifications, notify the carrier and your Hewlett-Packard Sales and Service Office immediately (see list at rear of this manual.)

2-3. PREPARATION FOR USE.

2-4. POWER REQUIREMENTS.

2-5. The VHF Signal Generator requires a power source of 115 or 230 volts at $\pm 10\%$, single phase, which can supply approximately 220 watts. A twoposition slide switch, on the inside rear chassis, permits operation from either a 115- or 230-volt power source. The number visible on the switch slider indicates the line voltage for which the signal generator is connected. The correct fuse rating for each line voltage is adjacent to the switch. To prepare the signal generator for operation, position the 115-230 volt switch so that the number visible on the slider corresponds to the available line voltage, and install a fuse of correct rating (marked on rear apron).

CAUTION

Before connecting the power cable, set the 115-230 volt switch for the line voltage to be used.

2-6. POWER CABLE.

2-7. To protect operating personnel, the National Electrical Manufacturers' Association (NEMA) recommends that the instrument panel and cabinet be grounded. This instrument is equipped with a threeconductor power cable which, when plugged into an appropriate receptable, grounds the instrument. The off-set pin of the three-prong connector is the ground pin. To preserve the protection feature when operating the instrument from a two-contact outlet, use a three-prong to two-prong adapter (Hewlett-Packard Stock No. 1251-0049) and connect the green pigtail on the adapter to ground.

2-8. REPACKING FOR SHIPMENT.

2-9. If the signal generator is to be packaged for shipment use the original shipping container and packing materials. If these have been discarded or are not in condition for reuse, obtain new materials from your local Hewlett-Packard Sales and Service Office (see rear of this manual for locations), or follow these general instructions:

a. Wrap the signal generator in heavy paper or plastic. (If the signal generator is being shipped to a Hewlett-Packard service facility, attach a tag indicating type of servicing required, return address, model number, and full serial number.)

b. Use a strong shipping container. A carton made of 500-600 pound test material will usually provide adequate protection.

c. Use enough shock-absorbing material (3 to 4 inch layer) around all sides of instrument to provide firm cushion and prevent movement inside the container. Protect the control panel with cardboard. With Hewlett-Packard "floater pack" packaging, the foam blocks provide sufficient shock protection, and additional material is unnecessary.

d. Seal the shipping container securely.

e. Mark the shipping container "FRAGILE" to assure careful handling.

2-10. In any correspondence refer to the signal generator by model number and full serial number.

SECTION III OPERATION

3-1. INTRODUCTION.

3-2. This section provides operating instructions for the Model 608E and 608F Signal Generators. Included are a listing of controls, indicators and connectors, pre-operation procedures, mode operating procedures, and phase-lock operation procedures (Model 608F only). Controls and indicators on both instruments are nearly identical; therefore, an illustration of the Model 608F will be referenced during this discussion, with the difference between instruments identified.

3-3. CONTROLS, INDICATORS AND CONNECTORS.

3-4. Except for the 115/230 volt power switch, all controls, indicators, and connectors are located on the front panel. Figure 3-1 locates and provides a functional description of each front-panel mounted control, indicator and connector. The 115-230 volt power switch is located inside the rear panel and is not shown.

3-5. PRE-OPERATION PROCEDURES.

3-6. Prior to use, certain procedures must be completed to obtain maximum accuracy during operation of the instrument. Consideration must be given to the possibility of RF loading, and calibration of the instrument is required. During the following discussion, each control used will be identified by an item number that is referenced to Figure 3-1.

3-7. RF OUTPUT LOADING.

CAUTION

Do not connect any source of RF or dc power to the RF OUTPUT connector on the signal generator. To do so may permanently damage the impedance-matching network in the Attenuator Section. Particular care must be observed when testing transceiver-type equipment to insure that the transmitter remains inactive while the equipment is connected to the signal generator.

NOTE

To protect the Attenuator when there is a possibility of voltage being applied to the RF OUTPUT connector, a Model 11509A Fuseholder is recommended. The fuseholder is connected between the test cable and RF OUT-PUT connector. 3-8. A resistive load of 50 ohms is used at the factory during calibration of the Attenuator Control dial. Therefore, for an accurate indication of output power the external load should be 50 ohms resistive. The internal impedance of the generator is close to 50 ohms, with a maximum VSWR of 1.2 existing when the generator is terminated in 50 ohms. This VSWR will have no important effect on the accuracy of the Attenuator Control calibration. However, a severe mismatch between the instrument and load will produce a considerable difference between the output voltage selected on the Attenuator Control dial and the actual voltage impressed across the external load. Particular care should be exercised in the selection of coaxial connectors. A coaxial connector that has been improperly assembled can produce a substantial increase in the standing-wave ratio.

3-9. TURN-ON PROCEDURES.

3-10. Turn-on procedures for the signal generator are as follows:

CAUTION

Do not obstruct the ventilating louvers on the side of the instrument cabinet. Safe operating temperature requires free air flow through the louvers.

a. Set AC POWER switch (item 12) to the OFF position.

b. Check position of 115-230 volt power switch (refer to paragraph 2-5) on rear of chassis.

c. Set MODULATION selector switch (item 20) to CW position.

d. Set RF OUTPUT control (item 5) full clockwise (maximum).

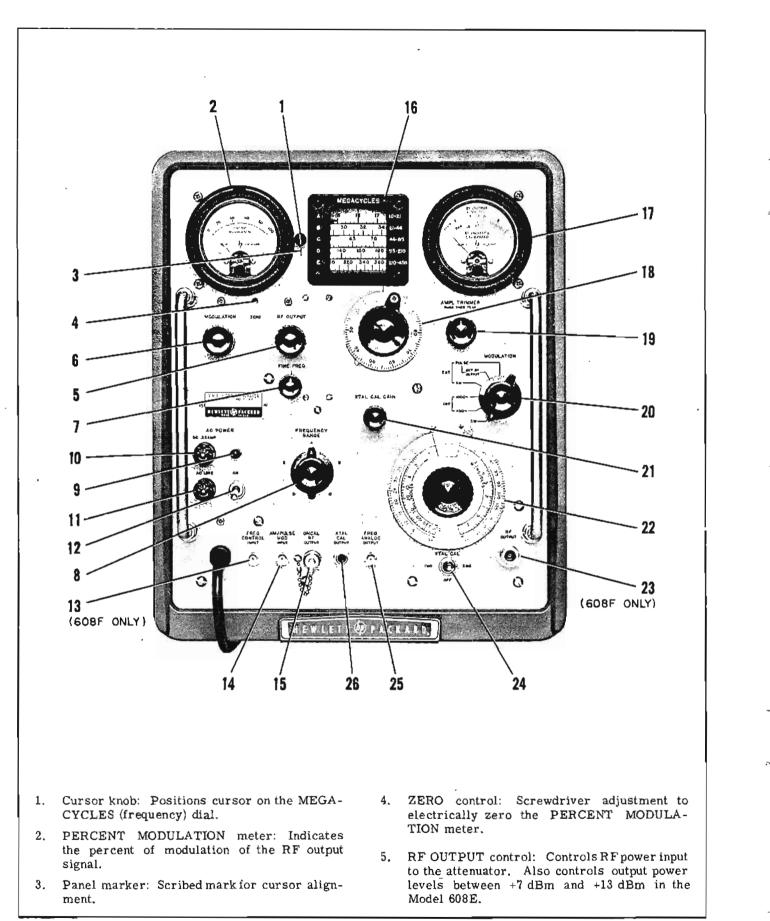
e. Set AC POWER switch to ON position.

f. Warmup equipment for 10 minutes prior to use. If greater frequency stability is required, extend warmup period to 60 minutes.

3-11. CALIBRATION PROCEDURES.

3-12. Prior to operation certain calibration of the instrument may be deemed necessary. The following discussion contains frequency and power output calibration procedures.

3-13, FREQUENCY CALIBRATION. The frequency (MEGA-CYCLES) dial in the signal generators are calibrated to be accurate within $\pm 0.5\%$ for the 608E and $\pm 1\%$ for the 608F.



- 6. MODULATION control: Adjusts the percent of modulation of the RF output signal as observed on the PERCENT MODULATION meter.
- 7. FINE FREQ control: Used for fine frequency adjustments
- 8. FREQUENCY RANGE switch: Five-position switch used to select the desired frequency range. Also positions the range pointer on the dial. Switch positions are as follows:

RANGE	MODEL 608E	MODEL 608F
Α	10-21 MHz	10-21 MHz
В	21-43 MHz	21-44 MHz
С	43-95 MHz	44-95 MHz
D	95-215 MHz	95-210 MHz
E	215-480 MHz	210-455 MHz

- 9. AC POWER lamp: Indicates when ac power is applied to the instrument.
- 10. DC .25 AMP fuse: Overload protective device for the +225 Vdc power supply.
- 11. AC LINE fuse: Protective device between the instrument and the ac power source.
- 12. AC POWER ON switch: Controls application of ac power to the instrument.
- 13. FREQ CONTROL INPUT connector (Model 608F only): Input connector for the dc control voltage when the instrument is used with a Model 8708A Synchronizer.
- 14. AM/PULSE MOD INPUT connector: Input connector for AM or pulse signals during external sine-wave or pulse operation. During internal sine-wave modulation, a 400 or 1000 Hz signal is available at the connector for synchronizing external equipment.
- 15. UNCAL RF OUTPUT connector: Output connector in the 608E delivers an uncalibrated, unmodulated RF signal to external equipment. In the Model 608F used with the Model 8708A Synchronizer, output connector delivers an uncalibrated FM signal.
- 16. MEGACYCLES dial: Indicates selected frequency range and frequency in MHz.
- 17. RF OUTPUT meter: Indicates level of RF power applied to the attenuator as selected by operation of the RF OUTPUT control. Also used in conjunction with the AMPL TRIMMER control to peak the RF output signal.
- 18. Frequency control: Used to select the output frequency within a frequency range as observed on the MEGACYCLES dial.

- 19. AMPL TRIMMER control: A two-function control for peaking the RF output. During pulse modulation, the control is tuned without pushing to peak the RF output. When sine-wave or CW operations are conducted, the control is pushed and then tuned.
- 20. MODULATION switch: A six-position switch that sets the instrument mode of operation. The switch positions are identified as follows:
 - PULSE: Sets the instrument for external pulse modulation.
 - PULSE SET RF OUTPUT: Used to establish the instrument operating point for external pulse modulation.
 - AM: Sets the instrument for external sinewave modulation.
 - 1000 ~ : Sets the instrument for internal AM modulation at 1000 Hz.
 - 400 ~ : Sets the instrument for internal AM modulation at 400 Hz.
 - CW: Sets the instrument to provide a CW output.
- 21. XTAL CAL GAIN control: Controls audio level of beat-frequency output at the XTAL CAL OUTPUT connector.
- 22. Attenuator control: Controls RF output levels of +7 dBm and below. Output levels selected by use of the Attenuator Control are accurate as long as the RF OUTPUT meter needle indicates +7 dBm (ATTENUATOR CALIBRATED).
- RF OUTPUT connector: Output connector for the calibrated RF signal (refer to CAUTION, paragraph 3-7).
- 24. XTAL CAL switch: Three-position toggle switch used when performing frequency calibrations. Switch positions are as follows:
 - 1 MC: Provides 1 MHz checkpoints from 10 to 270 MHz.
 - 5 MC: Provides 5 MHz checkpoints from 10-480 MHz in the Model 608E and from 10-455 MHz in the Model 608F.

OFF: De-energizes oscillator circuits.

- 25. FREQ ANALOG OUTPUT connector (Model 608F only): Delivers a resistance proportional to frequency to the Model 8708A Synchronizer during phase-lock operation.
- 26. XTAL CAL OUTPUT connector: Supplies calibrating signals to an operator for frequency calibration. Accommodates an earphone plug.

To obtain higher accuracy, a crystal-controlled calibrator has been included to provide a means of calibrating the instrument at any multiple of 1 MHz between 10 and 270 MHz, or any multiple of 5 MHz across the frequency range of the instrument. Procedures to crystal-calibrate the instrument are as follows:

a. Conduct turn-on procedures listed in paragraph 3-9.

b. Connect headphones (not supplied) to the XTAL CAL OUTPUT connector (item 26).

c. Set XTAL CAL switch (item 24) to 1 MC or 5 MC as applicable.

d. Rotate frequency control (item 18) until MEGA-CYCLES dial (item 16) is set on the 1 MHz or 5 MHz checkpoint nearest the frequency to be used.

e. Adjust XTAL CAL GAIN control (item 21) for a comfortable audio level in headphones.

NOTE

Extraneous beat notes generated by RF harmonics may be eliminated by properly setting the RF OUTPUT control (paragraph 3-10, step d) and the XTAL CAL GAIN control (paragraph 3-13, step e).

f. Slowly rotate frequency control around the selected checkpoint until a null (zero beat) is reached.

g. Rotate Cursor Knob (item 3) until cursor is aligned on MEGACYCLES dial with zero-beat point achieved in step f.

h. Rotate frequency control until selected frequency on MEGACYCLES dial appears directly under cursor.

i. Set XTAL CAL switch to OFF.

3.14. A FINE FREQ control (item 7) is located on the instrument front panel for making minor frequency adjustments. During normal operation no adjustment of the FINE FREQ control is required, and the knob arrow should be aligned with the front-panel marker. For a precision frequency selection using a frequency counter, the FINE FREQ control is used after the frequency has been selected by use of the frequency control. Tuning the knob arrow counterclockwise from the panel marker lowers the frequency, while a clockwise rotation raises the frequency.

3-15. POWER OUTPUT CALIBRATION. The RF leveling circuit in the instrument maintains a flat RF output over wide frequency ranges with proper initial adjustments. These adjustments are conducted using the RF OUTPUT and AMPL TRIMMER controls in conjunction with the RF OUTPUT meter. The take-off point for the detected RF signal displayed on the RF OUTPUT meter is at the input circuit of the attenuator. Therefore, calibrating the RF output into the attenuator ensures that subsequent operation of the attenuator control will produce an accurate calibrated RF output from the attenuator. RF output calibration procedures are as follows: a. Conduct turn-on procedures listed in paragraph 3-9.

b. Set the MODULATION switch (item 20) to the CW position.

NOTE

For maximum accuracy, power output calibrations should be conducted with the instrument set in the CW mode.

c. Set the FREQUENCY RANGE switch (item 8) to the desired range.

d. Tune the frequency control (item 18) five turns in from the low end of the selected frequency range as indicated on the MEGACYCLES dial (item 16).

e. Adjust the RF OUTPUT control (item 5) for an indication on the RF OUTPUT meter (item 17).

f. Depress the AMPL TRIMMER control (item 19) and tune for a peak RF indication on the RF OUTPUT meter.

g. Readjust the RF OUTPUT control until the RF OUTPUT meter pointer is aligned with the +7 dBm (ATTENUATOR CALIBRATED) mark.

NOTES

To calibrate range "E" tune to top of band and repeat steps e through g, before proceeding on to step h.

The RF OUTPUT control may be adjusted to a value >+7 dBm (RF OUTPUT meter reading); however, for maximum tube life it is recommended that the level be reduced when the higher level is not needed.

h. Tune the frequency control across the selected frequency range. The RF OUTPUT meter pointer should remain aligned with the ATTENUATOR CAL-IBRATED mark.

NOTE

If the pointer does not remain aligned with the ATTENUATOR CALIBRATED mark, the leveling circuit is not functioning properly. Stop tuning and readjust the AMPL TRIM-MER control to restore leveling. The latter adjustment is made without depressing the control.

i. Proceed with operating adjustments listed in paragraphs 3-18 and 3-20.

3-16. MODE OPERATING PROCEDURES.

3-17. The signal generators have four operating modes identified as CW operation, internal sine-wave modulation, external sine-wave modulation, and external pulse modulation. In addition, the RF output level established for CW operation can be remotely controlled by an externally supplied dc control voltage. The following discussion combines locally controlled CW operation and internal sine-wave modulation into operating instructions used with internally generated signals, external sine-wave modulation, external pulse modulation and remote-controlled CW operation. These operating procedures are used with externally supplied signals.

3-18. INTERNAL MODULATION.

3-19. The signal generators provide internal facilities to produce a CW output, or selection of either a 400 Hz or 1000 Hz sine-wave modulated signal output. Figure 3-2 illustrates and provides instructions for proper operation of the controls, indicators, and connectors used with internally generated signals.

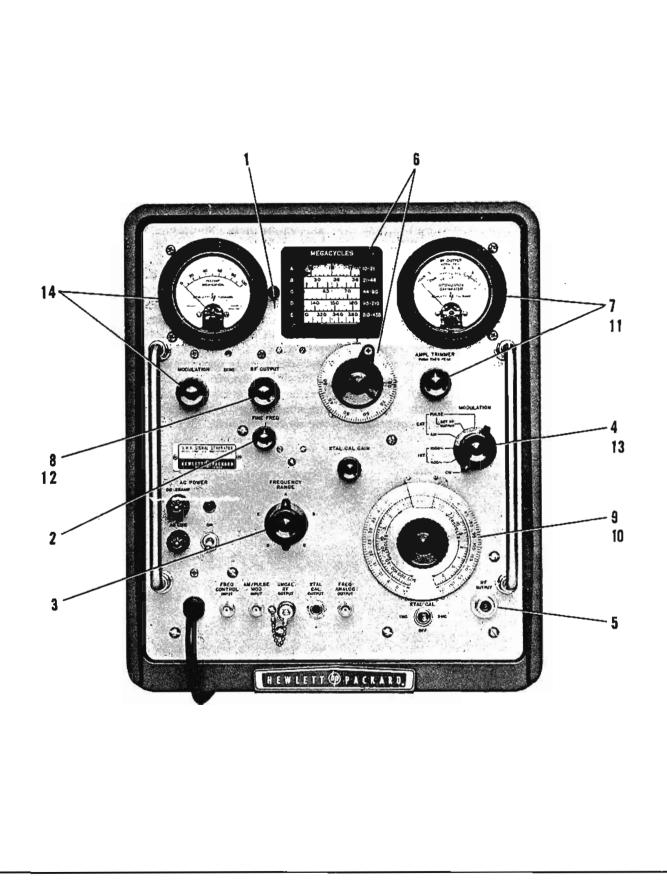
3-20. EXTERNAL MODULATION

3-21. The signal generators operate with externally supplied signals which are used to modulate the internally developed RF carrier. Also, an external dc voltage can be applied to provide remote on-off control of the RF carrier output from the instruments. Figure 3-3 illustrates and describes the operation of controls, indicators, and connectors used in the three operations.

3-22. PHASE-LOCK OPERATING PROCEDURES (MODEL 608F ONLY).

3-23. The RF Generator section in the Model 608F signal generator is designed for phase-lock operation with the Model 8708A Synchronizer. This method of operation provides a highly stable output from the instrument, and requires only normal control adjustments followed by cable connections between the signal generator and synchronizer. Figure 3-4 illustrates and describes operation of controls, indicators and connectors on the Model 608F when the instrument is used in phase-lock operation.

3-24. Configuration of the RF Generator Section in the Model 608F signal generator also permits the instrument to be used in narrow-band frequency or phase modulation operations. In this application, consideration must be given to the internal bias of -22 volts dc on the FREQ CONTROL INPUT connector, and the input impedance of approximately 5K ohms. A blocking capacitor must be added to the input circuit for passing the modulating signal while preventing any change in the reference reverse bias applied to the varicap diodes.



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Figure 3-2. Controls, Indicators, and Connectors Used in Operation with Internally Generated Signals

CW OPERATION

NOTE

Perform turn-on procedures listed in paragraph 3-9.

- 1. Align cursor knob with panel marker.
- 2. Align FINE FREQ control with panel marker.
- 3. Set FREQUENCY RANGE switch to desired range (frequency limits for each range shown on MEGACYCLES dial housing).
- 4. Set MODULATION switch to the CW position.

CAUTION

Do not connect any source of RF or dc power to the RF OUTPUT connector on the signal generator. To do so may permanently damage the impedance-matching network in the Attenuator Section. Particular care must be observed when testing transceiver-type equipment to insure that the transmitter remains inoperative while the equipment is connected to the signal generator (refer to NOTE in paragraph 3-7).

5. Connect the equipment under test to the RF OUTPUT connector observing that the signal generator is terminated in the proper load (refer to paragraph 3-8).

NOTE

If deemed necessary, conduct power output calibration procedures for the selected frequency range as listed in procedural steps d. through h. in paragraph 3-15.

6. Tune the Frequency Control until MEGA-CYCLES dial cursor is aligned with frequency to be used as indicated on dial face.

NOTE

To crystal calibrate the frequency dial, refer to procedures listed in paragraph 3-13.

- 7. Depress and tune AMPL TRIMMER control for maximum indication on RF OUTPUT meter.
- 8. Adjust RF OUTPUT control until RF OUTPUT Meter pointer is aligned with +7 dBm mark (ATTENUATOR CALIBRATED).

NOTE

Accuracy of the Attenuator Control setting and PERCENT MODULATION meter are within specifications only when the RF OUT-PUT meter is indicating ATTENUATOR CALIBRATED.

9. Adjust Attenuator Control for desired output level of +7 dBm or below.

NOTE

To select RF output levels from +7 dBm to +13 dBm in the Model 608E, perform steps 10 through 12. In this operation, the RF OUTPUT meter pointer is used in lieu of the Attenuator Control to indicate the power output level.

- Adjust the Attenuator Control to the +7 dBm dial mark.
- 11. Check that the RF OUTPUT meter pointer indicates ATTENUATOR CALIBRATED.
- 12. Adjust RF OUTPUT control for desired output level above +7 dBm as indicated by the RF OUTPUT meter pointer.

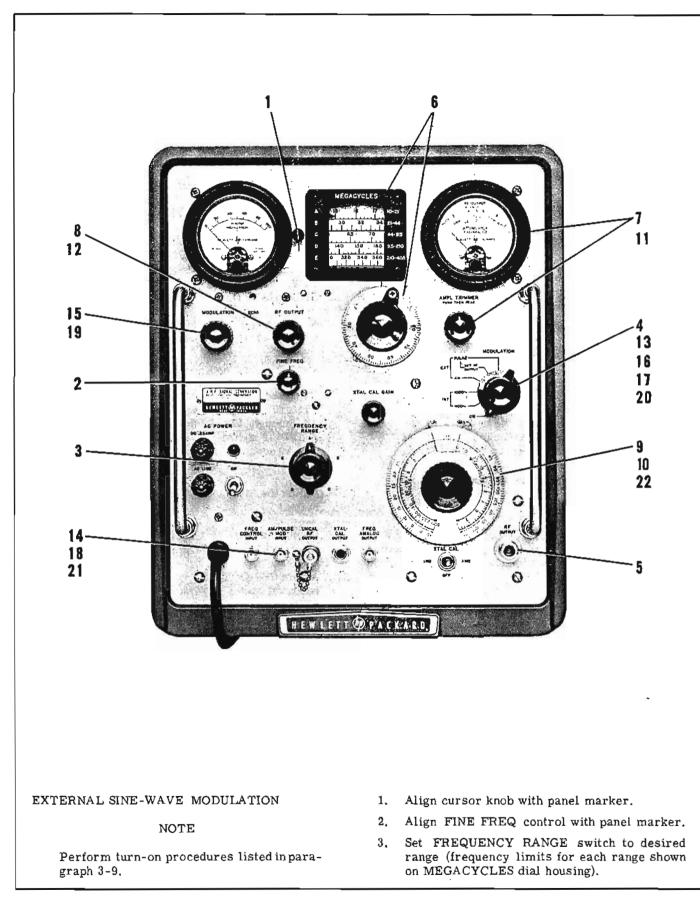
INTERNAL SINE-WAVE MODULATION

NOTE

Steps 1 through 12, as applicable, must be completed before setting the instrument for internal sine-wave AM modulation.

- Set MODULATION switch to 400 ~ or 1000
 position as desired.
- 14. Adjust MODULATION control for desired modulation percentage as indicated on the PER-CENT MODULATION meter.

Figure 3-2. Controls, Indicators and Connectors Used in Operation with Internally Generated Signals (Cont.)



4. Set MODULATION switch to CW position.

CAUTION

Do not connect any source of RF or dc power to the RF OUTPUT connector on the signal generator. To do so may permanently damage the impedance-matching network in the Attenuator Section, Particular care must be observed when testing transceiver-type equipment to insure that the transmitter remains inoperative while the equipment is connected to the signal generator (refer to NOTE in paragraph 3-7).

5. Connect the equipment under test to the RF OUTPUT connector observing that the signal generator is terminated in the proper load (refer to paragraph 3-8).

NOTE

If deemed necessary, conduct power output calibration procedures for the selected frequency range as listed in procedural steps d. through h. in paragraph 3-15.

6. Tune Frequency Control until MEGACYCLES dial cursor is aligned with frequency to be used as indicated on dial face.

NOTE

To crystal calibrate the frequency, refer to procedures listed in paragraph 3-13.

- Depress and tune AMPL TRIMMER control for maximum indication on RF OUTPUT meter.
- 8. Adjust RF OUTPUT control until RF OUTPUT meter pointer is aligned with +7 dBm mark (ATTENUATOR CALIBRATED).

NOTE

Accuracy of the Attenuator Control setting and PERCENT MODULATION meter are within specifications only when the RF OUT-PUT meter is indicating ATTENUATOR CALIBRATED.

9. Adjust Attenuator Control for desired output level of +7 dBm or below.

NOTE

To select RF output levels from +7 dBm to +13 dBm in the Model 608E, perform steps 10 through 12. In this operation, the RF OUTPUT meter pointer is used to indicate the power output level in lieu of the Attenuator Control.

10. Adjust the Attenuator Control to the +7 dBm dial mark.

- 11. Check that the RF OUTPUT meter pointer indicates ATTENUATOR CALIBRATED.
- 12. Adjust the RF OUTPUT control for the desired output level above +7 dBm as indicated by the RF OUTPUT meter pointer.
- 13. Set the MODULATION switch to the AM position.
- 14. Connect the external sine-wave generator to the AM/PULSE MOD INPUT connector and set generator for a sine-wave input between 20 Hz and 20 kHz.
- 15. Set MODULATION control for desired percentage of modulation as observed on the PER-CENT MODULATION meter.

EXTERNAL PULSE MODULATION

NOTE

Perform turn-on procedures if required, and complete steps 1 through 3.

16. Set MODULATION switch to the PULSE / SET RF OUTPUT position.

NOTE

Perform steps 5 through 12 as applicable.

- 17. Set the MODULATION switch to the PULSE position.
- Connect the external pulse generator to the AM/PULSE MOD INPUT connector and set generator for an input pulse level above 10 volts peak-to-peak.

EXTERNAL DC CONTROL

NOTE

Perform turn-on procedures if required, and complete steps 1 through 8.

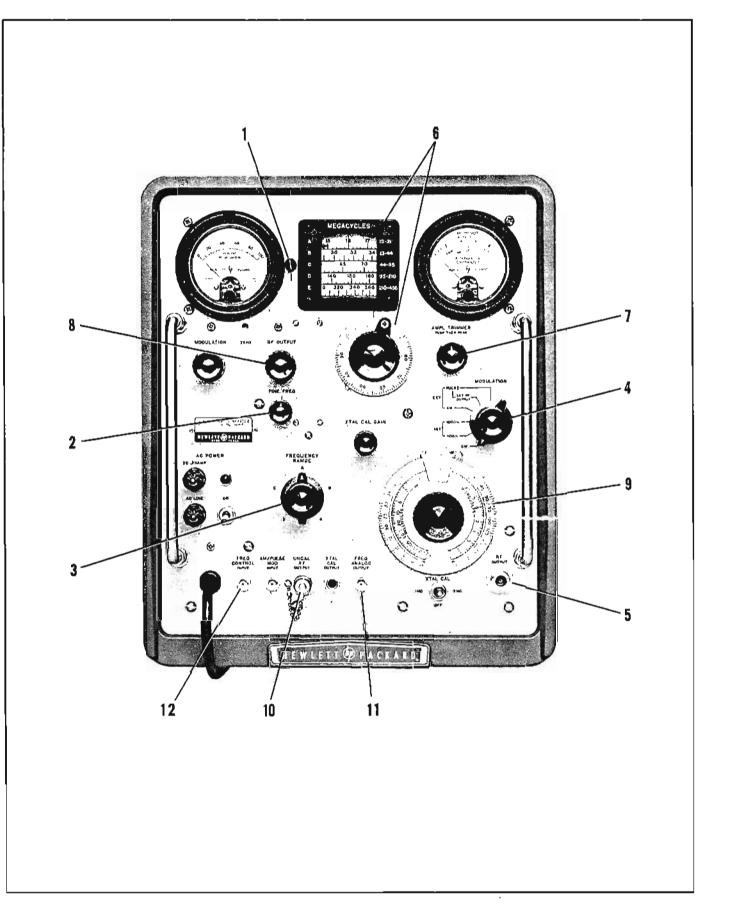
- 19. Set the MODULATION control to a maximum clockwise position.
- 20. Set MODULATION switch to the AM position.
- 21. Connect external power source to AM/PULSE MOD INPUT connector.

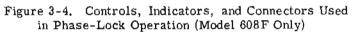
NOTE

Voltages required for external control of the RF carrier are +15 volts dc for full-on operation and -15 volts dc for full-off operation.

22. Adjust Attenuator Control and dc control voltage input for desired RF output level.

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NOTE

Perform turn-on procedures listed in paragraph 3-9, and prepare the Model 8708A Synchronizer for use with the Model 608F using procedures contained in Section III of the Operating and Service Manual for Synchronizer 8708A.

- 1. Align cursor knob with panel marker.
- 2. Align FINE FREQ control with panel marker.
- Set FREQUENCY RANGE switch to desired range (frequency limits for each range shown on MEGACYCLES dial housing).
- 4. Set MODULATION switch to the CW position.

CAUTION

Do not connect any source of RF or dc power to the RF OUTPUT connector on the Signal Generator. To do so may permanently damage the impedance-matching network in the Attenuator Section. Particular care must be observed when testing transceivertype equipment to insure that the transmitter remains inoperative while the equipment is connected to the signal generator (refer to NOTE in paragraph 3-7).

5. Connect the equipment under test to the RF OUTPUT connector observing that the signal generator is terminated in the proper load (refer to paragraph 3-8).

NOTE

If deemed necessary, conduct power output calibration procedures for the selected frequency range as listed in procedural steps d. through h. in paragraph 3-15.

6. Tune the Frequency Control until MEGA-CYCLES dial cursor is aligned with frequency to be used as indicated on the dial face.

NOTE

To crystal calibrate the frequency, refer to procedures listed in paragraph 3-13.

- 7. Depress and tune AMPL TRIMMER control for maximum indication on RF OUTPUT meter.
- 8. Adjust RF OUTPUT control until RF OUTPUT meter pointer is aligned with +7 dBm mark (ATTENUATOR CALIBRATED).

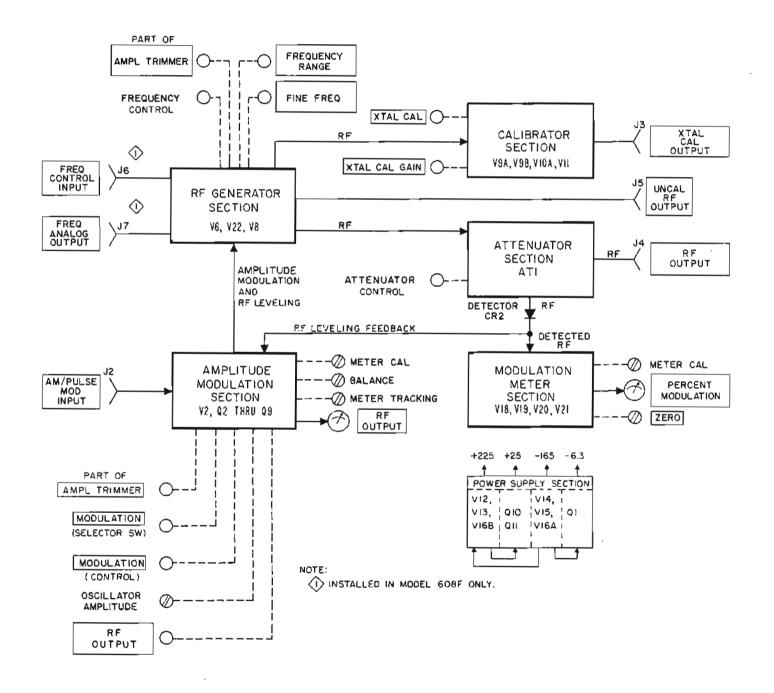
NOTE

Accuracy of the Attenuator Control setting and PERCENT MODULATION meter are within specifications only when the RF OUT-PUT meter is indicating ATTENUATOR CALIBRATED.

- 9. Adjust Attenuator Control for desired output level.
- Connect coaxial cable between UNCAL RFOUT-PUT connector on the Model 608F and the RF INPUT connector (beneath lighted lamp) on the Model 8708A Synchronizer.
- 11. Connect coaxial cable between FREQ ANALOG OUTPUT connector on the Model 608F and the FREQ ANALOG INPUT connector on the Model 8708A Synchronizer.
- 12. Connect coaxial cable between FREQ CONTROL INPUT connector on the Model 608F and the FREQ CONTROL OUTPUT connector on the Model 8708A Synchronizer.

NOTE

With the foregoing adjustments and connections completed phase-lock operation of the Model 608F can begin. Frequency or phase modulation of the RF carrier can be accomplished using the procedures listed in Figure 3-2 of the 8708A Manual.



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Figure 4-1. Overall Block Diagram, Models 608E and 608F

PRINCIPLES OF OPERATION

4-1. INTRODUCTION.

4-2. This section contains principles of operation for the Model 608E and Model 608F Signal Generators. Included herein is a general description of the instruments to a block diagram level, and a detailed circuit description which references the schematic diagrams. Principal dissimilarities between the Model 608E and 608F occur in the RF Generator Section; therefore, this section of each instrument will be described separately. The remaining sections in both instruments are nearly identical. These sections will be presented in a single description with the minor differences between instruments noted.

4-3. GENERAL DESCRIPTION.

4-4. As illustrated in Figure 4-1, the signal generators contain a RF Generator Section, Amplitude Modulation Section, Attenuator Section, Calibrator Section, Modulation Meter Section, and Power Supply Section. Each section is described in the succeeding paragraphs

4-5. RF GENERATOR SECTION, MODEL 608E.

4-6. The RF Generator Section in the Model 608E includer RF Oscillator V6, Buffer Amplifier V22, Power Amplifier V8, and associated circuitry. The RF Oscillator generates sine-wave signals from 10 MHz to 480 MHz in five frequency bands, which are identified in Figure 3-1 (item 8). Operation of the FREQUENCY RANGE switch selects tuned circuits in both the RF Oscillator and Power Amplifier stages for the desired operating range, and the Frequency Control adjustment selects frequencies within the range. The latter control tunes the RF Oscillator and Power Amplifier simultaneously. Minor frequency adjustments can be made by use of the FINE FREQ control connected in the output circuit of RF Oscillator V8. An uncalibrated RF output signal is also provided at the output circuit of Oscillator V6. The signal is applied to UNCAL RF OUTPUT connector J5 and is available for use with a frequency counter or other external measuring equipment. Buffer Amplifier V22 isolates RF Oscillator V6 and the output stages. A connection between the output stage of Buffer Amplifier V22 and the input circuit of Power Amplifier V8 serves as the modulation point, with the modulating signal being received from the Amplitude Modulation Section. The modulated RF signal is amplified in V8 and applied to the output circuit of the stage, which has been tuned by the FREQUENCY RANGE and Frequency Control adjustments. The AMPL TRIMMER control connected in the output circuit of Power Amplifier V8 is mechanically linked to a pushbutton switch in the Amplitude Modulation section. The interconnected control permits peaking the

RF output of the instrument, and maintaining a fixed RF output across the frequency range of the instrument. Tuning the control without engaging the pushbutton provides a fine-tune adjustment, which is used in conjunction with the RF OUTPUT control to maintain a fixed, leveled RF output regardless of the selected frequency. Depressing the control and then tuning permits a sharp peak to be obtained on the RF OUTPUT meter for optimizing the RF output of Power Amplifier V8.

4-7. RF GENERATOR SECTION, MODEL 608F.

The RF Generator Section in the Model 608F 4-8 Signal Generator is essentially the same as the Model 608E, containing the RF Oscillator, Buffer Amplifier, and Power Amplifier stages discussed in paragraph 4-5. Frequency capability of the instrument is between 10 MHz and 455 MHz in five frequency ranges. Differences between the frequency capabilities of the Model 608F and 608E can be ascertained by reference to Figure 3-1 (item 8). Two circuits and associated input and output connectors have been added to the Model 608F to permit the unit to be used with a Model 8708A Synchronizer. This equipment arrangement provides a highly stable, phase-locked RF signal output from the Model 608F, or permits the instrument to be used in narrow-band frequency or phase modulation applications. During use with a synchronizer a connection is made between UNCAL RFOUTPUT connector J5 on the Model 608F and synchronizer to provide a sample to the uncalibrated RF signal output from RF Oscillator V6. A second connection is made between the synchronizer and FREQ CONTROL INPUT connector J6 on the Model 608F. The latter connection delivers dc control voltage to varactors connected across the tank circuit of RF Oscillator V6 in the Model 608F. This dc control voltage reflects any frequency difference between the uncalibrated RF signal output of V6 and a calibrated reference frequency developed in the synchronizer. Hence, any drift in the output of RF Oscillator V6 is detected and an offsetting control voltage is returned to the Model 608F to maintain a correct frequency output. A third connection between the Model 608F and synchronizer is made at FREQ ANA-LOG OUTPUT connector J7. This connection provides a resistance to the synchronizer that is inversely proportional to the frequency in use. The resistance is used to maintain constant loop gain across the frequency band during phase-lock operation.

4-9. AMPLITUDE MODULATION SECTION.

4-10. The Amplitude Modulation Section consists of Modulation Oscillator V2, Differential Amplifier Q2 and Q3. Sine-Wave Amplifier Q4, Pulse Amplifier Q5, Amplifiers Q6 and Q7, Modulators Q8 and Q9, the RF OUTPUT meter, and associated circuits. Selection of the signal generator mode of operation is made by use of the MODULATION selector switch connected in the Amplitude Modulation Section. The four modes of operation are internal sine-wave modulation (400 or 1000 Hz), external sine-wave modulation, external pulse modulation, and cw operation; these four operations are discussed in the following paragraphs.

4-11. INTERNAL SINE-WAVE MODULATION. To select internal modulation, the MODULATION selector switch is set to either the 400 ~ or 1000 ~ position. Modulation Oscillator V2 provides a sine-wave output of fixed amplitude established by the Oscillator Amplitude control, a screwdriver adjustment within the instrument. The sine-wave output is then applied totransistor Q3 in the differential amplifier stage, and to AM/PULSE MOD INPUT connector J2. The signal at the connector is made available for synchronizing external equipment during internal modulation operation. The second signal applied to Differential Amplifier Q2-Q3 is a detected RF output signal received from diode CR2. This leveling feedback signal, which is used to counteract variations in the RF output signal, is applied to transistor Q2. The product of Differential Amplifier Q2-Q3 is amplified in Sine-Wave Amplifier Q4, and Amplifiers Q6 and Q7, then passed to Modulators Q8 and Q9. During internal and external modulation and when CW operation is conducted, Pulse Amplifier Q5 is disabled. Signal output from the modulator stage is applied to the input circuit of Power Amplifier V8 in the RF Generator Section where modulation of the RF signal and signal leveling occur. A MODULATION control between Modulator Oscillator V2 and transistor Q3 in the differential amplifier stage provides an adjustment to select the desired percentage of modulation as observed on the PER-CENT MODULATION meter discussed in paragraph 4-19. The RF OUTPUT control connected in the differential amplifier stage is used in conjunction with the Attenuator Control to calibrate RF power output from the instrument. Final selection of the desired output level up to +7 dBm is made using the Attenuator Control in both the Model 608E and 608F. In the Model 608E only, an RF output level above +7 dBm is selected by use of the RF OUTPUT control only. The AMPL TRIMMER control in the Amplitude Modulation Section is ganged to a tuning control in the output circuit of Power Amplifier V8 in the RF Generator Section. Operation of these controls is discussed in paragraph 4-5. The Meter Cal, Balance, and Meter Tracking controls are internal adjustments to set the RF OUTPUT meter and differential amplifier stage for proper operation.

4-12. EXTERNAL SINE-WAVE MODULATION. During external sine-wave modulation a connection is made between the external modulating equipment and AM/PULSE MODE INPUT connector J2 on the instrument, and the MODULATION selector switch is set to the AM position. The latter adjustment disables Modulation Oscillator V2, and routes the incoming modulation signal through the MODULATION control to transistor Q3 in the differential amplifier stage. Operation of the remaining circuits and controls in the Amplitude Modulation Section is identical to that described in paragraph 4-11. 4-13. EXTERNAL PULSE MODULATION. To set the signal generators for external pulse modulation, the desired RFoutput level and operating characteristics of the Amplitude Modulation Section are established with the MODULATION selector switch in the PULSE/ SET RFOUTPUT position. The switch position enables Pulse Amplifier Q5 and disables Modulation Oscillator V2 and Sine-Wave Amplifier Q4. With the MODULA-TION switch in the PULSE/SET RF OUTPUT position, the operating characteristics of Amplifier Q7 and Modulators Q8-Q9 in the Amplitude Modulation Section are established by use of the RF OUTPUT control, and the desired RFoutput level from the instruments selected as observed on the RF OUTPUT meter. Fine adjustment of the RF output level can be made at this time using the AMPL TRIMMER and RF OUTPUT controls conjunctively. Upon completion of these preliminary adjustments, the MODULATION selector switch is set to the PULSE position, and the external pulse source connected to the signal generators at AM/PULSE INPUT connector J2. The input signals are amplified by transistors Q5, Q6 and Q7 then passed to Modulators Q8-Q9. The output of the Modulator stage is applied to the input circuit of Power Amplifier V8 where modulation of the RF signal occurs. During pulse modulation, the MODULATION, RF OUTPUT and AMPL TRIMMER controls should not be operated after initial adjustments are made, and the PERCENT MODULATION and RF OUTPUT meters are inoperative. To readjust the RFoutput level, the external modulating source must be removed at connector J2, and the MODULATION switch returned to the PULSE/SET RF OUTPUT position. Adjustment of the RF OUTPUT and AMPL TRIMMER controls can then be made as previously discussed.

4-14. CW OPERATION. When the instrument is set for CW operation with the MODULATION selector switch in the CW position, Modulation Oscillator V2 and transistor Q3 in the differential amplifier stage are disabled. The RF leveling signal received from diode CR2 is passed through the Amplitude Modulation Section in a normal manner, and applied to the input circuit of Power Amplifier V8 in the RF Generator Section for stabilization of the RF power output level. The RF leveling signal is also applied to the RF OUT-PUT meter circuit in the Amplitude Modulation Section for RF level calibration. Operation of both the AMPL TRIMMER PUSH TO PEAK and RF OUTPUT controls is identical to that described in paragraph 4-11.

4-15. ATTENUATOR SECTION.

4-16. Attenuator Section AT1 couples RF energy from Power Amplifier V8 in the RF Generator Section to the load through RF OUTPUT connector J4, and provides control over the RFoutput level. The previously discussed RF leveling signal developed across diode CR2 is provided by a second RF output from the Attenuator Section. The Attenuator Control is used in conjunction with the RF OUTPUT meter and RF OUT-PUT control in the Amplitude Modulation Section to calibrate the RF output level of the instrument, and also to select the desired output level from the equipment after calibration procedures have been completed.

4-17. CALIBRATOR SECTION.

4-18. The Calibrator Section consists of 5-MHz Oscillator V9A, 1-MHz Oscillator V9B, Amplifiers V10A and V11, and the associated circuits. Operation of the oscillators is controlled by the XTAL CAL toggle switch that has three positions; OFF, 1 MC, and 5 MC. With the selector switch in either the 1 MC or 5 MC position, an output from the active oscillator is mixed with a sample RF signal from Power Amplifier V8 in the RF Generator Section to produce audio beat signals. The audio signals are passed through amplifiers V10A and V11 to a matching transformer. The transformer output is then applied to XTAL CAL OUTPUT connector J3 for use by an operator during frequency calibration operations. A XTAL CAL GAIN control connected in the output circuit of Amplifier V11 provides an audio output level adjustment.

4-19. MODULATION METER SECTION.

4-20. The Modulation Meter Section consists of stabilized wideband amplifiers V18 and V19, rectifier V20, and twin-triode V21 connected in a bridge circuit. Prior to operation the PERCENT MODULA-TION meter, also connected in the bridge circuit, is calibrated by use of the Meter Calibration control, an internal adjustment. Also, the bridge circuit is balanced to electrical zero by the ZERO adjustment located on the front panel of the instrument. The RF signal output of the Attenuator Section is detected across diode CR2 and applied to a filter network that produces the modulating signal. This signal is amplified in V18 and V19, rectified in V20, and applied to the bridge circuit. The bridge circuit is unbalanced in direct proportion to the peak amplitude of the input modulating signal; therefore, the PERCENT MODU-LATION meter connected across the bridge indicates the degree of unbalance in the bridge, or the percentage of modulation. The percentage of modulation is established by use of the MODULATION control in the Amplitude Modulation Section.

4-21. POWER SUPPLY SECTION.

4-22. The Power Supply Section consists of four series-regulated interrelated power supplies. The +225-volt power supply contains Comparison Amplifier V12 and a Series Regulator comprised of both sections of V13 and V16B. The +25-volt power supply contains Reference Amplifier Q10 and Series Regulator Q11. The -165 -volt power supply consists of Reference Tube V15, Comparison Amplifier V14, and Series Regulator V16A. The -6.3 volt supply contains a reference diode and Series Regulator Q1. The -165 volt supply is used to reference both the -6.3 volt and +225 volt supplies. The +225 volt supply is used to reference the +25 volt supply. A further discussion of these power supplies is contained in the detailed circuit description that follows.

4-23. DETAILED DESCRIPTION.

4-24. This discussion describes the circuits in the RF Generator Section (Model 608E), RF Generator

Section (Model 608F), Amplitude Modulation Section, Attenuator Section, Calibrator Section, Modulation Meter Section, and the Power Supply Section. Figures 7-4 and 7-5 will be used as reference during the description, and where transition from one diagram to the second is required appropriate notice will be made.

4-25. RF GENERATOR SECTION, MODEL 608E.

4-26. The RF Generator Section in the Model 608E includes RF Oscillator V6, Buffer Amplifier V22, and Power Amplifier V8. RF Oscillator V6 is a type 4042 triode connected across the +225 and -165 Vdc supplies in a Colpitts circuit. The oscillator tank circuit contains split-stator capacitor C17A and five separate RF transformers mounted on a revolving turret assembly. Capacitor segment C17A is mechanically linked to segment C17B in the output circuit of Power Amplifier V8, and the oscillator turret assembly is linked to an amplifier turret assembly also located in the output circuit of V8. Choice of frequency band is made by operating the FREQUENCY RANGE switch located on the front panel of the instrument. This function rotates the oscillator and power amplifier turret assemblies simultaneously to the selected range (Range A through E), and sets the associated inductor (L1 through L6, and L12 through L16) into the output circuits of RF Oscillator V6 and Power Amplifier V8 respectively, Capacitors C17A and C17B are then tuned by the front-panel mounted Frequency Control to the desired frequency within the range. Minor frequency adjustments can be made by tuning capacitor C68 (FINE FREQ control) in the tank circuit of Oscillator V6.

4-27. The plate circuit of RF Oscillator V6 is series fed through resistors R246 and R247. These resistors in conjunction with resistor R43 in the cathode of the stage serve to limit plate current within safe limits. The cathode of Oscillator V6 is bypassed by capacitor C25, which is part of the tube mounting plate and is not visible when the plate is in position. Bias for the stage is developed by cathode resistor R249 and grid resistor R248, with coupling capacitor C16 returning feedback to the grid to maintain oscillation. Capacitor C59 connected across a portion of the tuned circuit is used to calibrate the high end of the frequency dial after replacing an oscillator tube. Capacitor C18, installed in the Model 608E only, is used to improve tracking characteristics of the tuned circuit at higher frequencies. Adjustment of this internal control is critical, since a compromise must be reached between tracking capabilities and optimum power output. Misadjustment of capacitor C18 can produce improved tracking capability but a loss in RF power output. The control is factory aligned, and no further adjustment should be required.

4-28. All voltages applied to RF Oscillator V6 located in the RF Generator housing assembly are regulated, and brought into the stage through RFI filters. The +225 Vdc plate voltage is received through filter FL5, the -165 Vdc supply is received through filter FL2, and a regulated -6.3 heater voltage is received through filter FL4. 4-29. The secondary windings in the oscillator turret assembly couple RF energy from RF Oscillator V6 through a coaxial cable to Buffer Amplifier V22. Resistors R240, R242, R244, and R245 located in the secondary windings of coils L2 through L5 damp undesirable resonance in the cathode of Buffer Amplifier V22. Resistors R241 and R243 connected across the secondary windings of coils L4 and L5 limit the RFdrive at lower frequencies. Resistors R250, R251, and capacitor C218 form a take-off network for an uncalibrated RF output signal that is applied through coaxial cable W1 to UNCAL RF OUTPUT connector J5. Resistor R251 in the take-off network is 301 ohms in the Model 608E and 200 ohms in the Model 608F. This configuration is due to the difference in RF output level between instruments. The RF signal at connector J5 can be used with a frequency counter or other external measuring equipment. When the Model 608F is used with a Model 8708A synchronizer during phase-lock operation, the signal is used by the synchronizer.

4-30. Buffer Amplifier V22 is a type EC-8010 triode connected as a grounded-grid amplifier. The stage isolates RF Oscillator V6 and Power Amplifier V8, and reduces any incidental frequency modulation to a minimum level. Cathode resistors R255 and R253 operate with resistors R254 and R256 in the plate circuit to limit plate current to a safe value. Power to the stage is supplied through RFI filters FL2, FL4, and FL5, with the plate, cathode and filament supplies being decoupled by capacitors C222, C220, and C221 respectively. The plate circuit of Buffer Amplifier V22 is coupled to the cathode circuit of Power Amplifier V8 through a wide-band network consisting of dc blocking capacitor C224 and coil L22. Coil L20 in the plate circuit of V22 serves as an RF choke, while isolation network L11 blocks an RF signal path to the output circuit of the Amplitude Modulation Section.

4-31. Power Amplifier V8 is a grounded-grid, cathode-modulated amplifier that delivers the modulated or CW output signal to the Attenuator Section. The plate circuit of V8 contains the amplifier turret assembly which houses inductors L12 through L16. These inductors are placed into the output circuit of V8 by operation of the FREQUENCY RANGE switch discussed in paragraph 4-26. The second segment of Frequency Control capacitor C17 is also located in the plate circuit of Power Amplifier V8. Linkage between capacitor C17B and C17A in the RF Oscillator Stage permit tuning to a frequency within a range as discussed in paragraph 4-26. A second linkage between capacitor C17B and the AMPL TRIMMER control permit the capacitor to be funed separately or in conjunction with push-to-operate switch S6 in the Amplitude Modulation Section for peaking the RF output signal. This function is discussed further in paragraph 4-45.

4-32. When the instrument is set to provide an amplitude modulated output, the modulating signal and an RF leveling signal are received at the cathode of Power Amplifier V8 from the Amplitude Modulation Section through isolation network L11. During CW operation, only the RF leveling signal is received from the Amplitude Modulation Section, Diode CR7 connected in the signal path prevents damage to Power Amplifier V8 by limiting the negative swing of the cathode. Capacitor C32 is connected in the plate circuit of V8 in the Model 608F only, and is adjusted for improved tracking characteristics in the power amplifier stage at higher frequencies. Capacitors C30, C229, C230, and C231 connected across coils L16, L15, L14, and L13 respectively are also installed in the Model 608F only. These capacitors are selected intest to improve tracking between the power amplifier and oscillator stages. Nominal values for the capacitors are illustrated in Figure 7-4. The RF output signal from Power Amplifier V8 is inductively coupled through the amplifier turret assembly coil and applied to the Attenuator Section. This section of the instrument is discussed in paragraph 4-51.

4-33. RF GENERATOR SECTION, MODEL 608F.

4-34. The RF Generator Section in the Model 608F is essentially the same as the generator section in the Model 608E, except for the addition of circuits to permit the unit to be used with a Model 8708A Synchronizer. These circuit differences occur in the tank circuit of RF Oscillator V6 and the output stage of Power Amplifier V8.

4-35. Two varactor diodes, CR31 and CR32, are connected across the oscillator tank circuit in the Model 608F. The varactors are reverse-biased junctions that act as variable capacitors, changing capacitance with a change in the applied reverse bias. In quiescent condition the diodes are biased by a voltage divider network comprised of resistors R271, R267, and R268. A change in bias, and the resultant change in capacitance, is produced by a dc control voltage received at FREQ CONTROL INPUT connector J6. During operation with a Model 8708A Synchronizer, a connection is made between UNCAL RF OUTPUT connector J5 and the synchronizer to provide a sample of the RF output from the Model 608F. A second connection is made between the synchronizer and FREQ CONTROL INPUT connector J6 to receive the dc control voltage from the synchronizer. The frequency sample sent to the synchronizer is compared to a reference frequency developed in the synchronizer. A drift tendency in RF Oscillator V6 is compared with the reference frequency, and translated into a dc control voltage for application to varactors CR31 and CR32 in the Model 608F. The varactors change capacitance in direct proportion to the detected drift holding RF Oscillator V6 on frequency.

4-36. Potentiometer R269 is installed in the Model 608F only, and is mechanically linked to Frequency Control capacitor C17B in the output circuit of Power Amplifier V8. The potentiometer output resistance varies as the Frequency Control adjustment is tuned, presenting 4K ohms at the low-frequency end of the capacitor tuning range and zero ohms at the highfrequency end. The resistance is applied through FREQ ANALOG OUTPUT connector J7 to the dc control voltage circuit in the synchronizer. This loop arrangement controls the voltage level applied to varactors CR31 and CR32, thereby assuring correct operating conditions for the devices across a frequency range.

4-37. AMPLITUDE MODULATION SECTION.

4-38. The Amplitude Modulation Section consists of Modulation Oscillator V2, a portion of the Modulator and ALC Board, a RF OUTPUT meter and control, two MODULATION controls, and the associated circuitry. The Modulator and ALC Board contains Differential Amplifier Q2-Q3, Sine-Wave Amplifier Q4, Pulse Amplifier Q5, Amplifiers Q6 and Q7, and Modulators Q8 and Q9. Four modes of operation will be discussed during the description of the Amplitude Modulation Section; these are identified as internal sinewave modulation, external sine-wave modulation, external pulse modulation, and CW operation.

4-39. INTERNAL SINE-WAVE MODULATION. The internal sine-wave modulation mode is selected by placing MODULATION switch S7 to either the 400 ~ or 1000 ~ position (switch setting 2 or 3), which provides a 400 ± 40 Hz or a 1000 ± 100 Hz output from Modulation Oscillator V2. The oscillator consists of resistance-coupled amplifiers V2A and V2B connected in a Wein-Bridge circuit. At the resonant frequency, positive feedback from the plate of Amplifier V2B is applied to the grid of amplifier V2A sustaining oscillation. A change in frequency is obtained by inserting resistors R3-R5 or R4-R6 into the feedback network through two segments of MODULATION switchwafer 2F. A negative feedback circuit consisting of resistors R7, R8 and lamp DS1 is used to stabilize the oscillator, reduce distortion, and maintain a constant output level. If the amplitude of the signal increases, current through the lamp increases thereby raising the resistance of lamp DS1; consequently, the signal amplitude is held constant. Potentiometer R7 is an internal adjustment for adjusting the amplitude of the output signal. Heater voltage to the stage is applied at all times during instrument operation; however, plate voltage is applied to V2A and V2B only when MODULATION switch-wafer 2R is in position 2 or 3.

4-40. The sine-wave signal from Modulation Oscillator V2 is applied through dc blocking capacitor C205, MODULATION switch-wafer 1F, to MODULATION control R210. A second signal path for the sine-wave is provided through resistor R262 to AM/PULSE MOD INPUT connector J2 where the signal is available for synchronizing external equipment during the internal modulation mode. The signal at the wiper of MODU-LATION control R210 is passed through, a frequencycompensation network consisting of resistor R261, and capacitors C203 and C204. The signal is then routed through MODULATION switch-wafer 1R to the base of transistor Q3 in the differential amplifier stage. The second input to the differential amplifier is a detected signal developed across diode CR2 and filter network C38 and R119 located in the bolometer portion of the Attenuator Section. This leveling signal, which is representative of the RF power output from the instrument, is passed through RFI filter FL8 to the base of transistor Q2. The leveling signal is also applied through resistor R201, Meter Cal potentiometer R202.

MODULATION switch-wafer 4F, and AMPL TRIM-MER pushbutton switch S6 to RF OUTPUT meter M1. Operating characteristics of switch S6 and the RF OUTPUT meter are contained in paragraph 4-45.

4-41. Balance potentiometer R203 is a board adjustment in the emitter circuit of Differential Amplifier Q2 and Q3 used to set the stage for proper operation. The RF OUTPUT control R212 connected with resistor R213 in avoltage divider network sets the operating characteristics of the differential amplifier stage, and is used in conjunction with the Attenuator Control to calibrate and control the RF output from the instrument. The output of the differential amplifier is the product of the leveling signal on the base of transistor Q2 and the modulating signal on the base of transistor Q3. This product signal is then applied to the base of Sine-Wave Amplifier Q4.

4-42. The base circuit of amplifier Q4 contains a lowfrequency compensating network consisting of resistor R214 and capacitor C202. In the Model 608E resistor R214 is 200 ohms, and in the Model 608F the resistor value is 392 ohms. Amplifier Q4 is biased by voltage divider network R217 and R215, with diode CR21 installed as a protective device to maintain a safe relationship between the base and emitter circuits of Q4 during operation. The collector circuit of Q4 also contains two frequency-compensating networks consisting of resistor R219 and capacitor C208, and resistor R218 and capacitor C207. The output of amplifier Q4 is developed across resistor R216 and applied to the base circuit of Amplifier Q6.

4-43. During internal or external sine-wave modulation, Amplifier Q6 receives the leveling and modulating signals from Amplifier Q4. During CW operation, only the leveling signal is received at Amplifiers Q4 and Q6. During external pulse modulation, Amplifier Q4 is inhibited and the input pulse signals are received at Amplifier Q6 from Pulse Amplifier Q5. The external pulse modulation and CW modes of operation are discussed later in this section. The emitter of Amplifier Q6 is biased by a voltage divider network consisting of resistors R223 and R224, and during internal sine-wave operation the modulating and leveling signals are developed across load resistor R222 for application to Amplifier Q7.

4-44. Resistor R229 provides emitter bias for Amplifier Q7, while the collector voltage is developed across load resistor R228 and resistor R233. This voltage divider network also establishes the base conditions on Modulators Q8 and Q9. During the pulse modulation mode, the collector voltage on Amplifier Q7 can be controlled by operation of the RF OUTPUT control; this function is discussed in paragraph 4-47. The collector output of Amplifier Q7 is applied to the base of Modulators Q8 and Q9 connected in parallel emitterfollower configuration. The emitter bias for the stage is derived from a voltage divider network consisting of resistor R33, Zener diode CR26, and resistor R258 in the RF Generator Section. Diode CR25 is connected as a protective device between the base and emitter circuits of the stage. The modulator output is applied through RFI filter FL6 and isolation network L11 to the cathode of Power Amplifier V8, where modulation and RF leveling occur.

4-45. A feedback path for the output of Modulators Q8 and Q9 is provided through filter network R235 and capacitor C200 which removes the modulating signal. The leveling signal is passed through MODULATION switch-wafer 4F and applied to the normally-open contacts of AMPL TRIMMER pushbutton switch S6 when the MODULATION switch is in positions 1 through 4. As previously mentioned, the AMPL TRIMMER pushbutton is mechanically linked to AMPL TRIM-MER capacitor C17B in the output circuit of Power Amplifier V8. During normal operation the detected feedback from diode CR2 in the Attenuator Section is applied through resistor R201, potentiometer R202, and the normally-closed contacts of pushbutton switch S6 to RF OUTPUT meter M1: therefore, the meter normally displays the leveled RF output of the instrument. Rotation of RF OUTPUT control R212 changes the conduction level of Differential Amplifier Q2-Q3 and the signal output level of Modulators Q8-Q9. This change is reflected in the RF output of Power Amplifier V8, and displayed on RF OUTPUT meter M1. Operation of AMPL TRIMMER control C17B without engaging the pushbutton provides a fine-tune adjustment for use with the RF OUTPUT control to maintain a constant, leveled RF output across a frequency range. When AMPL TRIMMER pushbutton S6 is depressed the leveling signal feedback from Modulators Q8-Q9 is displayed on meter M1, and operation of the control permits a sharp peak to be observed on the meter. Hence, the RF output of Power Amplifier V8 can be peaked. During internal and external sine-wave operation, MODULATION control R210 controls the level of the modulation and is used in conjunction with the PERCENT MODULATION meter discussed in paragraph 4-59.

4-46. EXTERNAL SINE-WAVE MODULATION. Operation of the instruments during external sine-wave modulation is similar to the internal mode, and is established by setting MODULATION switch S7 to the AM position (switch position 4). Modulation Oscillator V2 is disabled by removing the stage plate voltage at MODULATION switch-wafer 2R, and a signal path of the external sine-wave signal is established through MODULATION switch-wafer 1F. A cable connection is made between the external source and AM/PULSE MOD INPUT connector J2, with the incoming signal being routed to MODULATION control resistor R210. The signal is then applied through the frequencycompensation network and MODULATION switch-wafer 1R to the base of transistor Q3 in the differential amplifier stage. Operation of the remaining circuits and controls in the Amplitude Modulation Section is the same as described during internal sine-wave modulation.

4-47. EXTERNAL PULSE MODULATION. During this mode of operation MODULATION switch S7 is set first to the PULSE/SET RF OUTPUT position (switch position 5) to establish operating conditions in the output circuit of the Amplitude Modulation Section, and

consequently the RF output level of the instrument. The switch is then set to the PULSE position (switch position 6) to pulse modulate the RF carrier. In both positions Modulation Oscillator V2 is disabled and the input signal path through MODULATION control resistor R210 is interrupted. The position of MODU-LATION switch-wafer 4F interrupts the leveling signal path through the normally-open contacts of AMPL TRIMMER pushbutton switch S6, and removes the push-to-peak function of the AMPL TRIMMER control during pulse operation. However, the detected signal from diode CR2 in the Attenuator Section is received through resistor R201 and potentiometer R202 for application to RF OUTPUT meter M1. Therefore, in the PULSE/SET RF OUTPUT position, the AMPL TRIMMER and RF OUTPUT controls can be used to set the desired RFoutput level as indicated on the RF OUTPUT meter. The two controls are operated only in the PULSE/SET RF OUTPUT position.

4-48. Transistor Q3 in the differential amplifier stage has an inhibiting voltage applied to the base through resistor R211 and MODULATION switch-wafer 1R. With transistor Q3 cut off, transistor Q2 conducts heavily and places a disabling voltage on the base of Amplifier Q4. This condition interrupts to leveling signal path through the Amplitude Modulation Section. Resistor R212 is removed from the circuit by the position of MODULATION switch-wafer 1R; however, the RF OUTPUT control is also connected to potentiometer R226 in the collector circuit of Amplifier Q7 for rf output level control. With the MODULATION switch in position 5, a positive voltage is applied to the base of Pulse Amplifier Q5 through resistors R220 and R225 to enable the amplifier. The position of MODULATION switch-wafer 3R forms a voltage divider network consisting of resistors R228, R226, and R270 in the collector circuit of Amplifier Q7 and the base circuits of Modulators Q8 and Q9. Operation of RF OUTPUT potentiometer R226 controls the conduction level of transistors Q7, Q8, and Q9, and the output level of Power Amplifier V8.

4-49. While the MODULATION switch is still in the PULSE/SET RF OUTPUT position, the RF OUTPUT control is adjusted for an indication on RF OUTPUT meter M1. AMPL TRIMMER control C17B in the output stage of Power Amplifier V8 is then peaked for maximum RF output. The RF OUTPUT is readjusted for an indication of +7 dBm (Attenuator Calibrated) on RF OUTPUT meter M1. With the preliminary adjustments completed, the external modulating equipment is connected to AM/PULSE MOD INPUT connector J2, and MODULATION switch S7 is set to the PULSE position (switch position 6). The incoming pulse train is passed through MODULATION switch-wafer 3F to the base of Pulse Amplifier Q5, and then routed through Amplifier Q6 and the remaining circuits of the Amplitude Modulation Section as previously described.

4-50. CW OPERATION. To set the equipment for CW operation, MODULATION switch S7 is placed to the CW position (switch position 1). Modulation Oscillator

V2 is removed from the circuit by the position of switch-wafer 2R, and the sine-wave input circuit is interrupted by the setting of switch-wafer 1F. A detected RF output signal from diode CR2 in the Attenuator Section is applied to the base of transistor Q2 in the differential amplifier stage for the development of a leveling signal as previously described. RF OUT-PUT meter M1 also receives the detected signal through AMPL TRIMMER pushbutton switch S6. Transistor Q3 in the differential amplifier stage is terminated in a network consisting of resistor R263 and capacitor C206; however, RF OUTPUT control resistor R212 is connected in the circuit and controls conduction of the differential amplifier stage. Pulse Amplifier Q5 is disabled by the ground applied to the base circuit through MODULATION switch-wafer 3F. Leveling signal output of the differential amplifier stage is passed through Amplifiers Q4, Q6, Q7 and Modulators Q8-Q9 to the input circuit of Power Amplifier V8. RF OUTPUT meter M1 displays a leveled RF output from the instrument as selected by the RF OUT-PUT control and Attenuator Control. AMPL TRIM-MER pushbutton switch S6 is returned to operation in conjunction with AMPL TRIMMER capacitor C17B for peaking the RF output using the leveling signal feedback from Modulators Q8-Q9.

4-51. ATTENUATOR SECTION.

4-52. To extract power from the RF Generator Section, a piston-type attenuator is used. The attenuator housing projects through the rear of the generator housing, and provides an open-ended termination near the power amplifier plate-circuit inductor. A singleturn pick-up loop at the end of the attenuator probe couples energy to an impedance-matching network consisting of capacitors C69 and C37, and resistors R58 and R59. Capacitor C37 is a movable slug in the probe body that is used to adjust the standing-wave ratio to a minimum value when RF OUTPUT connector J4 is terminated in a 50-ohm load.

4-53. The front-panel mounted Attenuator Control moves the probe in the attenuator housing, and controls the amount of RF energy passed to the output connector. In both the Model 608E and 608F, calibration is accomplished by adjusting the RF OUTPUT and AMPL TRIMMER controls for a steady +7 dBm (ATTENUATOR CALIBRATED) indication on the RF OUTPUT meter across the frequency range in use. An adjustment of the Attenuator Control after calibration procedures are completed provides the RF output level indicated on the control dial as long as the RF OUTPUT meter needle remains in the ATTENUA-TOR CALIBRATED position. In the Model 608E only, an rf output between +7 dBm and +13 dBm can be selected. The Attenuator Control is set to the +7 dBm position (maximum clockwise), and the RF OUTPUT and AMPL TRIMMER controls are adjusted for the RF OUTPUT meter indication of +7 dBm as previously discussed. The desired output level above +7 dBm is then selected by operation of the RF OUTPUT control only, and the output is read directly from the RF OUT-PUT meter.

4-54. Two parallel wires crossing the open end of the attenuator housing are used to provide a sample of RF energy delivered by Power Amplifier V8. The energy is applied to the Bolometer through ferritebead RFC1, which provides isolation between the Bolometer and Attenuator. The Bolometer contains frequency-response coil L17, diode CR2 and resistor R60. The RFsignal detected across resistor R60 and diode CR2 is passed through an RF filter consisting of capacitor C38 and resistor R119. The detected signal is then applied through RFI filter FL8 to the Modulation Meter Section discussed in paragraph 4-60, and to transistor Q2 in the Amplitude Modulation Section discussed in paragraph 4-40.

4-55. CALIBRATOR SECTION.

4-56. The Calibrator Section consists of 5-MHz Oscillator V9A, 1-MHz Oscillator V9B, Harmonic Generator and Mixer CR1, Amplifiers V10A and V11, and output transformer T3. The oscillator and mixer circuits are illustrated in Figure 7-4, while the amplifiers and transformer are located on Figure 7-5.

4-57. The 5-MHz Oscillator is a crystal-controlled, electron-coupled oscillator utilizing the pentode half of a type 6U8 tube. The plate and screen circuits are tuned to the crystal frequency of 5 MHz by coils L6 and L7. Trimmer capacitor C23 is used for finefrequency adjustment of the oscillator. The 1 MHz Oscillator consists of the triode portion of the tube, with coil L18 used to adjust the oscillator to a point where the fifth harmonic is synchronized with the 5-MHz output of Oscillator V9A. When XTAL CAL switch S4 is set to the 5-MC position, only Oscillator V9A is active. In the 1-MC position both oscillators are operating, with Oscillator V9A referencing the 1-MHz oscillator.

4-58. The output signal from Oscillators V9A or V9B is passed across dc blocking capacitor C24 to diode CR1 which acts as a harmonic generator and mixer. The anode of diode CR1 is inductively coupled to the output circuit of Power Amplifier V8 by running a lead near the plate of the amplifier. Oscillator harmonics within the diode are mixed with RFoutput signals to produce audio-beat frequencies across resistor R50 in the grid circuit of Amplifier V10A. The output of amplifier V10A is again amplified in V11 and applied across capacitor C36 to 600-ohm, line-matching transformer T3. The secondary of transformer T3 is connected to XTAL CAL OUTPUT connector J3 where the calibrating signals can be monitored. XTAL CAL GAIN resistor R56 provides front-panel control of the audio level output at connector J3.

4-59. MODULATION METER SECTION.

4-60. The Modulation Meter Section consists of a detector network in the Bolometer portion of the Attenuator Section, Amplifiers V18 and V19, Rectifier

V20, and a balanced-bridge circuit containing PER-CENT MODULATION meter M2. The Bolometer is illustrated in Figure 7-4, and the remainder of the section appears in Figure 7-5.

4-61. The output of diode CR2, capacitor C38 and resistor R119 in the Bolometer is a dc signal which is representative of the RF output level, and the modulating signal when internal or external modulation is in process. The output is passed through RFI filter FL8 and applied to blocking capacitor C39, which couples the modulating signal to Amplifier V18. The output of Amplifier V18 is amplified again in V19 and passed to the plates of Rectifier V20. A stabilization feedback circuit connects the plate of Amplifier V19 and cathode of Amplifier V18, with resistor R99 installed in the circuit for frequency compensation. C55 ac couples the plate of V19 to the cathode of V18.

4-62. Rectifier V20 is connected in a balanced-bridge detector circuit consisting of twin-triode V21, PER-CENT MODULATION meter M2, and associated circuitry. With no modulating signal applied to the circuit. the steady-state dc potential at the plate of Amplifier V19 is coupled to the grids of V21A and V21B. In this condition, equal current flows on both sides of the bridge, and the PERCENT MODULATION meter reads zero. ZERO potentiometer R106 is a front-panel adjustment to compensate for variations in tube and component values. When a modulation mode of operation is in progress, the peak value of the rectified modulating signal is applied to the grid of V21B. The bridge is unbalanced accordingly, and produces an indication on the PERCENT MODULATION meter that is indirect proportion to the amplitude of the modulating signal. Triode V21A is unaffected during the detection cycle, since the modulating signal is filtered by resistors R100, R101, and capacitor C58.

4-63. POWER SUPPLY SECTION.

4-64. Four series-regulated power supplies are contained in the power supply section. The power supplies provide +225, -6.3, -165, and +25 volts dc to the Model 608 Signal Generators. The four supplies are connected to a common input source through ON switch S3. A slide switch located on the rear panel of the instrument sets the generators for use with either a 115-Vac or 230-Vac power source, with AC LINE fuse F1 connected between the power supplies and source as a protective device. The supplies are illustrated in Figure 7-5 and are described in the following paragraphs.

4-65. THE +225 VOLT SUPPLY. Input ac voltage to the +225-volt supply is received across transformer

T1. The secondary contains bridge rectifier CR8 through CR11 and filter capacitor C40. The supply is series regulated through triodes V13A, V13B, and V16B, with the output connected to the load through front-panel mounted DC 0. 25 AMP fuse F3. The -165 Volt Power Supply output is used as reference, being applied through a voltage divider consisting of resistors R70 through R72. Any change in the load is sensed across +225 Volts Adj potentiometer and applied to the control grid of Comparison Amplifier V12. The plate of amplifier V12 controls the conduction level of the series regulators, and holds the output voltage constant. The +225 volt supply is used to reference the +25 volt supply discussed in paragraph 4-69.

4-66. THE -6.3 VOLT SUPPLY. Input ac voltage to the -6.3-volt supply is also received across transformer T1. Rectification and filtering occur across bridge rectifier CR16 through CR19 and capacitors C75. Transistor Q1 operates as a comparison amplifier and series regulator, with the base circuit referenced by -7.5 Vdc developed from the -165 Vdc supply across resistor R260 and Zener diode CR20. The output voltage of 7.2 Vdc is applied to RFI Filter FL4 where 0.9 Vdc is dropped. The -6.3 volts is then used as a filament supply for RF Oscillator V6, Buffer Amplifier V22, and Power Amplifier V8. Filament voltage for the remaining tubes in the instrument is provided by a separate secondary winding of transformer T1. The latter winding also supplies voltage to a light behind the MEGACYCLES dial, and to the AC POWER light located on the front panel. Both lamps are lighted when switch S3 is in the ON position.

4-67. THE -165 VOLT SUPPLY. Input ac voltage to the -165 volt supply is received across transformer T1, and rectified by diodes CR12 through CR15. Triode V16A is the series regulator, with the grid controlled by the plate of Comparison Amplifier V14 whose cathode is referenced by gas tube V15. The voltage level is controlled -165 Volts Adj potentiometer R80, with the output applied to the load and to the +225 and -6.3 volt supplies for reference.

4-68. THE +25 VOLT SUPPLY. Components of the +25 volt supply are located on the Modulator and ALC Board, which also contains a portion of the Amplitude Modulation Section circuits discussed in paragraph 4-37. Ac power to the supply is received across transformer T2, and rectified by diodes CR29 and CR30. Transistor Q11 is the series regulator, with the conduction level being control by Amplifier Q10. The supply is referenced by +225 Vdc received across resistor R239.

SECTION V MAINTENANCE

5-1. INTRODUCTION.

5-2. This section contains inspection requirements, performance and test procedures, troubleshooting techniques, and repair instructions for the Model 608E and 608F Signal Generators. Generally, maintenance procedures for both instruments are identical and will be presented in one set of instructions. Where differences occur, separate instructions for each instrument will be included. The test equipment required for maintenance is listed in Table 5-1. Test instruments other than those listed may be used if the performance equals or exceeds the minimum specification listed. Two test instruments must be fabricated; these include a 100-ohm termination and loop antenna, which are illustrated in Figure 5-1.

5-3. PERIODIC INSPECTION.

5-4. CLEANING.

5-5. If the equipment has been subjected to unusual conditions (excessive moisture, dust, heat, vibration, etc.) it is suggested that the instrument be removed from the cabinet and inspected for dirt or moisture accumulation, loosened components, or any possible sign of damage. Forced air under medium pressure is recommended for dusting and drying, although care must be taken not to vary the settings of the internal-adjustment potentiometers and capacitors during the process.

5-6. LUBRICATION.

5-7. The signal generators are thoroughly lubricated at the factory. Full-shielded ball bearings are used in many applications and require no subsequent attention. Ball bearings that are not fully shielded require only light machine oil. After prolonged use of the instrument, lubricated points should be checked for excessive dust accumulation or drying of lubricant.

5-8. This maintenance is particularly important when the instrument is operated continuously or for several hours daily. Calibration accuracy and instrument dependability can be maintained only by proper lubrication of the two worm gears in the tuning mechanism. Other rubbing surfaces and sleeve bearings, including the small pulleys used in the attenuator drive system, should be lubricated quarterly with a light slideway oil such as Shell Tonna G. In all cases avoid over lubrication. The points to be lubricated are listed in Table 5-2 and illustrated in Figure 5-2.

5-9. PERFORMANCE TESTS.

5-10. PURPOSE.

5-11. The following paragraphs check performance of the signal generators during incoming inspection tests, periodic instrument evaluation. troubleshooting, and during equipment calibration periods. The tests can be performed without access to the interior of the instruments. The specifications listed in Table 1-1 are the performance standards. Unless otherwise specified in the test procedures, the following controls are set to the indicated position for each test:

MODULATION switch.	•	• •	•	•	•	•	•	•	•		CW
Cursor knob	•	Al	igne	ed	wit	:h	pa	ine	e1	ma	arker
FINE FREQ control .		Al	igne	ed	wit	h	pa	ine	el	ma	arker
XTAL CAL switch						•					OFF

5-12. FREQUENCY ACCURACY AND RESETTABILITY TEST.

a. Connect signal generator intest setup illustrated in Figure 5-3.

b. Set FREQUENCY RANGE switch to range A.

c. Adjust FREQ OUTPUT and attenuator controls to obtain proper triggering of electronic counter.

d. Adjust frequency control to the low end, approximate center and high end of range A; record MEGA-CYCLES dial setting and electronic counter indication at each position. Electronic counter should indicate within $\pm 1\%$ of MEGACYCLES dial setting.

e. Retune frequency control to each MEGACYCLES dial setting recorded in step d. Electronic counter should indicate within $\pm 0.1\%$ of indication recorded in step d.

f. Repeat steps d. and e. for FREQUENCY RANGE B through E.

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Instrument	Critical Specifications	Recommended Model
Electronic Counter	Frequency Range: 10 MHz to 445 MHz Accuracy: ±0.005%	HP 5245L with 5253B plug-in unit
Power Meter	Range: 10 μ W to 10 mW Accuracy: $\pm 5\%$ of full scale	НР 432А
Thermistor Mount	Frequency Range: 10 to 480 MHz SWR: 1.6 max	HP 478A
Coaxial Attenuator Coaxial Attenuator	Attenuation: 10 dB Attenuation: 20 dB	HP 8491A HP 8492A
Oscillator	Frequency Range: 20 Hz to 30 kHz Envelope Distortion: 0.2% max Output: 1 V rms Output Impedance: 600Ω	HP 200CD
VTVM	Voltage Range: 0.1 mV to 300V Decibel Range: -72 to +52 dB Frequency Range: 20 Hz to 6 MHz	HP 400E
Distortion Analyzer	Range: 5 Hz to 600 kHz Accuracy: ±3%	HP 334A
Deviation Meter	Frequency Range: 50 to 150 MHz Accuracy: $\pm 3\%$ Deviation Range: 0 to 5 kHz	Measurements Corp. Model 140
Power Supply	-2 to -32 V variable	HP 6217A
Oscilloscope	Vertical Bandwidth: 5 MHz Vertical Sensitivity: 5 mV/cm	HP 140A with 1402A and 1420A plug-in units
Adapter	None	UG-201A/U
Crystal Detector	Frequency Range: 10 MHz to 12.4 GHz VSWR: 1.2 max Deviation from Square Law: <0.5 dB over a range of at least 30 dB Input Impedance: 50Ω Impedance Output: 15KΩ, 10pf	HP 423A
Dummy Load	Resistance: $100\Omega \pm 10\%$ Wattage: $1/2$ W	See Figure 5-2
Pulse Generator	Pulse Width: $10\mu sec$ Pulse Amplitude: $10 v$ peak into $2K\Omega$ Pulse Rise and Decay: $0.02 \mu sec$ max Repetition Rate: 50 to 5000 pps	HP 214A
Loop Antenna	Special	See Figure 5-1
Spectrum Analyzer	Freq: 400 MHz Sensitivity: 1 μ V. Absolute calibration	HP 140S with 8554L and 8552A Plug-Ins.

Table 5-1. Test Equipment Required for Maintenance

Instrument	Critical Specifications	Recommended Model	
DC Ammeter	Range: 15 to 30 mA Accuracy: <u>+</u> 5%	HP 428B	
SWR Meter	Sensitivity: at least μv full scale Attenuator Range: 40 dB Attenuator Accuracy: $\pm 0, 1$ dB	HP 415E	
Double Balanced Mixer	Frequency Range: 10 to 480 MHz	HP 10514A	
Frequency Comb Generator	Frequency Range: 10 to 480 MHz Comb Frequency: 1 MHz	HP 8406A	
Filter	Cutoff Frequency: 100 kHz		
Frequency Meter/ Filter Kit	Discriminator Output: 100 kHz/V Disc. Out Accuracy: Better than 1% Internal Filter: 10 kHz, low pass	HP 5210A/10531A	
Oscilloscope	Bandwidth: 0 to 20 MHz Vert Sensitivity: 50 $\mu V/cm$	HP 140 with 1406A plug-in unit	
LAB Dual Range, DC Power Supply	Output: 0 to 20 Vdc	HP 6204B	
Electronic Counter	Resolution: 5 digits Function: Frequency to > 350 MHz Input Impedance: 50 ohms	HP 5246L with 5252A plug-in unit	
Synchronizer	Compatible with the 608F	HP 8708A	

Table 5-1. Test Equipment Required for Maintenance (Cont.)

LOOP ANTENNA

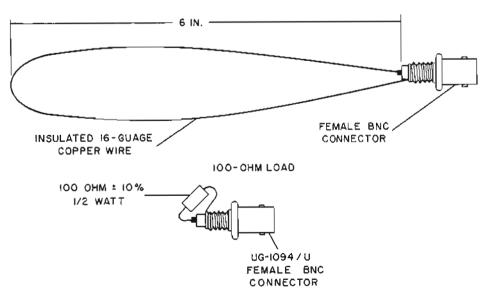


Figure 5-1. Fabricated Test Equipment

Model 608E/F

Table	5-2.	Lubrication	Chart

Lubrication Point (see Figure 5-2)	Lubricant	
Oscillator and amplifier worm gears	"Molykote (type Z)"	
AMPL TRIMMER stop mechanism	Light machine oil	
AMPL TRIMMER toggle nut	Light machine oil	
AMPL TRIMMER drive link	Light machine oil	
Amplifier worm drive shaft	Light machine oil	
RF Amplifier sliding coupler	Light machine oil	
Attenuator pulleys	Light machine oil	
Attenuator drive shaft bearing (front)	Light machine oil	
Attenuator drive shaft bearing (rear)	Light machine oil	

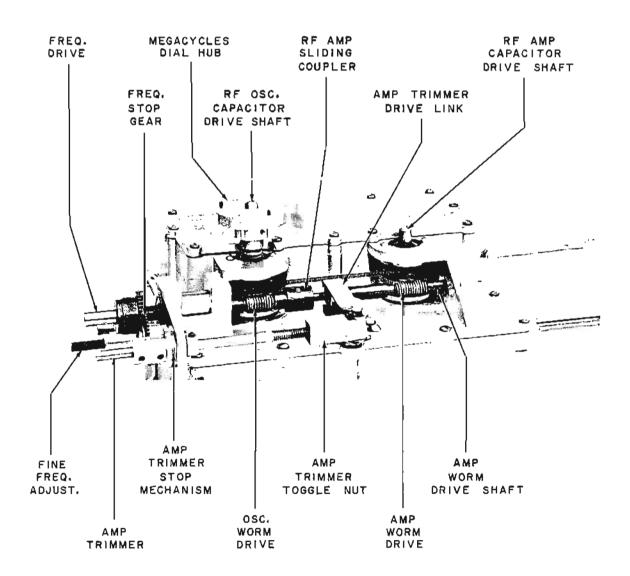
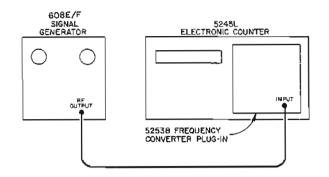
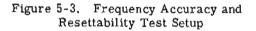


Figure 5-2. Lubrication Points, RF Tuner Drive Mechanism





5-13. CALIBRATOR ACCURACY TEST.

a. Connect signal generator intest setup illustrated in Figure 5-3.

b. Set FREQUENCY RANGE switch to range E.

c. Set XTAL CAL switch to 5 MC position.

d. Plug headset into XTAL CALOUTPUT connector.

e. Adjust Model 608E frequency control for MEGA-CYCLES dial reading of 480 MHz, or Model 608F for MEGACYCLES dial reading of 455 MHz.

f. Adjust RF OUTPUT and attenuator controls for proper triggering of electronic counter.

g. Readjust frequency control for zero beat in headset. Adjust XTAL CAL GAIN control as required.

h. Electronic counter should indicate within $\pm 0.01\%$ (± 48 kHz) of 480 MHz in Model 608E, or $\pm 0.01\%$ (± 45.5 kHz) of 455 MHz in Model 608F at zero beat in headset.

i. Set XTAL CAL switch to 1 MC position.

k. Adjust frequency control on Model 608E or 608F for MEGACYCLES dial reading of 270 MHz at zero beat in headset.

l. Repeat steps f. and g. Electronic counter should indicate within $\pm 0.01\%$ (± 27 kHz) of 270 MHz.

5-14. FREQUENCY DRIFT TEST.

a. Connect signal generator in test setup illustrated in Figure 5-3.

b. Set FREQUENCY RANGE switch to range A.

c. Adjust frequency control for a MEGACYCLES dial reading of 15 MHz.

d. Adjust RF OUTPUT and attenuator controls for proper triggering of electronic counter. Allow signal generator to stabilize, i.e., for one hour after turn on and ≈ 25 minutes after a frequency change.

e. Record electronic counter indication.

f. Wait 10 minutes and again record electronic counter indication. Frequency should be within $\pm 0.005\% (\pm 750 \text{ Hz})$ of the indication recorded in step e.

g. Repeat steps e. and f. for the following FRE-QUENCY RANGE and MEGACYCLES dial settings. Allow the signal generator to stabilize for 25 minutes after changing the frequency before making the frequency drift measurement.

FREQ RANGE	FREQ.	DRIFT
B	30 MHz	<u>+</u> 1.50 kHz
С	75 MHz	<u>+</u> 3.75 kHz
a	150 MHz	\pm 7.50 kHz
E	340 MHz	±17.00 kHz

5-15. RESIDUAL FM TEST.

a. Connect signal generator intest setup illustrated in Figure 5-4.

b. Set FREQUENCY RANGE switch to range A.

c. Adjust frequency control for MEGACYCLES dial reading of 10 MHz.

d. Adjust RF OUTPUT and AMPL TRIMMER controls for a +7 dBm (ATTENUATOR CALIBRATED) indication on RF OUTPUT meter.

e. Adjust attenuator control for output setting of 0 dBm.

f. Readjust frequency control for a 50 kHz indication on frequency meter on the 100 kHz range.

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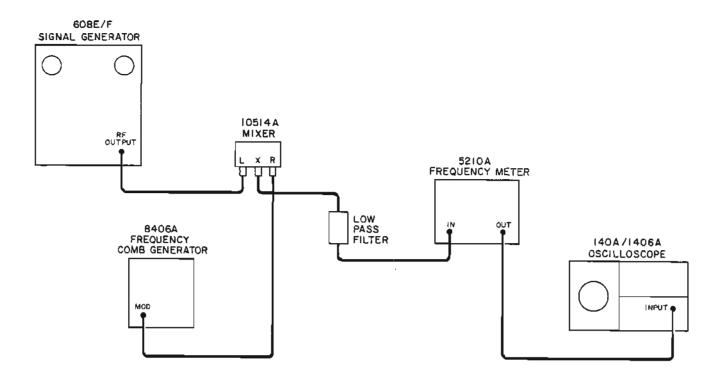


Figure 5-4. Residual and Incidental FM Test Setup

g. Measure peak-to-peak amplitude of signal displayed on oscilloscope. Measurement should not exceed 100 μ V peak-to-peak (10 Hz peak-to-peak maximum deviation).

h. Repeat steps d. through f. at the FREQUENCY RANGE and MEGACYCLES dial settings listed with applicable oscilloscope indications in the following chart.

FREQ. RANGE	FREQ.	OSCILLOSCOPE INDICATION	
		RESIDUAL FM	MAXIMUM DEV
В	30 MHz	300 µV pk-pk	30 Hz pk-pk
с	70 MH2	700 μ V pk-pk	70 Hz pk-pk
D	150 MHz	1.5 mV pk-pk	150 Hz pk-pk
E	300 MHz	3 mV pk-pk	300 Hz pk-pk
E	400 MHz	4 mV pk-pk	400 Hz pk-pk
Έ	420 MHz	4.2 mV pk-pk	420 Hz pk-pk
Ê	450 MHz (608E)	4.5 mV pk-pk	450 Hz pk-pk
Е	455 MHz (608F)	4.55 mV pk-pk	455 Hz pk-pk
Έ	480 MHz (608E)	4.8 mV pk-pk	480 Hz pk-pk

5-16. INCIDENTAL FM TEST.

a. Connect signal generator intest setup illustrated in Figure 5-4.

b. Set FREQUENCY RANGE switch to range A.

c. Adjust frequency control for MEGACYCLES dial setting of 10 MHz.

d. Adjust RF OUTPUT and AMPL TRIMMER controls for a +7 dBm (ATTENUATOR CALIBRATED) indication on RF OUTPUT meter.

e. Set MODULATION switch to INT 1000 $\sim\,$ position.

f. Adjust MODULATION control for a 50% indication on PERCENT MODULATION meter.

g. Adjust attenuator control for output of 0 dBm.

h. Readjust frequency control for a 50 kHz indication on frequency meter on 100 kHz range.

i. Measure peak-to-peak amplitude of signal displayed on oscilloscope. Measurement should not exceed 2.0 mV peak-to-peak (200 Hz peak-to-peak maximum deviation).

j. Repeat steps h. and i. at the FREQUENCY RANGE and MEGACYCLES dial settings listed with applicable oscilloscope indications in the following chart. 5-17. MAXIMUM RF OUTPUT TEST.

a. Connect signal generator intest setup illustrated in Figure 5-5.

b. Set FREQUENCY RANGE switch to range A.

c. Adjust RF OUTPUT control to the full clockwise position.

d. Adjust attenuator control to the full clockwise position.

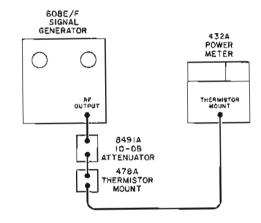


Figure 5-5. RF Output Test Setup

		OSCILLOSCOF	PE INDICATION
FREQ. RANGE	FREQ	INCIDENTAL FM	MAXIMUM DEV
в	30 MHz	6 mV pk-pk	600 Hz pk-pk
с	70 MHz	14 mV pk-pk	1400 Hz pk-pk
D	150 MHz	20 mV pk-pk	2000 Hz pk-pk
E	300 MHz	20 mV pk-pk	2000 Hz pk-pk
Е	400 MHz	20 mV pk-pk	2000 Hz pk-pk
Е	420 MHz	20 mV pk-pk	2000 Hz pk-pk
E	450 MHz (608E)	20 mV pk-pk	2000 Hz pk-pk
E	455 MHz (608F)	20 mV pk-pk	2000 Hz pk-pk
E	480 MHz (608E)	20 mV pk-pk	2000 Hz pk-pk

e. Rotate frequency control through entire limits of range A. Depress and tune AMPL TRIMMER control when necessary during tuning cycle to keep the rf output peaked.

f. Observe power meter for minimum output during tuning cycle. Indication should be a minimum of +3 dBm over the range for the 608E, and -3 dBm across the range for the 608F.

g. Repeat steps e. and f. for frequency ranges B through E.

5-18. FREQUENCY RESPONSE AND LEVELING TEST.

a. Connect signal generator intest setup illustrated in Figure 5-5.

b. Set FREQUENCY RANGE switch to range D.

c. Adjust frequency control for a MEGACYCLES dial reading near center of frequency range D.

d. Adjust attenuator control for a dial reading of +7 dBm.

e. Adjust RF OUTPUT control approximately onequarter turn from counter-clockwise stop.

f. Adjust AMPL TRIMMER control-knob arrow to eight-o'clock position and tune (without depressing) slowly clockwise. RF OUTPUT meter needle should follow right and kick back suddenly. Stop tuning AMPL TRIMMER control when needle kicks.

g. Adjust RF OUTPUT control for an RF OUTPUT meter indication of +7 dBm (ATTENUATOR CALI-BRATED).

h. Vary Frequency Control a few megacycles either side of MEGACYCLES dial setting made in step c. If RF OUTPUT meter needle does not lock on +7 dBm, slowly readjust RF OUTPUT and AMPL TRIMMER controls until needle remains at +7 dBm position.

i. Vary frequency control across frequency range D. Peak-to-peak power meter variation should be less than ± 1 dBm over range and RF OUTPUT meter needle should remain at +7 dBm position.

j. Repeat steps c. through i. for frequency ranges A through C and range E. The AMPLIFIER TRIMMER adjustment for leveling on bands A through D typically is the initial setting of AMPL TRIMMER plus one readjustment. Band E is typically the initial setting plus up to three readjustments.

5-19. RF OUTPUT ACCURACY TEST.

a. Connect signal generator in test setup illustrated in Figure 5-5 omitting 8491A 10-DB Attenuator. b. Set FREQUENCY RANGE switch to range C.

c. Adjust frequency control for MEGACYCLES dial reading of 70 MHz.

d. Depress and adjust AMPL TRIMMER control for maximum indication RF OUTPUT meter.

e. Adjust RF OUTPUT control for +7 dBm (ATTEN-UATOR CALIBRATED) indication on RF OUTPUT meter.

f. Set attenuator control to 0 dBm position. Power meter should indicate 0 dBm ± 1 dB.

5-20. UNCALIBRATED RF OUTPUT TEST.

a. Connect signal generator in test setup illustrated in Figure 5-6.

b. Model 608E should provide indication of -1.8 dBm minimum on power meter. Model 608F should provide power meter indications as follows:

10 to 215 MNz: -1.8 to +7.0 dBm 215 to 400 MHz: +2.0 to +6.0 dBm 400 to 450 MHz: +1.0 to +5.0 dBm

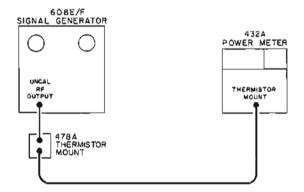


Figure 5-6. Uncalibrated RF Output Test Setup

5-21. EXTERNAL SINE-WAVE MODULATION AND MODULATION METER TEST.

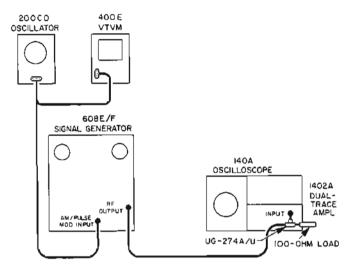
a. Connect signal generator intest setup illustrated in Figure 5-7.

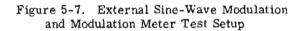
b. Set FREQUENCY RANGE switch to range A.

c. Adjust frequency control for MEGACYCLES dial reading of 15 MHz.

d. Adjust oscilloscope gain for envelope amplitude of 5 cm peak-to-peak.

e. Set MODULATION switch to EXT AM position.





f. Adjust oscillator for a 2-kHz, 1-volt rms output as observed on VTVM.

g. Adjust MODULATION control maximum clockwise. PERCENT MODULATION meter should indicate 95% modulation.

h. Adjust MODULATION control for 90% modulation indication on PERCENT MODULATION meter.

i. Measure peak-to-peak amplitude (in centimeters) of modulation envelope from uppermost peak to uppermost trough. Amplitude should be 4.5 ± 0.5 cm (90 $\pm10\%$ modulation).

j. Repeat steps h. and i. to test the following modulation percentages:

Percent Modulation Meter (%)	Amplitude (cm)	Equivalent Modulation Percentage (%)
80 70 60 50 40 30 20 10	$\begin{array}{c} 4. \ 0 \ \pm \ 0. \ 25 \\ 3. \ 5 \ \pm \ 0. \ 25 \\ 3. \ 0 \ \pm \ 0. \ 25 \\ 2. \ 5 \ \pm \ 0. \ 25 \\ 2. \ 0 \ \pm \ 0. \ 25 \\ 1. \ 5 \ \pm \ 0. \ 25 \\ 1. \ 0 \ \pm \ 0. \ 25 \\ 1. \ 0 \ \pm \ 0. \ 25 \\ 0. \ 5 \ \pm \ 0. \ 25 \end{array}$	$\begin{array}{r} 80 \pm 5 \\ 70 \pm 5 \\ 60 \pm 5 \\ 50 \pm 5 \\ 40 \pm 5 \\ 30 \pm 5 \\ 20 \pm 5 \\ 10 \pm 5 \\ 10 \pm 5 \end{array}$

5-22. INTERNAL MODULATION TEST.

a. Set FREQUENCY RANGE switch and frequency control to any desired position.

b. Adjust the RFOUTPUT control for ATTENUATOR CALIBRATED condition.

c. Adjust Attenuator control for 0 dBm.

d. Set MODULATION switch to INT 1000 $\,^\circ$ position.

e. Adjust MODULATION control clockwise and PERCENT MODULATION meter should indicate at least 95% modulation.

f. Adjust MODULATION control to maximum counter-clockwise position. PERCENT MODULATION meter should indicate 0%.

g. Set MODULATION switch to INT 400 v position.

h. Repeat steps e and f.

5-23. ENVELOPE DISTORTION TEST.

a. Connect signal generator in test setup illustrated in Figure 5-8.

b. Set FREQUENCY RANGE switch to range C and adjust frequency control for a MEGACYCLES dial reading of 70 MHz.

c. Set MODULATION switch to CW position.

d. Adjust RF OUTPUT control for indication on RF OUTPUT meter.

e. Depress and tune AMPL TRIMMER control for peak indication on RF OUTPUT meter.

f. Readjust RF OUTPUT control for +7 dBm (AT-TENUATOR CALIBRATED) indication on RF OUTPUT meter.

g. Adjust oscillator for a 10 kHz, 1 V rms, output.

h. Set MODULATION switch to AM position.

i. Adjust MODULATION control for 30% modulation indication on PERCENT MODULATION meter.

j. Measure distortion using 334A Distortion Analyzer. Distortion should be <2%.

k. Adjust MODULATION control for 70% modulation and measure distortion at desired frequencies. Distortion should be <5%.

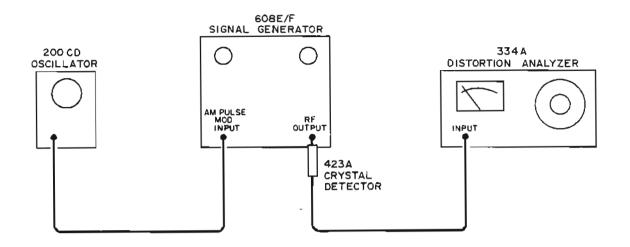


Figure 5-8. Envelope Distortion Test Setup

5-24. EXTERNAL PULSE MODULATION AND RESIDUAL LEVEL TEST.

a. Connect signal generator intest setup illustrated in Figure 5-9.

b. Set FREQUENCY RANGE switch to range C.

c. Adjust frequency control for a MEGACYCLES dial reading of 44 MHz.

d. Set MODULATION switch to PULSE SET RF LEVEL position.

e. Adjust RF OUTPUT control for indication on RF OUTPUT meter.

f. Depress and adjust AMPL TRIMMER control for peak indication on RF OUTPUT meter.

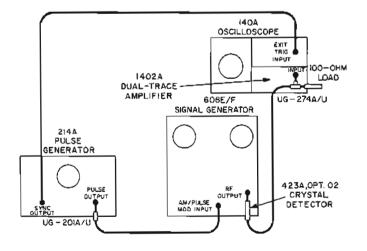


Figure 5-9. Pulse Modulation Test Setup

g. Readjust RF OUTPUT control for a +7 dBm (ATTENUATOR CALIBRATED) indication on RF OUT-PUT meter.

h. Adjust attenuator control for output indication of +7 dBm.

i. Set MODULATION switch to PULSE position.

j. Adjust pulse generator for 10-volts positive output, 10 μ s wide.

k. Oscilloscope should indicate a combined pulse rise and decay time of $4 \mu \text{sec}$ maximum and a pulse width of 10 $\pm 1.5 \mu \text{sec}$.

1. Set FREQUENCY RANGE switch to range E.

m. Adjust frequency control for MEGACYCLES dial setting of 220 MHz.

n. Repeat steps d. through k.

o. Connect signal generator intest setup illustrated in Figure 5-5 omitting 10-dB attenuator.

p. Repeat steps d. through h.

q. Set MODULATION switch to PULSE position. Power meter should indicate -13 dB minimum.

5-25, RF LEAKAGE TEST.

a. Connect signal generator intest setup illustrated in Figure 5-10.

b. Set FREQUENCY RANGE switch to range E.

c. Adjust frequency control for MEGACYCLES dial reading of 400 MHz.

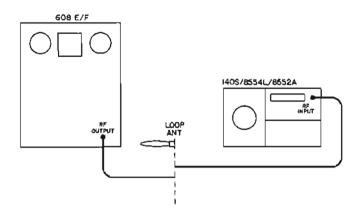


Figure 5-10. RF Leakage Test Setup

d. Set MODULATION switch to CW position.

e. Adjust RF output control for ATTENUATOR CALIBRATED condition.

f. Adjust Attenuator control for $1 \mu V$ output.

g. Adjust spectrum analyzer for reference of 1 $\mu V.$

h. Disconnect the cable from signal generator RF OUTPUT connector. Set ATTEN to 0.1 μ V.

i. Connect loop antenna (see Figure 5-1) to cable.

j. Probe with loop antenna over entire signal generator cabinet area. The spectrum analyzer indication should not exceed 1 μ V.

5-26. FREQUENCY CONTROL INPUT TEST (MODEL 608F ONLY).

a. Connect signal generator in test setup illustrated in Figure 5-11.

b. Set FREQUENCY RANGE switch to range A.

c. Adjust frequency control for MEGACYCLES dial reading of 10 MHz.

d. Connect power supply and adjust for -2 Vdc at FREQ CONTROL INPUT. NOTE frequency counter reading.

e. Adjust power supply for -32 Vdc at FREQ CONTROL INPUT. The frequency counter should indicate a frequency change of more than 0.2%, or 20 kHz.

f. Set frequency to high end of range A, 21 MHz.

g. Return power supply voltage to -2 Vdc at the FREQ CONTROL INPUT. The frequency counter should indicate a frequency change of more than 1%, or 210 kHz.

h. Repeat this test on other bands of interest.

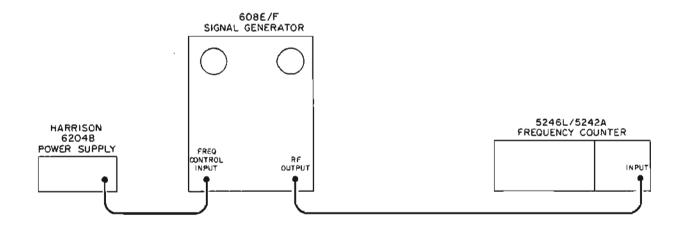


Figure 5-11. Frequency Control Input Test Setup

a. Connect ohmmeter to FREQ ANALOG OUTPUT jack on signal generator.

b. Rotate frequency control for a MEGACYCLES dial reading at the extreme low end. Ohmmeter should indicate between 2500 and 3500 ohms.

c. Rotate frequency control for a MEGACYCLES dial reading at the extreme high end. Obmmeter should indicate zero ohms.

5-28. SYNCHRONIZATION TEST WITH MODEL 8708A SYNCHRONIZER (MODEL 608F).

a. Connect signal generator intest setup illustrated in Figure 5-12.

b. Set 608F and 8708A controls as follows:

$608\,\mathrm{F}$

FREQUENCY RANGE				. A
FREQUENCY	high	end	of	band
MODULATION switch				CW

8708A

FREQ RANGE									Α
FREQ TUNING						c	ent	er	ed
MODULATION	•	•	•		•		•	C	W

d. Slowly tune 608F across band A. The 8708A meter pointer will move to left, indicating that the FREQ CONTROL OUTPUT voltage is changing to maintain phase lock. When end-of-lock range is reached, the 8708A meter pointer will jump to right indicating phase-lock is re-established on the next lock point. On the oscilloscope, the dc level will shift to approximately +3.0 V when the frequency is increased, and approximately -1.0 V when the frequency is decreased. The oscilloscope trace should not jump outside the the +3V and -1V limits (this indicates weak locking due to either 608F or the 8708A).

e. Repeat steps c. and d. for frequency ranges B through E.

5-29. ADJUSTMENT.

5-30. The following procedures list the internal adjustments necessary for proper operation of the signal generator. The test equipment required to perform the adjustments are contained in Table 5-1.

WARNING

Performance of the following routines require that the dust cover be removed from the signal generator. Lethal voltages exist within the unit. Exercise extreme caution when performing the procedures.

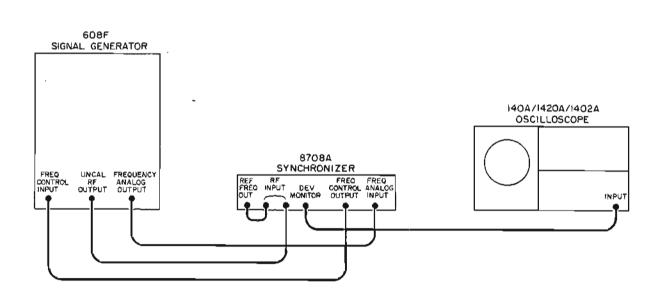


Figure 5-12. Synchronization Test Setup

Model 608E/F

5-31. POWER SUPPLY VOLTAGE ADJUSTMENTS,

5-32. There are four adjustable dc operating voltages: -165, +225, +25, and -6.3 volts. Adjust these voltages only if proven by accurate measurement to be outside the tolerances specified below and only if voltage error is not caused by excessive current drawn through the regulator or by inadequate voltage supplied to the regulator from the power supply rectifiers.

5-33. Because of the interdependence of these power supplies, the -165 volt regulator must be set first, the +225 volt regulator second, and the +25 and -6.3 volt regulators last. After adjusting any of these regulators, check the output voltage of each other regulator to be sure it is still within specified tolerances.

a. Connect an adjustable transformer to control line voltage applied to the signal generator.

b. Set signal generator controls as follows:

MODULATION control				
RFOUTPUT		•		, max. ccw
FINE FREQ				centered
Frequency				
AMPL TRIMMER				max. cw
MODULATION selecto	r			CW
Attenuator				max. cw
Power switch				ON
XTAL CAL				OFF

c. Adjust line voltage transformer to deliver 115 (or 230) volts. Be sure 115/230 volt switch is set correctly.

d. Measure, and if necessary, adjust the following voltages:

Voltage	Adjustment (Figure 5-13)
$ \begin{array}{r} -165 \pm 5 \\ +225 \pm 5 \\ +25 \pm 1 \\ -7.2 \pm 6.3 \end{array} $	R80 R71 Replace CR28 if necessary Replace CR20 if necessary

5-34. MODULATION METER ADJUSTMENT.

5-35. MECHANICAL ZERO. Adjust mechanical zero as follows:

a. Set MODULATION selector to CW.

b. Connect shorting lead across PERCENT MODU-LATION meter terminals.

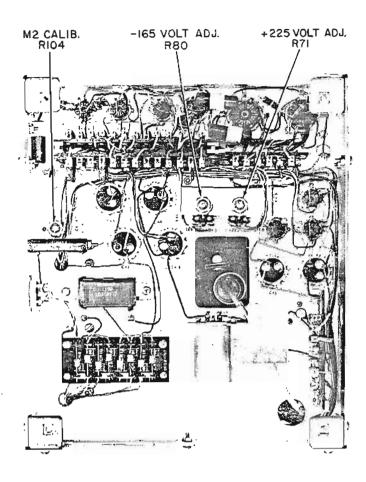


Figure 5-13. Location of Power Supply and Modulation Meter Adjustments

c. Adjust mechanical zero-adjust screw clockwise until pointer is at left of meter zero and moving towards meter zero; stop adjustment when meter needle is exactly at zero. If needle overshoots, repeat adjustment.

d. Carefully adjust mechanical zero-adjust screw a few degrees to free screw from meter suspension. If pointer moves off zero, repeat step c.

e. Remove shorting lead from meter terminals.

5-36. ELECTRICAL ZERO. Adjust electrical zero as follows:

a. Adjust R106 (front-panel ZERO adjustment) fully clockwise. PERCENT MODULATION meter should indicate at least 30% modulation.

b. Adjust line voltage from 102 to 128 (or 204 to 256) volts. PERCENT MODULATION meter indication should not change more than $\pm 10\%$ of full scale.

c. Adjust line voltage to 115 (or 230) volts. Adjust R106 for zero indication on PERCENT MODULATION meter.

5-37. CALIBRATION OF PERCENT MODULATION METER.

5-38. Recalibration of the PERCENT MODULATION meter may be necessary following a repair of the modulation measuring circuits (V18-V21) or after replacement of the meter itself.

NOTE

Be sure the mechanical and electrical zero is adjusted properly; refer to paragraph 5-34.

a. Connect 608E/F and oscilloscope as shown in Figure 5-14.

b. Set the MODUALTION selector to CW.

٩	FREQUENCY RANGE A band
	MEGACYCLES dial 10 MHz
	MOD. SELECTOR 1000 ν
	AMP. TRIMMER Adjust for maximum output
	RF OUTPUT ATTENUATOR CALIBRATED
	MODULATION

c. Set the 140/1420/1402 oscilloscope as follows:

VERT. SENS 0.2 V/cm
HOR. SWEEP TIME 0.2 ms/cm
TRIGGER SOURCE + INT or - INT

d. Adjust vertical position to place pattern at center of screen.

e. Set MODULATION selector to 1000-. Adjust 608E/F Attenuator for a 4 cm pattern on screen.

5-39. ADJUSTMENTS.

a. Increase MODULATION control until sine wave pattern appears on oscilloscope. Adjust trigger level for stable pattern.

b. Increase MODULATION control until peak-topeak pattern is 6 cm high, as shown in Figure 5-15. Adjust R104 (METER CAL on rear panel) for a reading of 50% on the PERCENT MODULATION meter.

c. Increase MODULATION control until the peakto-peak pattern is 8 cm high. Adjust trigger level as necessary to keep stable pattern. The PERCENT MODULATION meter should read approximately 100%.

5-40. This check will determine if the modulation meter circuitry (V18-V21) is frequency sensitive. It will also check the relative output of the Modulation Oscillator (V2).

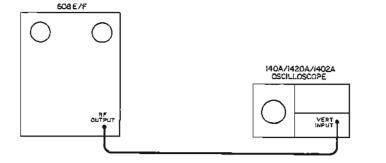


Figure 5-14. Modulation Meter Calibration Test Setup

a. Return the MODULATION control to a meter reading of 50%.

b. Turn the MOD. SELECTOR switch to 400 %. Meter should read $\pm 5\%$ of that obtained in step a.

5-41. AUTOMATIC OUTPUT LEVELING ADJUSTMENT.

a. Connect dc voltmeter (50-volt range) to junction of CR26-R235 (Figure 5-16) and ground.

b. Set RF OUTPUT control fully counterclockwise and MODULATION selector to CW.

c. Adjust R204 (Figure 5-14) to point where dc voltmeter shows no further voltage increase with adjustment of R204.

d. DC voltmeter should indicate greater than 30 volts.

e. Set MODULATION selector to $400 \ varsistication$, and AM position; dc voltmeter should indicate greater than 30 volts at each switch position.

f. Set MODULATION selector to PULSE/SET RF OUTPUT; voltmeter should indicate greater than 14 volts.

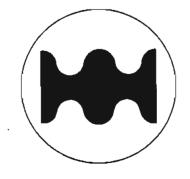
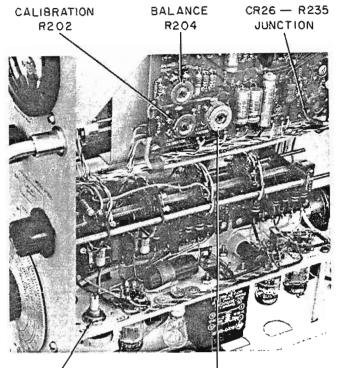
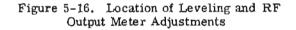


Figure 5-15. Pattern on 140 Oscilloscope at 50% Modulation



MODULATION OSCILLATOR R7

TRACKING R209



g. Set MODULATION selector to PULSE; voltmeter should indicate greater than 25 volts.

h. Set RF OUTPUT control fully clockwise.

i. Set MODULATION selector to CW and PULSE/ SET RF OUTPUT positions; dc voltmeter should indicate +0.3 to +1.0 volt dc at each switch position.

j. Set MODULATION selector to PULSE; voltmeter should indicate greater than 20 volts.

5-42. RF OSCILLATOR ADJUSTMENT.

a. Connect signal generator in test setup shown in Figure 5-11 omitting Harrison Power Supply.

b. Set FREQUENCY RANGE switch to band A.

c. Center FINE FREQ and cursor controls.

d. Adjust frequency control for 21 MHz.

e. Check counter frequency indication; counter should indicate 21 MHz $\pm 0.5\%$ (± 105 kHz). If not, adjust capacitor C59 (Figure 5-17) for proper counter indication.

f. Repeat steps b. through e. for each of the following frequencies, adjusting C59 for the best compromise at the listed frequencies:

Band		Frequency
в		43 MHz
С		95 MHz
D	-	210 MHz
E		455 MHz

5-43. UNCALIBRATED RF OUTPUT, MINIMUM RF OUTPUT, AND RF AMPLIFIER TRACKING ADJUSTMENT.

a. Connect signal generator in test setup shown in Figure 5-5, but delete 10-dB attenuator.

b. Set FREQUENCY RANGE switch to band A.

c. Rotate frequency control through entire range plus one-half turn at edges of band, while adjusting AMPL TRIMMER for maximum RF output. Observe Model 432A power meter for following indications:

UNCAL RF OUTPUT (608E): -1.8dBm minimum

RF OUTPUT (608E): 13 dBm minimum

UNCAL RF OUTPUT (608F): -1.8dBm minimum 10-215 MHz

+2.0dBm minimum 215-400 MHz

+1.0dBm minimum 400-430 MHz

RF OUTPUT (608F): +10 dBm minimum.

NOTE

Although Model 608F is capable of well in excess of its published specification, prolonged use at greater power levels than +7 dBm should be avoided. However, it is usable for a short time to make these adjustments.

d. AMPL TRIMMER control should peak with at least one-quarter turn adjustment at either end.

e. If necessary, the power amplifier tuning coils can be adjusted to obtain the proper output at the RF OUTPUT connector as follows:

Band A: Reposition 3 or 4 turns on coil.

Bands B through D: Reposition shorted turns on coils.

Band E: Spread or compress coil loop.

f. Repeat steps b through d for bands B through E.

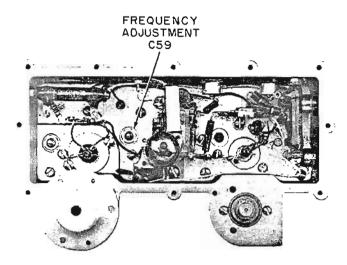


Figure 5-17. Location of RF Oscillator Adjustment

- 5-44. RF OUTPUT METER ACCURACY ADJUSTMENT.
- 5-45. MECHANICAL ZERO AND ELECTRICAL ADJUSTMENT.
 - a. Adjust mechanical zero as follows:
 - b. Set MODULATION selector to CW.

c. Connect shorting leads across PERCENT MODULATION meter.

d. Adjust mechanical zero-adjust screw clockwise until pointer is at left of meter zero and moving towards meter zero. Stop adjustment when meter needle is exactly on zero. If needle overshoots, repeat adjustment.

e. To electrically calibrate meter, proceed as follows:

f. Connect signal generator in test setup illustrated in Figure 5-5, omitting 10-dB attenuator.

g. Set MODULATION switch to CW position.

h. Set MODULATION control completely counterclockwise.

i. Set FREQUENCY RANGE switch to range C, and adjust frequency control to center mark on MEGACYCLES dial (approximately 75-80 MHz).

j. Adjust RF OUTPUT control for indication on RF OUTPUT meter.

k. Depress and adjust AMPL TRIMMER control for peak indication on RF OUTPUT meter.

m. Readjust RF OUTPUT control for +7 dBm (ATTENUATOR CALIBRATED) indication on RF OUTPUT meter,

n. Adjust Attenuator control for output of -1.0 dBm.

o. Adjust RF OUTPUT control for power meter readings shown below. Calibrate the output meter by adjusting R202 and R209 as indicated.

NOTE

In the Model 608F, the RF OUTPUT meter is calibrated using full-scale deflection when the power meter is indicating +5 dBm. During normal operation only half-scale (+7 dBm) deflection is used, indicating the maximum RF output capabilities of the instrument.

POWER OUTPUT (432A Rdg) dBm	OUTPUT METER Rdg- dB	ADJUST see Figure 5-16
+5.0	+13.0	R202 CH
-4.0	+4.0	R209
+5.0	+13.0	R202
-4.0	+4.0	R209
+2.0	9,8-10.2	
-1.0	6.9-7.2	

5-46. RF OUTPUT METER/ATTENUATOR FRE-QUENCY RESPONSE AND AUTOMATIC LEVEL OUTPUT ADJUSTMENT.

a. The following steps will indicate if an adjustment is necessary.

b. Connect signal generator in test setup as shown in Figure 5-5, but delete the 10-dB attenuator.

c. Adjust AMPL TRIMMER for best tracking compromise over bands A through D.

d. Adjust RF OUTPUT control for ATTENUATOR CALIBRATED (+7 dBm) indication on meter.

e. Vary frequency control over entire frequency range of bands A through D. The RF OUTPUT meter indication should remain constant within ± 1 dB over entire range of bands A through D. There should be no further adjustment of the AMPL TRIMMER necessary, however one readjustment is acceptable (initial setting plus one readjustment) per band.

f. Set FREQUENCY RANGE switch to band E.

g. Adjust frequency control to lower end of band, and adjust AMPL TRIMMER for maximum RF output.

h. Adjust frequency control over lower half of band E. Power meter indication should remain constant within ± 1 dB over lower half of band E.

i. Adjust frequency control to middle of band E, and adjust AMPL TRIMMER for maximum RF output.

j. Adjust frequecy control over upper half of band E and adjust AMPL TRIMMER for maximum output. Power meter indication should remain constant within ± 1 dB over upper half of band E.

k. Set attenuator and FREQUENCY RANGE as follows and observe power meter for proper indication:

NOTE

When checking attenuator on band E, readjust AMPL TRIMMER for maximum RF output and then adjust RF OUTPUT control for +7 dBm on RF OUTPUT meter.

Attenuator Setting dBm	Frequency Range Band	Power Output dBm
-1.0	C	0 to -2.0
-1.0	A	0 to -2.0
-1.0	В	0 to -2.0
-1.0	D	0 to -2.0
-1.0	E	0 to -2.0
+7.0	\mathbf{E}	+6.0 to +8.0
+7.0	D	+6.0 to +8.0
+7.0	С	+6.0 to +8.0
+7.0	B	+6.0 to +8.0
+7.0	A	+6.0 to +8.0

m. If the power output is not within the above specified limits, proceed as follows:

n. Setup RF OUTPUT meter adjustments as shown in paragraph 5-45.

CAUTION

All readings are referenced to Band C. There are several variables which have to be taken into account. The amplifier coils for bands A through D can be moved closer or further away from the attenuator probe pickup loop. In general, the power level becomes more positive by moving the coils closer to the attenuator, and more negative by moving the coils further away. THERE IS HIGH VOLTAGE ON ALL AMPLIFIER COILS, AND ON THE STATOR OF TUNING CAPACITOR C17. MAKE CERTAIN THAT THERE IS SUFFI-CIENT CLEARANCE BETWEEN THE COILS AND THE ATTENUATOR PROBE PICKUP LOOP. Band E presents a different problem. The amplifier coil is not adjustable, so the procedure calls for changing the bolometer response to conform to other bands.

o. Check band C and adjust if necessary to bring into specifications.

p. Check E band. If power output is not within specifications, adjust either coil L17 or resistor R60 (see Figures 5-18 and 5-19). After adjusting either L17 or R60, always repeat RF OUTPUT meter adjustments, as in paragraph 5-44.

CAUTION

Some 608E/F instruments may have a large blob of solder on the pigtail lead of R60. This was done at the factory to lessen the effects of lead inductance. This effectively shortens the pigtail lead of R60 without cutting the lead, which may at a later date prove unsatisfactory and necessitate replacement.

q. Check and adjust, if necessary, bands D, B, and A.

5-47. CRYSTAL CALIBRATOR ADJUSTMENT.

a. Remove signal generator chassis from cabinet, and remove side plate from RF generator housing.

b. Plug headset into XTAL CAL OUTPUT jack.

c. Set signal generator controls as follows:

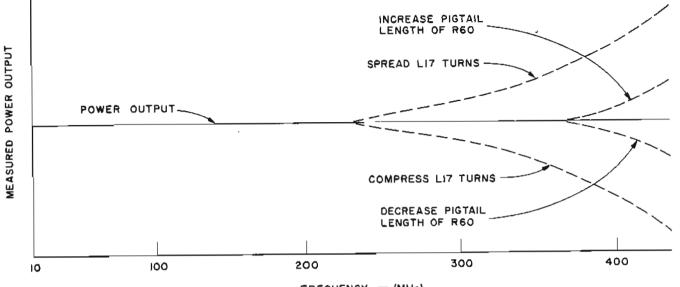
FREQUENCY RANGE	
XTAL CAL GAIN	max. cw
XTAL CAL	1 MC

d. Adjust L18 (Figure 5-19) in clockwise direction. Two separate beat frequencies should be heard in headset within one full turn of L18. Adjust L18 to audible null between the two beat notes.

e. Set signal generator controls as follows:

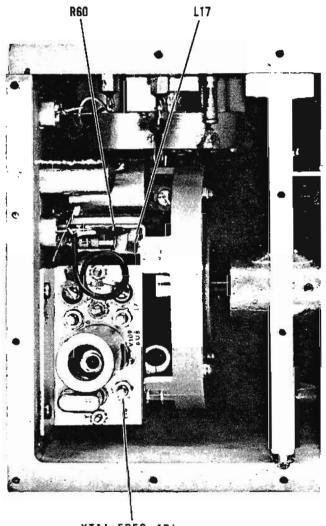
FREQUENCY	RANGE										. ban	dΕ
RF OUTPUT	control	•		•	•	•	•	•	•	•	max.	cw

f. Set XTAL CAL to 5 MC.



FREQUENCY - (MHz)

Figure 5-18. Frequency-Response Adjustments



XTAL FREQ AD} L18

Figure 5-19. Location of Frequency-Response and Crystal-Calibrator Adjustments

g. Adjust frequency control from 450 to 455 MHz. Two beat notes should occur, one at 450 MHz and one at 455 MHz.

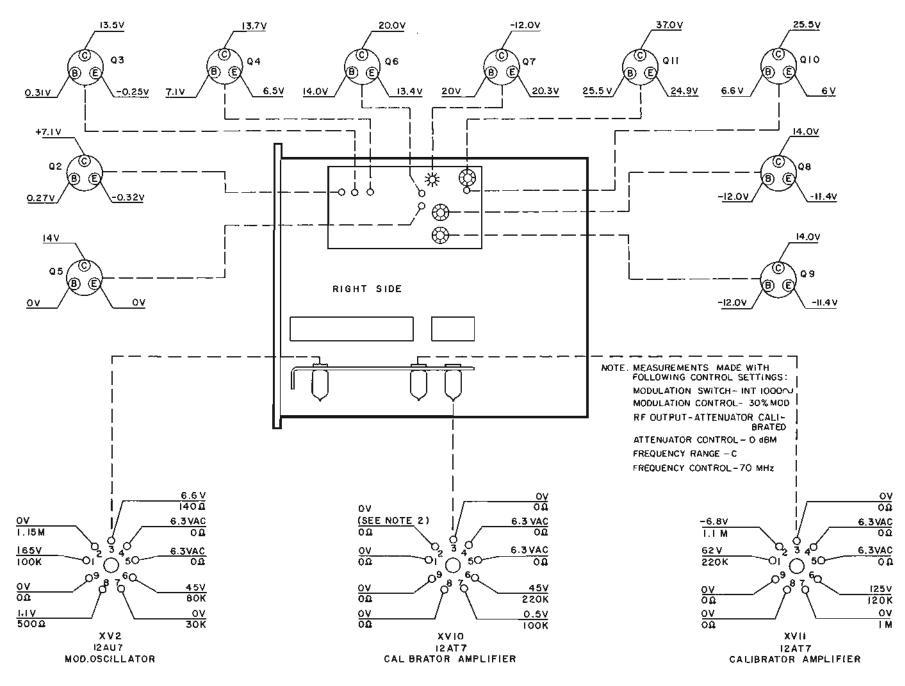
5-48. TROUBLESHOOTING.

5-49. The following paragraphs contain information for troubleshooting the signal generators, and procedures for isolating troubles in transistor circuits. The procedures are not intended to correct for small out-of-tolerance conditions, but are directed toward correcting a major fault.

5-50. ISOLATING A TROUBLE TO A CIRCUIT SECTION.

5-51. To isolate a trouble to a particular circuit, a troubleshooting tree (table 5-7) is presented. The tree presents five major problems that can occur during operation of the signal generators, and includes block-to-block instructions for maintenance personnel to trace a malfunction to a particular circuit. Normal shop troubleshooting practices must be employed after the trouble has been isolated to the circuit section.

5-52. As a further aid in circuit troubleshooting the instrument, voltage and resistance measurements are presented in Figures 5-20 and 5-21. Figure 5-20 lists the voltage and resistance measurements during INT 1000 \sim operation, and Figure 5-21 depicts the voltage measurements on the ALC board during CW and PULSE/SET RF OUTPUT operation.



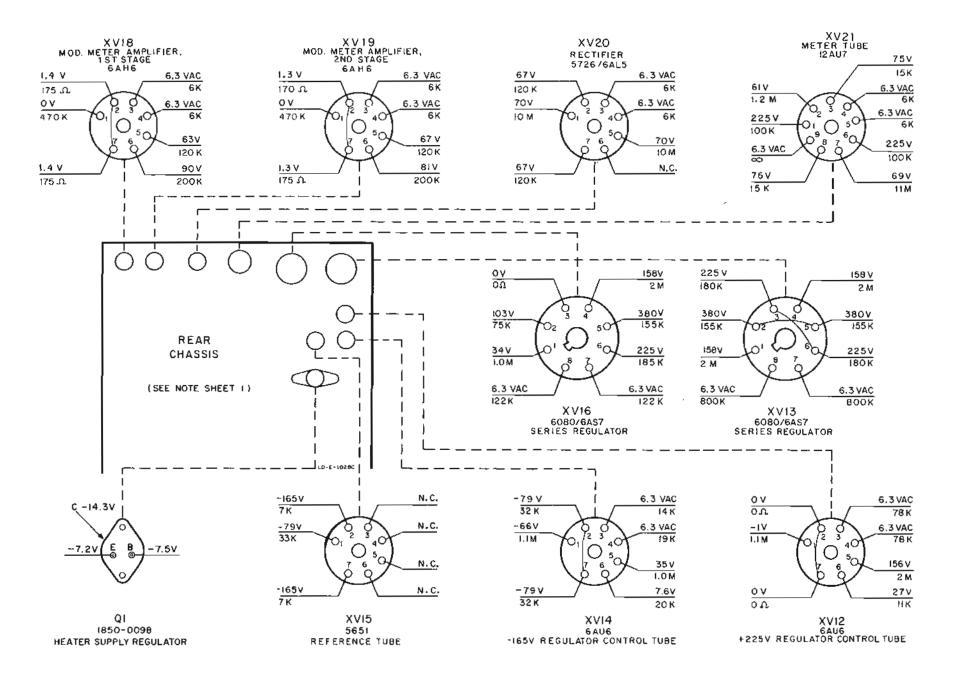
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Figure 5-20. Voltage and Resistance Measurements, INT 1000 ~ Operation (Sheet 1 of 3)



5-20

Figure 5-20. Voltage and Resistance Measurements, INT 1000 ~ Operation (Sheet 2 of 3)

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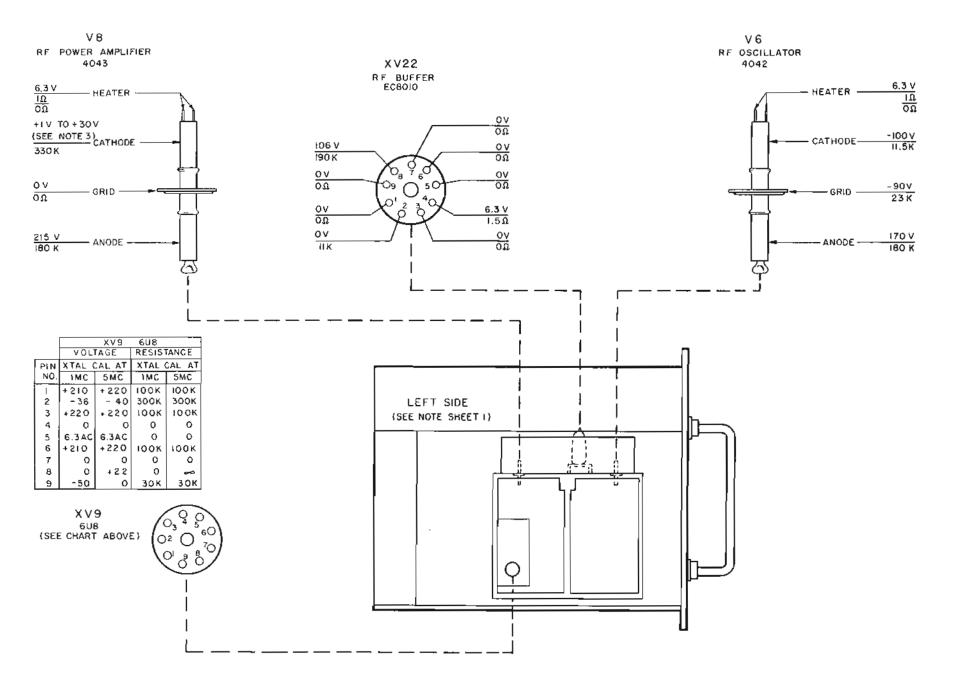
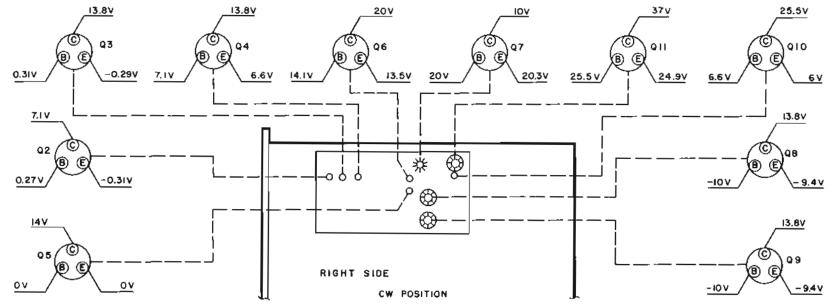
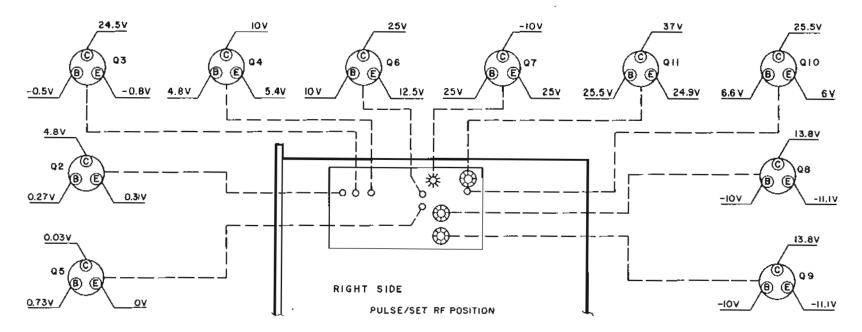


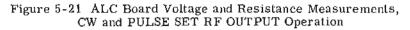
Figure 5-20. Voltage and Resistance Measurements, INT 1000 - Operation (Sheet 3 of 3)

Model 608E/F



NOTE: DIAL CONTROL FREQUENCY, RF OUTPUT, AND ATTENUATOR CONTROL SETTINGS SAME AS FIGURE 5-20





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5-53. Numerous test points have been established in the signal generators and are identified by a circlestar designator on Figures 7-4 and 7-5. The dc voltages listed with the test-point designators were obtained with the instrument in an INT 1000 \sim mode of operation.

5-54. ISOLATING TROUBLE IN TRANSISTOR CIRCUITS.

5-55. The following procedures and data are given to aid in determining whether a transistor is operational. Tests are given for both in-circuit and out-of-circuit transistors.

5-56. IN-CIRCUIT TESTING. The common causes of transistor failures are internal short- and opencircuits. In transistor circuit testing the most important consideration is the transistor base-emitter junction. Like the control grid of a vacuum tube, this is the operational control point in the transistor. This junction is essentially a solid-state diode. For the transistor to conduct, the diode must conduct; that is, the diode must be forward biased. As with simple diodes, the forward-bias polarity is determined by the materials forming the function. Use the transistor symbol on the schematic diagram to determine the bias polarity required to forward-bias the base-emitter junction. The A part of Figure 5-22 shows transistor symbols with terminals labeled. Notice that the emitter arrow conventionally points toward the type N material. The other two columns of the illustration compare the biasing required to cause conduction and cutoff in transistors and vacuum tubes. If the transistor base-emitter diode (junction) is forward-biased the transistor conducts. If the diode is heavily forwardbiased, the transistor saturates. However, if the base-emitter diode is reverse-biased the transistor is cut-off. The voltage drop across a forward biased emitter-base diode varies with transistor collector current. For example, a germanium transistor has a typical forward-bias, base-emitter voltage of 0.2 -0.3 volt when collector current is 1 - 10 mA, and 0.4-0.5 volt when collector current is 10 - 100 mA. In contrast, forward bias voltage for silicon transistors is about twice that for germanium types: about 0.5 -0.6 volt when collector current is low, and about 0.8 -0.9 volt when collector current is high.

5-57. Figure 5-22, part B, shows simplified versions of the three basic transistor circuits and gives the operating characteristics of each. When examining a transistor stage, first determine if the emitter-base diode is biased for conduction (forward-biased) by measuring the voltage difference between emitter and base. When using an electronic voltmeter, do not measure directly between emitter and base: there may be sufficient loop current between the voltmeter leads to damage the transistor. Instead, measure to a common point (e.g., chassis). If the emitter-base diode is forward-biased, check for amplifier action by short-circuiting base to emitter while observing collector voltage. The short-circuit eliminates baseemitter bias and should cause the transistor to stop conducting (cut off). Collector voltage should then shift to near the supply voltage. Any difference is due to leakage current through the transistor and, in general, the smaller this current, the better the transistor. If collector voltage does not change, the transistor may have either an emitter-collector short circuit of emitter-base open circuit.

5-58. TESTING TRANSISTORS WITH AN OHMME-TER. The two common causes of transistor failure are internal short- and open-circuits. Remove the transistor from the circuit and use an ohmmeter to measure internal resistance. See Table 5-3 for measurement data.

CAUTION

Most ohmmeters can supply enough current or voltage to damage a transistor. Before using an ohmmeter to measure transistor forward or reverse resistance, check open-circuit voltage and short-circuit current output ON THE RANGE TO BE USED. Open-circuit voltage must not exceed 1.5 volts and short-circuit current must be less than 3 ma. See Table 5-4 for safe resistance ranges for some common chmmeters.

5-59. REPLACEMENT AND REPAIR.

5-60. The following procedures contain instructions for work on etched circuit boards, replacement of electron tubes. attenuator probe replacement and repair, replacement of Lamp DS1, and diagrams for locating parts in the signal generator.

5-61. ETCHED CIRCUIT BOARD REPAIR.

5-62. The etched circuit boards in the Signal Generator are of the plated-through type consisting of metallic conductors bonded to both sides of insulating material. The metallic conductors are extended through the component mounting holes by a plating process. Soldering can be done from either side of the board with equally good results. Table 5-5 lists recommended tools and materials. Following are recommendations and precautions pertinent to etched circuit repair work.

a. Avoid unnecessary component substitution; it can result in damage to the circuit board and/or adjacent components.

b. Do not use a high-power soldering iron on etched circuit boards. Excessive heat may lift a conductor or damage the board.

A. TRANSISTOR BIASING						
DEVICE	SYMBOL	CUT OFF	CONDUCTING			
VACUUM TUBE		-15V	+200V			
N P N TRANSISTOR			+.3V HAIN CURRENT			
PNP TRANSISTOR	COLLECTOR BASE		3V CURRENT			

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B. AMPLIFIER CHARACTERISTICS							
CHARACTERISTIC	COMMON BASE	COMMON Emitter	COMMON Collector				
INPUT Z	30-50 Ω	500-1500 Ω	20-500Κ Ω				
OUTPUT Z	300-500K Ω	30-50K Ω	50-1000 Ω				
VOLTAGE GAIN	500-1500	300-1000	< 1				
CURRENT GAIN	< 1	25-50	25-50				
POWER GAIN	20-30 d B	25-40 dB	10-20 dB				

Figure 5-22. Transistor Biasing and Operating Characteristics

Transistor Type		Connect C	Measure			
		Pos. lead to				
	Small	emitter	base*	200 - 500		
PNP	Signal Power	emitter	collector	10K - 106K		
Ger-		emitter	base*	30 - 50		
manium		emitter	collector	several hundred		
	Small Signal	base*	emitter	1K - 3K		
		collector	emitter	very high (might read open)		
NPN Silicon		base	emitter	200 - 1000		
	Power	collector	emitter	high, often greater than 1 M		
*To test for transistor action, add collector-base short. Measured resistance should decrease.						

Table 5-3. Out-of-Circuit Transistor Resistance Measurement

c. Use a suction device (see Table 5-5) or wooden toothpick to remove solder from component mounting holes. DO NOT USE A SHARP METAL OBJECT SUCH AS AN AWL OR TWIST DRILL FOR THIS PURPOSE. SHARP OBJECTS MAY DAMAGE THE PLATED-THROUGH CONDUCTOR.

d. After soldering, remove excess flux from the soldered areas and apply a protective coating to prevent contamination and corrosion. See Table 5-5 for recommendations.

5-63. COMPONENT REPLACEMENT.

a. Remove defective component from circuit board.

b. Remove solder from mounting holes using a suction desoldering aid (Table 5-5) or wooden toothpick.

c. Shape leads of replacement component to match mounting hold spacing.

d. Insert component leads into mounting holes, and position component as original was positioned. DO NOT FORCE LEADS OF REPLACEMENT COMPO-NENT INTO MOUNTING HOLES. Sharp leads ends may damage plated-through conductor.

NOTE

Axial lead components, such as resistors and tubular capacitors, can be replaced without unsoldering. Clip leads near body of defective component, remove component and straighten leads left in board. Wrap leads of replacement component one turn around original leads. Solder wrapped connection, and clip off excess lead.

5-64. ETCHED CONDUCTOR REPAIR.

5-65. A broken or burned section of conductor can be repaired by bridging the damaged section with a length of tinned copper wire. Allow adequate overlap and remove any varnish from etched conductor before soldering wire into place.

		Open	Short	Lea	ıd
Ohmmeter	Safe Range(s)	Ckt Voltage	Ckt	Color	Pola- rity
412A	R x 1K R x 10K R x 100K R x 100K R x 1M R x 10M	1.0V 1.0V 1.0V 1.0V 1.0V 1.0V	1 mA 100 μA 10 μA 1 μA 0.1 μA	Red Blk	+
410C	R x 1K R x 10K R x 100K R x 100K R x 1M R x 10M	1.3V 1.3V 1.3V 1.3V 1.3V 1.3V	0.57mA 57 μA 5.7 μA 0.5 μA 0.05μA	Red Blk	+
410B	R x 100 R x 1K R x 10K R x 100K R x 100K R x 1M	1, 1V 1, 1V 1, 1V 1, 1V 1, 1V 1, 1V	1.1 mA 110 μA 11 μA 1.1 μA 0, 11μA	Blk Red	÷ -
Simpson 260	R x 100	1.5V	1 mA	Red Blk	+ -
Simpson 269	R x 1K	1.5V	0, 82 mA	Blk Red	+
Triplett 630	R x 100 R x 1K	1.5V 1.5V	3.25 mA 325 µA	Vari Se	es with erial
Triplett 310	R x 10 R x 100	1,5V 1.5V	750 μA 75 μA	Nun	nber

Table	5-4.	Safe	Ohmm	eter	Ranges	for	Transistor
		Resis	tance I	Meas	urement	s	

Table 5-5. Etched Circuit Soldering Equipment

Item	Use	Specification	Item Recommended	
Soldering Tool	Soldering Unsoldering	Wattage rating: 37,5 Tip Temp: 750-800°F Tip Size: 1/8'' OD	Ungar #776 Handle with Ungar #1237 Heating Unit	
Soldering Tip	Soldering	Shape: chisel	Ungar #PL113	
General Purpose	Unsoldering	Size: 1/8"		
De-soldering aid	Unsoldering multi- connection compo- nents (e.g., tube sockets	Suction device to remove molten solder from connection	Soldapult by the Edsyn Company, Arleta, California	
Resin (flux) solvent	Remove excess flux from soldered area before application of protective coating	Must not dissolve etched circuit base board ma- terial or conductor bonding agent	Freon Acetone Lacquer Thinner Isopropyl Alcohol (100% dry)	
Solder	Component replacement Circuit board repair Wiring	Resin (flux) core, high tin content (60/40 tin/lead), 18 gauge (SWG) pre- ferred		
Protective Coating	Contamination, corro- sion protection after soldering	Good electrical insulation, corrosion-prevention properties	Krylon #1302* Humiseal Protective Coat- ing, Type 1B12 by Columbia Technical Corp. Woodside 77, New York	

5-66. TRANSISTOR AND SEMICONDUCTOR DIODE REPLACEMENT.

a. Do not apply excessive heat. See Table 5-5 for soldering tool specifications.

b. Use a heat sink, such as pliers, between semiconductor body and hot soldering iron.

c. When installing a replacement semiconductor ensure sufficient lead length to dissipate heat of soldering by maintaining about the same length of exposed lead as used for original semiconductor.

d. Refer to Table 5-6 for checks following replacement of a semiconductor.

5-67. REPLACEMENT OF ELECTRON TUBES.

5-68. When replacing tubes in the Model 608E/F, it is recommended that a check be made of the operation of the instrument before and after each tube trial and, if no improvement is noted, the original tube be returned to the socket. Figure 5-23 locates all electron tubes in the equipment. Table 5-6 lists all tubes, together with the checks which should be made following tube replacement.

5-69. OSCILLATOR TUBE REPLACEMENT. Replacement of RF Oscillator tube V6 may affect the calibration of the frequency dial and the amplitude of the RF Generator heater-voltage supply. In addition, r

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Reference Designation	Check	Paragraph
V2	Internal Modulation	5~22
V6	Frequency Accuracy	5-12
	Frequency Drift	5-14
V8	Maximum RF Output	5-17
	Uncalibrated RF Output	5-20
V9	Calibrator Accuracy	5-13
V10	Calibrator Accuracy	5-13
V11	Calibrator Accuracy	5-13
V12	Power Supply (+225)	5-31
V13	Power Supply (+225)	5-31
V14	Power Supply (-165)	5-31
V15	Power Supply (-165)	5-31
V16	Power Supply (-165)	5-31
V18	Modulation Meter Accuracy	5-21
V19	Modulation Meter Accuracy	5-21
V20	Modulation Meter Accuracy	5-21
V21	Modulation Meter Accuracy	5-21
V22	Maximum RF Output	5-17
Q1	Power Supply	5-31
Q2	Leveling	5-18
Q 3	Leveling	5-18
2.4	Internal Modulation	5-22
Q4	Leveling	5-18
~	Internal Modulation	5-22
Q5	Leveling	5-18
	External Pulse Modulation	5-24
Q6	Leveling	5-18
	Internal Modulation	5-22
Q7	Leveling	5-18
00	Internal Modulation	5-22
Q8	Leveling	- 5-18
23	Internal Modulation	5-22
Q9	Leveling	5-18
010	Internal Modulation	5-22
Q10	Power Supply (+25)	5-31
Q11 CR1	Power Supply (+25) Crystal Calibrator Accuracy	5-31 5-13
CR2	Modulation Meter Accuracy	5-13
CR7	Leveling	5-21
CR8	Power Supply (+225)	5-31
CR9	Power Supply $(+225)$	5-31
CR10	Power Supply (+225)	5-31
CR11	Power Supply (+225)	5-31
CR12	Power Supply (-165)	5-31
CR13	Power Supply (-165)	5-31
CR14	Power Supply (-165)	5-31
CR15	Power Supply (-165)	5-31
CR16	Power Supply (-6.3)	5-31
CR17	Power Supply (-6.3)	5-31
CR18	Power Supply (-6.3)	5-31
CR19	Power Supply (-6.3)	5-31
CR20	Power Supply (-6.3)	5-31
CR21	Leveling	5-18
CR23	External Pulse Modulation	5-24
CR24	External Pulse Modulation	5-24
CR25	Leveling	5-18
CR26	Leveling	5-18
CR27	Power Supply (+25)	5-31
CR28	Power Supply (+25)	5-31
CR29	Power Supply (+25)	5-31
CR30	Power Supply (+25)	5-31

Table 5-6. Checks Following Tube	and Semiconductor Replacement
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plate current requirements of type 4042 pencil triodes may differ widely from tube to tube in a given application. Therefore, when V6 is replaced, heater voltage must be checked and, if necessary, reset to the proper value; plate current must be held to between 18 and 27 milliamperes by tube selection; and calibration of the frequency dial must be checked.

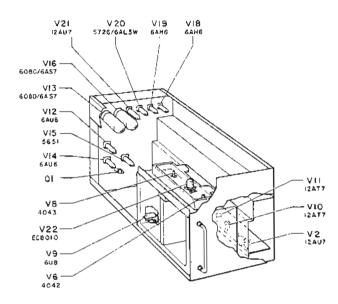


Figure 5-23. Location of Electron Tubes

5-70. To replace Oscillator V6, refer to Figure 5-24 and proceed as follows:

a. Remove frequency dial and top plate from RF Generator housing to gain access to tube compartment. (The frequency dial is accurately indexed on its hub by two pins to assure exact positioning when dial is replaced on hub.)

b. Remove socket from base of V6 by straight pull.

c. Remove cathode clip from tube.

d. Remove the two BH screws holding retainer plate, then remove plate and fiber spacer.

e. Lift tube gently from hole by straight pull.

f. Replace tube in reverse order of above steps.

g. Connect a dc voltmeter to inside lead of FL4. Meter should indicate 6.3 Vdc. If necessary, change value of CR20 to obtain 6.3 Vdc.

h. Using Model 428 DC Ammeter, measure current in lead to feed-through capacitor C15.

i. Set the FREQUENCY RANGE switch to band E. Current meter should indicate between 18 and 27 mA.

If meter indication is not within these limits, substitute another 4042 oscillator tube.

j. Using internal beat-frequency calibrator, check frequency calibration throughout range of signal generator, noting points that are significantly off frequency.

k. To correct calibration at the high-frequency end of all bands simultaneously, adjust trimmer capacitor C18 in the 608E, and C59 in the 608F, which is accessible in tube compartment in RF Generator housing. This adjustment has only minor effect at low-frequency ends of all ranges.

5-71. RF AMPLIFIER AND BUFFER TUBE RE-PLACEMENT. To remove RF Amplifier V8 and Buffer V22, refer to Figure 5-24 and proceed as follows:

NOTE

Replacement of the RF Amplifier and Buffer tubes can affect the amplitude of the RF generator heater-voltage supply and may limit the power output from the signal generator.

a. Remove frequency dial and top plate from RF Generator housing to gain access to tube compartment.

b. Replace Buffer V22.

c. For Amplifier V8, remove socket from base of V8 by straight pull.

d. Remove cathode clip from tube.

e. Using special wrench (located on instrument chassis, near RF Generator housing), loosen threaded retainer ring which holds V8 in housing. Remove retainer ring and neoprene washer.

f. Remove old tube and replace with new type 4043 tube.

g. Check power output throughout full frequency range of signal generator. With AMPL TRIMMER set for maximum output at each frequency, a full-scale reading should be obtained over entire frequency range.

5-72. CRYSTAL OSCILLATOR TUBE OR QUARTZ CRYSTAL REPLACEMENT. To gain access to the Calibrator Oscillator, remove the side cover on the RF Generator housing. The entire Crystal Oscillator is mounted on a bracket on the rear wall of the amplifier compartment. Location of oscillator tube V9, crystal Y1, and associated adjustments are shown in Figure 5-27. If V9 or Y1 is replaced, the calibrator should be checked and, if necessary, adjusted.

5-73. The crystal-controlled frequency calibrator is composed of two circuits: the 5-MHz crystal-controlled oscillator, which utilizes the pentode half of a type 6U8 tube, and the 1-MHz oscillator, utilizing the triode section of the 6U8. Each section is adjustable, and both sections should be adjusted at the same time. The frequency of the crystal can only be adjusted over a range of approximately 100 Hz. The frequency of the 1-MHz oscillator then locks in with the 5-MHz frequency and is either right on frequency or is very far off. To adjust the crystal-controlled calibrator, refer to paragraph 5-47.

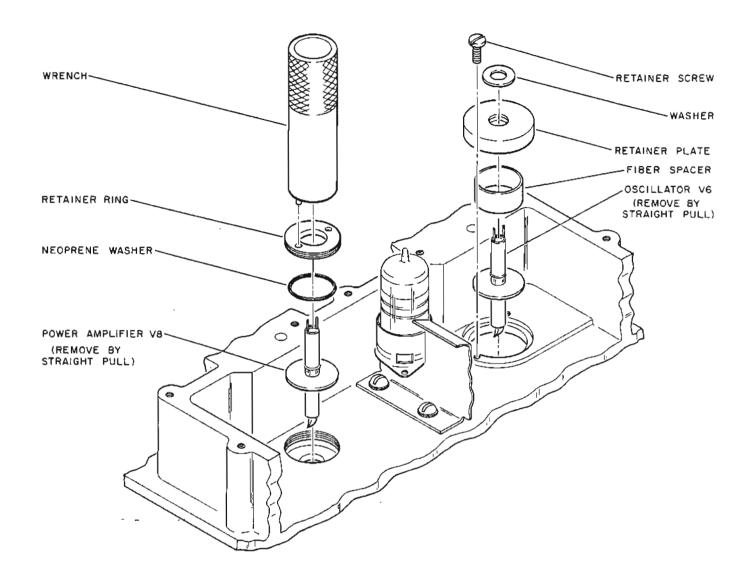


Figure 5-24. Replacement of Oscillator and Amplifier Tubes

5-74. ATTENUATOR PROBE REPAIR AND REPLACEMENT.

CAUTION

During removal and replacement of the probe, extreme care must be exercised. The probe consists of a cylindrical metal tube with a series of spring contact fingers around its periphery at one end. These fingers can be accidentally bent or twisted. It should be noted that one or two of the fingers are bent toward the center of the probe slightly; do not attempt to straighten them, since these fingers have been bent to assure clearance between the probe and the end of the guide slot in the attenuator housing. It is of greatest importance to make certain that the probe is not subjected to shock. If subjected to shock, the electrical components attached to the end of the probe can be broken or their positions altered, with a consequent change in the electrical characteristics of the probe.

5-75. GENERAL. The impedance-matching network in the output attenuator can be damaged if power from an external source is applied to the RF OUTPUT connector. If electrical components of the output attenuator are damaged, repair or replacement is necessary. Damage to the impedance-matching network may be confirmed by measuring the VSWR of the attenuator at the RF OUTPUT connector.

NOTE

Attenuator VSWR is determined by positioning of the probe variable capacitor (C37) and the relative position of the two resistors (R58 and R59) on the end of the probe. If replacement resistors are returned to their original positions, the VSWR will not be changed appreciably. However, if equipment is available for checking attenuator VSWR, such a check is advisable following replacement of resistors.

5-76. REMOVAL FROM HOUSING. If investigation shows an attenuator to be defective, proceed as follows:

a. Turn attenuator control on front panel until probe reaches end of its travel at rear of attenuator housing.

b. Remove nut and washers that hold drive cable in probe drive screw (Figure 5-25) in top of attenuator probe.

c. Remove probe split drive screw from probe body by removing inner nut and unscrewing.

d. Remove probe by carefully sliding it out of attenuator housing.

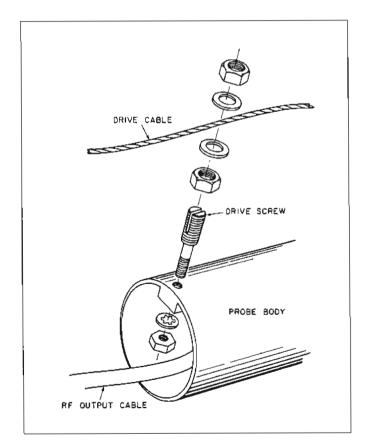


Figure 5-25. Disassembly of Attenuator Drive Cable

5-77. PROBE REPAIR. Capacitor C69 (Figure 5-26) is held in place by a mounting strap and a setscrew. To replace C69, unsolder the pickup loop at the capacitor end (ground slide of loop), loosen the setscrew, and slide the capacitor out of the clamp. Replace by reversing this procedure. The capacitor must be firmly seated in the hole in the end of the probe before tightening the setscrew. Be careful when tightening the setscrew; is tightened excessively.

5-78. Note location of the two resistors so that new ones can be mounted as nearly as possible in the same position. Install new resistors; duplicate the original resistor mounting positions and lead lengths as closely as possible. Make connections without excessive binding or pulling on ends of resistors as they are fragile.

CAUTION

These resistors are easily damaged by excessive heat. Use low-temperature soldering iron and solder connections as quickly as possible.

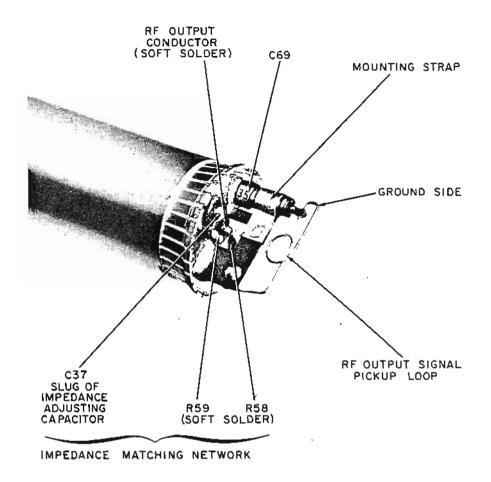
5-79. Measure resistors for any possible change in resistance due to soldering. The VSWR will not be changed appreciably.

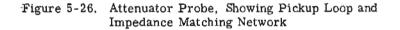
5-80. The VSWR can be checked at several frequencies on band E, and the position of resistors R_{05}^{+} and R58 adjusted for optimum VSWR. Adjustment on the band E will automatically correct for minimum VSWR on the lower bands. Laboratory test equipment, such as an impedance bridge or reflectometer setup, will be required for measuring the VSWR. These VSWR checks must be made with attenuator inserted in attenuator housing in the instrument, as results obtained will be slightly different with attenuator outside of the housing.

5-81. PROBE REPLACEMENT. If repair is not possible, both probe and cable must be replaced. Remove the RF OUTPUT connector from the front panel, and release the cable from the clamp holding the cable to the top of the side gusset. The entire probe assembly may then be removed from the instrument. Replacement probes are complete with cable and panel connector; and require no adjustment of the impedancematching network upon installation.

5-82. REPLACEMENT IN HOUSING.

a. Insert new or repaired probe in attenuator housing. Care must be taken in starting probe into housing, since diameter at probe contact fingers is slightly greater than inner diameter of housing. Contact fingers should be depressed slightly while starting probe into its housing.





CAUTION

Under no circumstances should the probe be forced into the housing.

b. Replace split drive screw in probe, making certain that screw slot is parallel to axis of housing.

c. Set attenuator drive cable in screw slot and replace both washers and nut. Do not tighten nut. Cable must move freely through the slot until probe penetration has been set.

d. Secure RF cable to clamp on side gusset.

e. Connect instrument to a source of 115-volt for 230 volt) power, and turn on POWER switch. Unless otherwise specified, operating controls should be set as follows:

MODULATION Selector	CW
Frequency dial	70 MHz
FREQUENCY RANGE	
AMPL TRIMMER . Adjust for max.	RF output
RF OUTPUT meter	7 dBm
MODULATION control	max. ccw
Attenuator	0 dBm

f. Connect a power meter (such as Model 432A), through a thermistor mount (Model 478A or equivalent), to RF OUTPUT connector.

g. Remove RF Generator side plate so that clearance between attenuator probe and coils of RF Amplifier tank may be observed.

CAUTION

The following step must be executed as carefully as possible to insure that the pickup loop does not make contact with any of the power amplifier coils. These coils are at B+ potential, and contact between any one of them and the attenuator pickup loop will damage attenuator components.

h. With attenuator dial set exactly on 0 dBm, manually advance attenuator probe into the housing until RF output signal is exactly 1 milliwatt (0 dBm) as read on external power meter. i. Tighten down nut on split drive screw so that probe may be actuated by its drive system. Carefully check to see that there is clearance between various amplifier coils and pickup loop when attenuator dial is set to +7 dBm.

5-83. REPLACEMENT OF LAMP DS1.

5-84. Lamp DS1 acts as a positive temperature coefficient resistance, maintaining Modulator Oscillator output constant. The characteristics of the type S6 lamps used for this purpose ordinarily vary widely from lamp to lamp, causing a corresponding variation in the amplitude of the oscillator output from one Model 608E/F to another. To compensate for differences between lamps, potentiometer R7 is provided for adjustment of the oscillator output voltage level. After lamp I1 has been replaced, the oscillator output voltage, as measured at pin 3 of V2, should be adjusted to 2 volts rms; if it cannot be adjusted to this value, another lamp must be tried.

5-85. LOCATION OF PARTS.

5-86. To enable maintenance personnel to properly identify and locate replacement parts during adjustments, troubleshooting, and repair periods Figures 5-27 through 5-33 are presented at the rear of this section. A parts identification diagram for the ALC board is contained on Figures 7-3, 7-4 and 7-5.

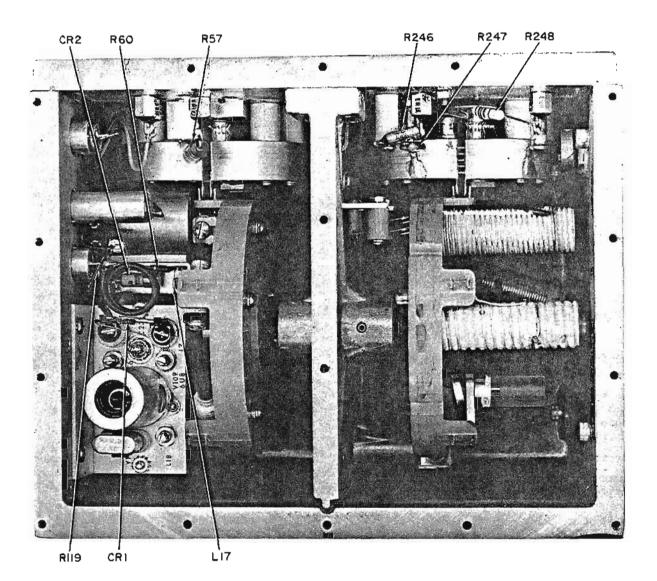
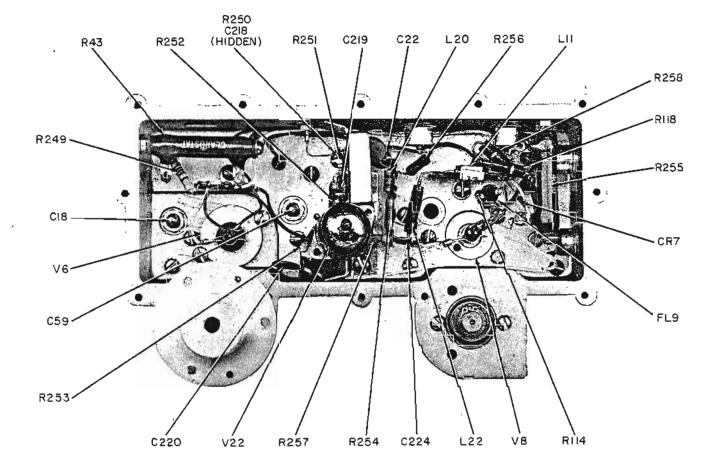
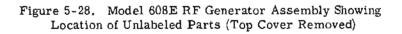


Figure 5-27. RF Generator Assembly Showing Location of Unlabeled Parts (Side Cover Removed)





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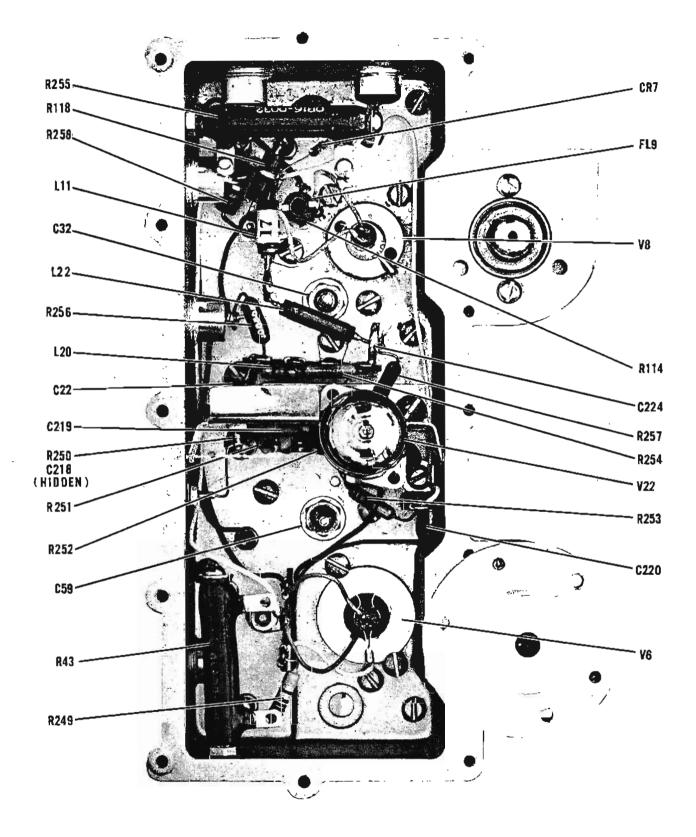


Figure 5-29. Model 608F RF Generator Assembly Showing Location of Unlabeled Parts (Top Cover Removed)

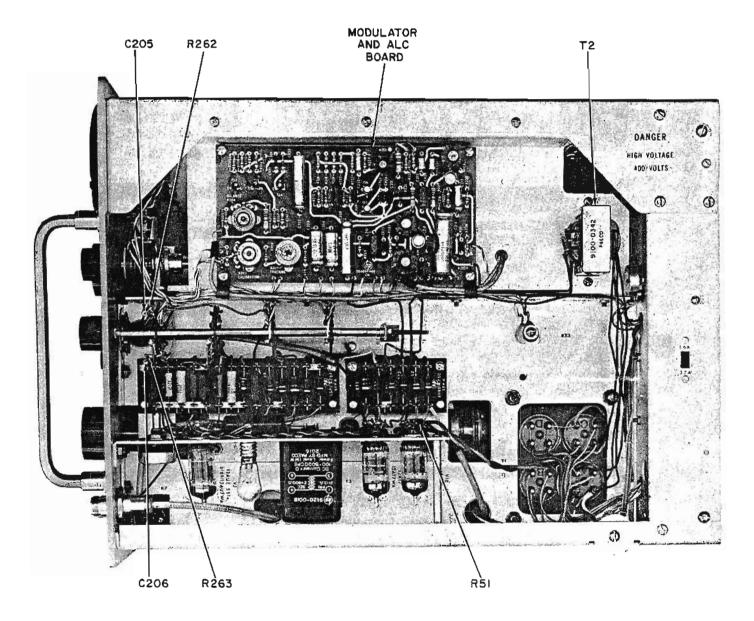


Figure 5-30. Right Side View Showing Location of Unlabeled Chassis Components

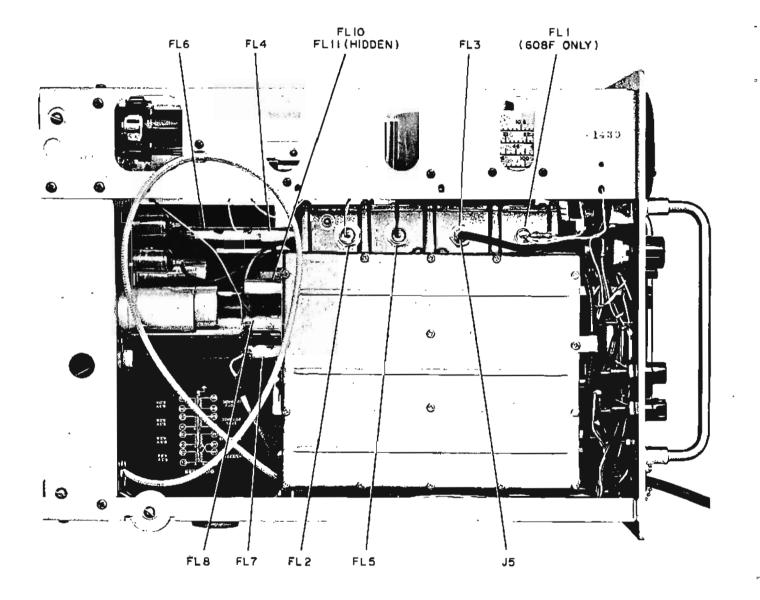


Figure 5-31. Left Side View Showing Location of Unlabeled Components

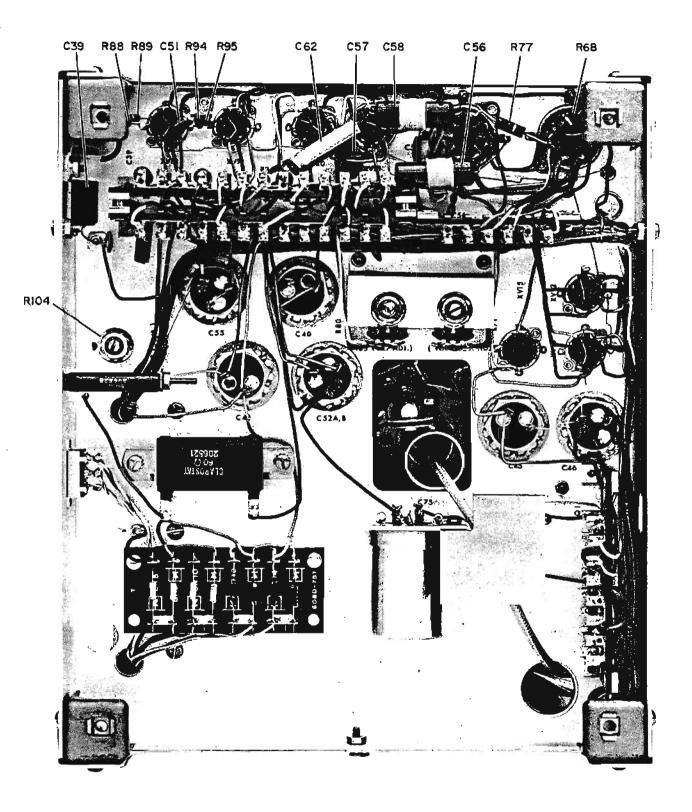
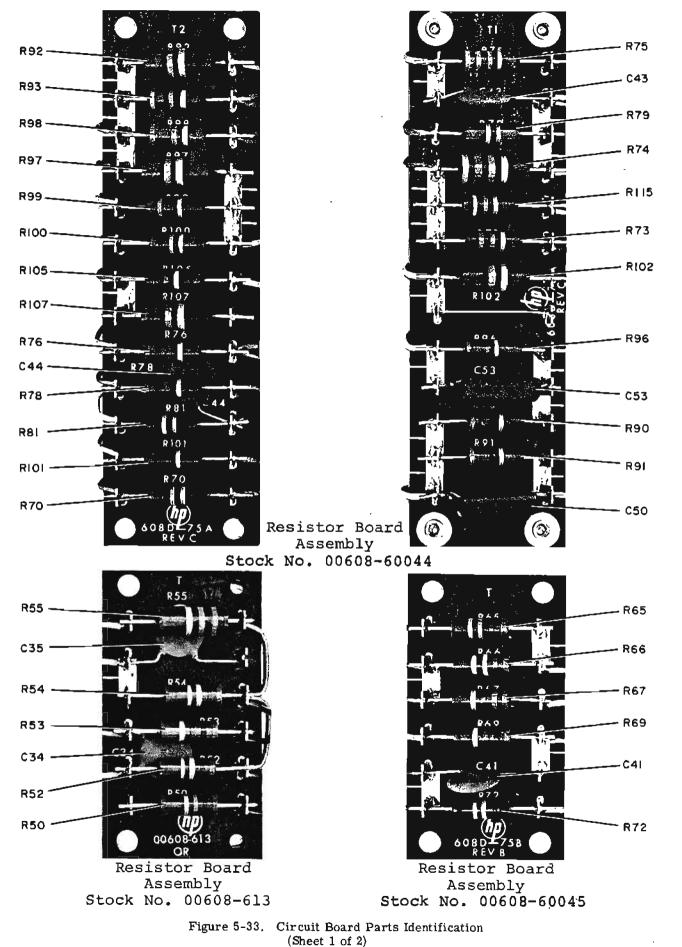
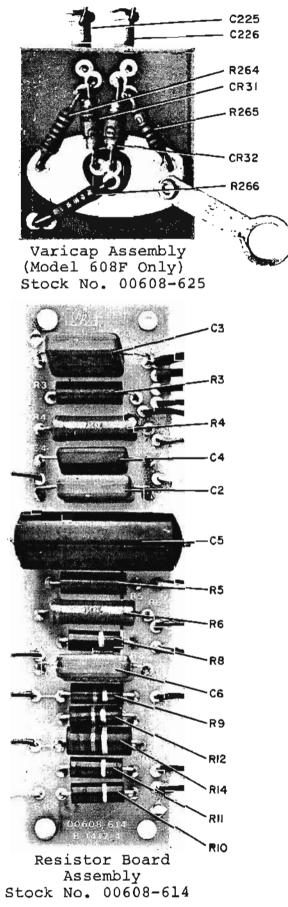
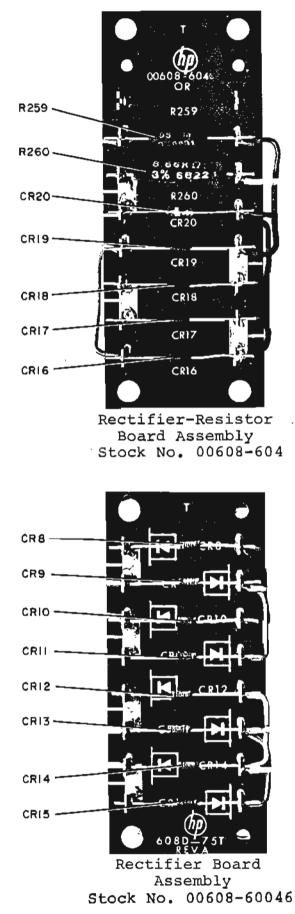


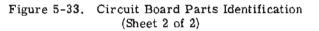
Figure 5-32. Rear View Showing Location of Unlabeled Chassis Components



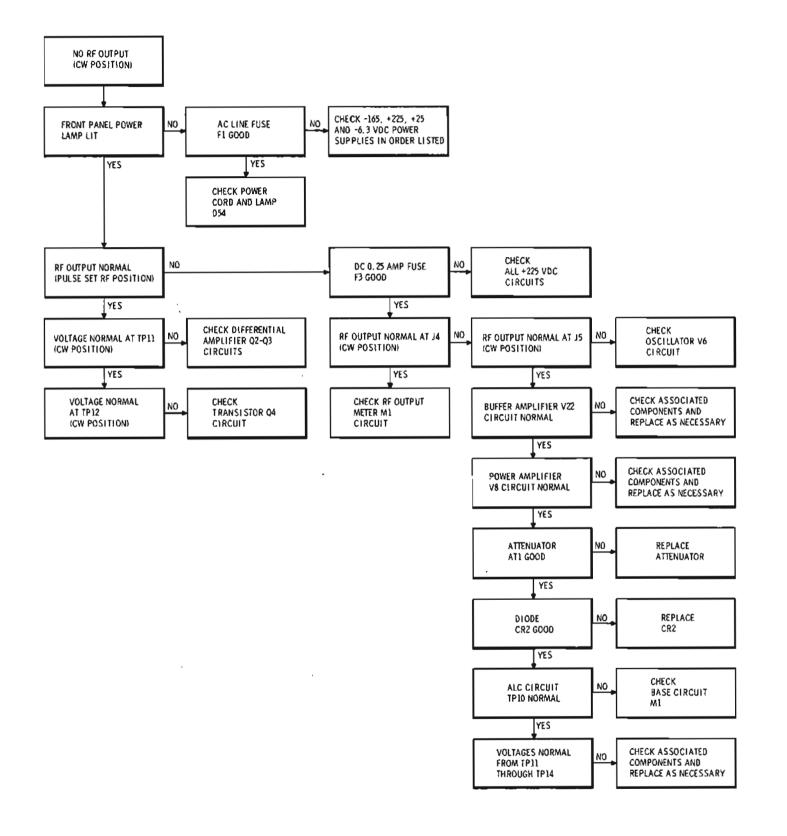
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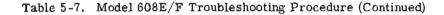


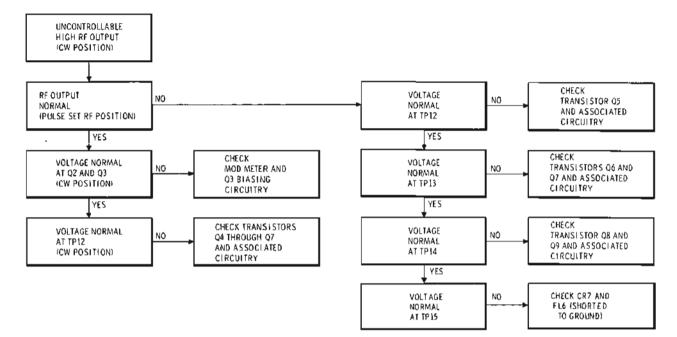


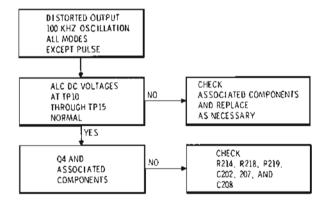
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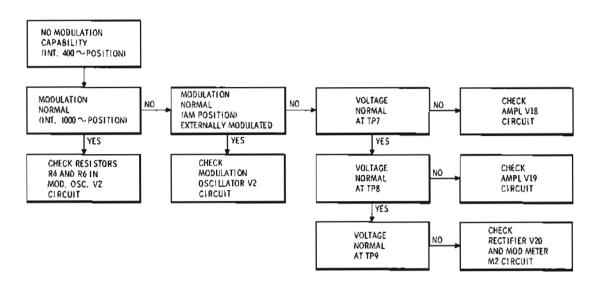
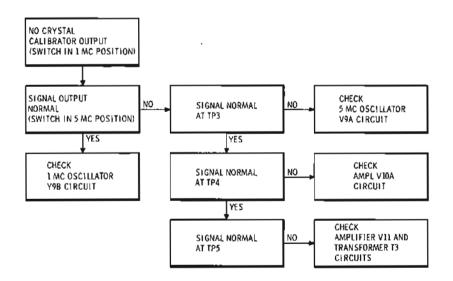


Table 5-7. Model 608E/F Troubleshooting Procedure (Continued)



SECTION VI REPLACEABLE PARTS

6-1. INTRODUCTION

6-2. This section contains information for ordering parts. Table 6-2 lists abbreviations used in the parts list and throughout the manual. Table 6-3 lists all replaceable parts in reference designator order. Table 6-4 contains the names and addresses that correspond to the manufacturer's code numbers.

6-3. REPLACEABLE PARTS LIST

6-4. Table 6-3 is the list of replaceable parts and is organized as follows:

a. Electrical assemblies and their components in alphanumeric order by reference designation.

b. Chassis-mounted parts in alphanumeric order by reference designation.

c. Mechanical parts.

6-5. The information given for each part consists of the following:

a. The Hewlett-Packard part number.

b. Part number check digit (CD).

c. The total quantity (Qty) for the entire instrument.

d. The description of the part.

e. A typical manufacturer of the part in a five-digit code.

f. The manufacturer's number for the part.

6-6. ORDERING INFORMATION

6-7.To order a part listed in the replaceable parts table, quote the Hewlett-Packard part number (with the check digit), indicate the quantity required, and address the order to the nearest Hewlett-Packard office. The check digit will ensure accurate and timely processing of your order.

6-8. To order a part that is not listed in the replaceable parts table, include the instrument model number, instrument serial number, description and function of the part, and the number of parts required. Address the order to the nearest Hewlett-Packard office.

6-9. FACTORY SELECTED PARTS (*)

6-10. Parts marked with an asterisk (*) are factory selected parts. The value listed in the parts list is the nominal value.

NOTE

Within the USA, it is better to order directly from the HP Parts Center in Mountain View, California. Ask your nearest HP office for information and forms for the "Direct Mail Order System".

REFERENCE DESIGNATIONS

 A assembly AT attenuator; isolator; termination B fan; motor BT battery C capacitor CP coupler CR diode; diode thyristor; varactor DC directional coupler DL delay line DS annunciator; signaling device (audible or visual); lamp; LED 	E miscellaneous electrical part F fuse FL filter H hardware HY circulator J electrical connector (stationary portion); jack K relay L cofi; inductor M meter MP miscellaneous mechanical part	 P electrical connector (movable portion); plug Q transistor: SCR; triode thyristor R resistor RT thermistor S switch T transformer TB terminal board TC thermocouple TP test point 	 U, integrated circuit; microcircuit V electron tube VR voltage regulator; breakdown diode W cable; transmission path; wire X socket Y crystal unit (piezo- electric or quartz) Z tuned cavity; tuned circuit
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Table 6-2. Abbreviations (1 of 2)

ABBREVIATIONS

A ampere	COEF coefficient
ac alternating current	СОМ соттоп
ACCESS accessory	COMP composition
ADJ adjustment	COMPL complete
A/D analog-to-digital	CONN connector
AF audio frequency	CP cadmium plate
AFC Automatic	CRT cathode-ray tube
frequency control	CTL , complementary
AGC automatic gain	transistor logic
control	CW continuous wave
AL aluminum	cw clockwise
ALC automatic level	cm centimeter
control	D/A digital-to-analog
control	
AM amplitude modula-	dB, decibel
tion	dBm decidel referred
AMPL amplifier	to 1 mW
APC automatic phase	dc direct current
control	deg degree (temperature
control ASSY assembly	interval or differ-
AUX auxiliary	
avg average	o ence)
	degree (pane
AWG American wire	o angle) C degree Celsius
gauge	C degrée Celsius
BAL balance	c (centigrade)
BCD binary coded	C degree Celsius (centigrade) F degree Fahrenheit
decimal	K degree Kelvin
BD board	DEPC deposited carbon
BE CU beryllium	DET detector
copper	diam diameter
copper	
BFO beat frequency	DIA diameter (used in
oscillator -	parts list)
BH binder head	DIFF AMPL differential
BKDN breakdown	amplifier
BP bandpass	div division
BPF bandpass filter	DPDT double-pole,
BRS brass	double-throw
	DR dríve
BWO backward-wave	
oscillator	DSB double sideband
CAL calibrate	DTL diode transistor
ccw counter-clockwise	logic
CER ceramic	DVM agata voltmeter
CHAN	ECL emitter coupled
cm centimeter	logic
CMO cabinet mount only	EMF . electromotive force
COAX coaxial	
COAA COAXIA	

	. electronic data
pre	ocessing
ELECT	electrolytic
ENCAP	., encapsulated
	external
F	
	farad
	field-effect
tra	nsistor
F/F	flip-flop
FH	flat head
FILH	fillister head
FM freq	uency modulation
FP	front panel
FREG	frequency
	fixed
E	gram
GE	germanium
GHz ,	gigahertz
GL . ,	glass
	ground(ed)
н	henry
b	hour
HET	heterodyne
	hexagonal
HD	head
HDW	hardware
HF	high frequency
HG	mercury
н	high
HP	Hewlett-Packard
HPF	. high pass filter
HR	hour (used in
nn	
	rts list)
	high voltage
Hz	Hertz
IC	integrated circuit
ID	. inside diameter
IF	intermediate
	quency
	impregnated
in	inch
ш	inch
INCD	incandescent
	include(s)
INP	input
INS	insulation

INT internal
kg kilogram
RHz kilohertz
k\4kilohm
kV, kilovolt
lbpound
LC inductance-
capacitance
LED light-emitting diode
LF low frequency
LG long
LH left hand
LIM limit
LIM linear taper (used
in narte liet)
lin linear
lin linear LK WASH lock washer
LO low; local oscillator
LOG logarithmic taper
(used in parts list)
log logrithm(ic)
log logrithm(ic) LPF low pass filter
LV low voltage
m meter (distance)
mA milliampere
mA milliampere MAX maximum MΩ megohm
MΩ megohm
MEG meg (10^6) (used)
MEG meg (10°) (used
MEG meg (10°) (used in parts list)
MEG meg (10°) (used in parts list) MET FLM metal film
MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide
MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency;
MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in
MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list)
MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallle oxide MF medium frequency; microfarad (used in parts list) MFR manufacturer
MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list) MFR, manufacturer mg, milligram
MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list) MFR manufacturer mg milligram MHz megahert2
 MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list) MFR manufacturer mg milligram MHz megahertz mH millihenry
 MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list) MFR manufacturer mg milligram MHz megahertz mH millihenry mho mho
 MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list) MFR manufacturer mg milligram MHz milligram MHz millihenry mho mho
 MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list) MFR manufacturer mg milligram MHz milligram MHz millihenry mho mho
 MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list) MFR manufacturer mg milligram MHz megahert2 mH millihenry mho mho MIN minute (tlme) minute (plane
MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list) MFR, manufacturer mg milligram MHz milligram MHz milligram MHz millihenry mho minilihenry mho
 MEG meg (10°) (used in parts list) MET FLM metal film MET OX metallic oxide MF medium frequency; microfarad (used in parts list) MFR manufacturer mg milligram MHz megahert2 mH millihenry mho mho MIN minute (tlme) minute (plane

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NOTE All abbreviations in the parts list will be in upper-case.

MOD , modulator MOM momentary
MOM momentary
MOM momentary MOS metal-oxide
semiconductor
sentconductor
ms millisecond MTG mounting
MTG mounting
MTR meter (indicating
device)
mV million)t
mVac millivolt, ac mVac millivolt, ac mVdc millivolt, dc
mvac munvoit, ac
mVdc millivolt, dc
mVpk millivolt, peak
mVpk millivolt, peak mVp-p millivolt, peak-
to-peak
nVrms millivolt, rms
mW
MUX multiplex
MY mylar μA microampere
μA microampere
UF microfarad
UH microbenry
μF microfarad μH microhenry μmbo micromho
us microsecond
μs microsecond
μV microvolt
μ Vac microvolt, ac
$\mu V dc$, , , , microvolt, dc
UVpk microvolt, peak
UVD-D microvolt, peak-
μVac microvolt, ac μVdc microvolt, dc μVpk microvolt, peak μVp-p microvolt, peak
to-peak
UVrms microvolt mrs
μVrms microvolt, rms μW microwatt nA nanoampere NC no connection N/C normally closed
μVrms microvolt, rms μW microwatt nA nanoampere NC no connection N/C normally closed
μVrms microvolt, rms μW microwatt nA nanoampere NC no connection N/C normally closed NE neon
WVrms microvolt, rms //W microvolt, rms //W microwatt nA nsnoampere NC no connection N/C normally closed NE neon NEG negative
μVrms microvolt, rms μW microvolt, rms μW microwatt nA ngnoampere NC no connection N/C normally closed NEG negative nF ngnofarad
μVrms microvolt, rms μW microvolt, rms μW microwatt nA ngnoampere NC no connection N/C normally closed NEG neon NEG ngative nF nanofarad NI PL nickel plate
VVTMS microvolt, rms µVTMS microvolt, rms µW microwatt nA nanoampere NC no connection N/C no connection N/C no connection N/C neon NEG neon NEG neofarad NI PL nickel plate N/O normally open
VVTMS microvolt, rms µVTMS microvolt, rms µW microwatt nA nanoampere NC no connection N/C no connection N/C no connection N/C neon NEG neon NEG neofarad NI PL nickel plate N/O normally open
LOPpeak LWrms microvolt, rms LW microvolt, rms LW microvolt, rms LW microvolt, rms NC no connection N/C no connection N/C no connection N/C normally closed NEG neon NEG nanofarad NI PL nickel plate N/O normally open NORM normal
LOPpeak LWrms microvolt, rms LW microvolt, rms LW microvolt, rms LW microvolt, rms NC no connection N/C no connection N/C no connection N/C normally closed NEG neon NEG nanofarad NI PL nickel plate N/O normally open NORM normal
LOPPeak UVrms microvolt, rms IW microvolt, rms IW microvolt, rms IW nanoampere NC no connection N/C normally closed NE neon NEG neon NEG nanofarad NI PL nickel plate N/O normally open NOM normal NORM normal NPN negative-positive-
LOPPeak UVrms microvolt, rms IW microvolt, rms IW microvolt, rms IW nanoampere NC no connection N/C no connection NEG neon NEG nanofarad NI PL nickel plate N/O normally open NOM normal NPN negative-positive- negative negative
μVrms microvolt, rms μW microvolt, rms μW microvolt, rms μW microvolt, rms nA nanoampere NC no connection N/C normally closed NEG nanofarad NI PL nanofarad NI PL normally open NOM normal NORM normal NPN negative-positive- negative NPO
UVTMS microvolt, rms UVTMS microvolt, rms IW microwatt nA nanoampere NC no connection N/C normally closed NE neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG normally open NOM normal NORM normal NPN negative-positive- negative-positive- zero (zero tempera-
UVrms microvolt, rms UVrms microvolt, rms UW microwatt nA nanoampere NC normally closed NEG negative nF nanofarad NI PL nickel plate N/O normally open NOM normally open NOM normally open NOM normal NPN negative-positive- negative NPO negative-positive- negative NPO negative-positive- zero (zero tempera- ture coefficient)
UVINS microvolt, ITMS UVINS microvolt, ITMS UW microwatt nA nanoampere NC normally closed NE normally closed NE negative nF negative nF normally open NOM normally NOM normally NOM normally NOM normal NOM normal NOM normal NOM normal NOM normal NOM normal NOM normal NOM normal NOM
UVrms microvolt, rms UVrms microvolt, rms UW microwatt nA nanoampere NC normally closed NEG negative nF nanofarad NI PL nickel plate N/O normally open NOM normally open NOM normally open NOM normal NPN negative-positive- negative NPO negative-positive- negative NPO negative-positive- zero (zero tempera- ture coefficient)
UVINS microvolt, ITMS UVINS microvolt, ITMS UW microwatt nA nanoampere NC normally closed NE normally closed NE negative nF negative nF normally open NOM normally NOM normally NOM normally NOM normal NOM normal NOM normal NOM normal NOM normal NOM normal NOM normal NOM normal NOM
LO-peak UVrms microvolt, rms UW microvolt, rms UW microvolt, rms UW manoampere NC no connection N/C no connection N/C normally closed NEG neon NEG nanofarad NI PL nanofarad NI PL normally open NOM normally open NOM normally NPN negative-positive- negative positive- network not recommended for field replace- ment
UVrms microvolt, rms UVrms microvolt, rms IW microvolt, rms NC no connection N/C normally closed NEG neon NEG neon NCM normal NORM normal NORM normal NPN negative-positive- negative-positive- zero (zero tempera- ture coefficient) NRFR not recommended for field replace- ment NSR not separately
UVrms microvolt, rms UVrms microvolt, rms UW microwatt nA nanoampere NC no connection N/C normally closed NE neon NEG negative nF negative nF nickel plate N/O normally open NOM normally open NOM normally open NOM normally open NOM normally open NOR normally open NOR normally open NOR normally open NOR normally open NPO negative-positive zero (zero tempera- ture coefficient) NR FR not recommended for field replace- ment NSR not separately replaceable
UVrms microvolt, rms UVrms microvolt, rms UW microvolt, rms UW manoampere NC normally closed NEG normally closed NEG normally closed NF normally closed NI PL negative nF normally open NOM normally open NPO negative-positive zero (zero tempera- ture coefficient) NRFR . not recommended for field replace- ment NSR not separately replaceable ns nanogecond
UVTMS microvolt, rms UVTMS microvolt, rms IW microwatt nA nanoampere NC no connection N/C normally closed NE neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG neon NOM neon NOM normal NORM normal NORM normal NPN negative-positive negative positive zero (zero tempera- ture coefficient) NRFR not recommended for field replace- ment NSR not separately replaceable ns
JUVINS microvolt, ITAS JUW microvolt, ITAS JUW microwatt nA nanoampere NC no connection N/C normally closed NE neon NEG neon NEG negative nF negative nF nickel plate N/O normally open NOM normal NORM normal NPN negative-positive negative-positive zero (zero tempera- ture coefficient) NRFR not recommended for field replace- ment NSR not separately replaceable ns
UVTMS microvolt, rms UVTMS microvolt, rms IW microwatt nA nanoampere NC no connection N/C normally closed NE neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG neon NEG neon NACM neon NOM normal NORM normal NORM normal NPN negative-positive negative positive zero (zero tempera- ture coefficient) NRFR not recommended for field replace- ment NSR not separately replaceable ns

OD outside diameter
OH oval head OP AMPL operational
OP AMPL operational
amplifier
OPT option
OSC oscillator
OX oxide
oz ounce
Ωοhm
P peak (used in parts
list)
modulation
PC printed circuit
PCM pulse-code modula-
tion; pulse-count
modulation
PDM pulse-duration
modulation
pF picofarad PH BRZ phosphor bronze PHL Phillips
PHL Phillips
PIN positive-intrinsic-
negative
PIV peak inverse
voltage
pk
PL phase lock
PL phase lock PLO phase lock
0 scillator
PM phase modulation PNP positive-perative-
PNP positive-negative-
positive
PORC porcelain
POS . , positive; position(s)
(used in parts list)
POSN position
POT
POT potentiometer p-p peak-to-peak
PP peak-to-peek (used
in parts list)
PPM pulse position
modulation
PREAMPL preamplifier
PRF pulse-repetition
frequency
PRR pulse repetition
rate
pspicosécond
PT point
PT point PTM pulse-time
modulation
PWM pulse-width
modulation
modemelon
N
NI

PWV peak working
voltage
RC resistance-
capacitance
D D OT
RECT rectifier REF reference REG regulated
REF reference
REG regulated
REPL replaceable
REPL replaceable
RF radio frequency RFI radio frequency
RFI radio frequency
interference
RH round head; right
hand
RLC resistance-
rebo resistance-
inductance-
capacitance
RMO rack mount only
rms root-mean-square
rms root-mean-square RND round
ROM read-only memory
R&P rack and panel
R&P rack and panel RWV reverse working
RWV reverse working
VOILARE
S scattering parameter
s second (time)
s second (unite)
" . second (plane angle)
S-B , slow-blow (fuse)
(used in parts list)
SCR silicon controlled
rectifier; screw
SE selenium SECT, sections SEMICON semicon-
SECT sections
SEMICON semicon-
durant durant
ductor
SHF superhigh fre-
quency
SI silicon
SIL silver
SIL
SL slide
SIL silver SL slide SNR signal-to-noise ratio
SNR signal-to-noise ratio
SPDT signal-to-noise ratio
SPDT signal-to-noise ratio SPDT single-pole, double-throw
SPDT signal-to-noise ratio SPDT single-pole, double-throw
SPDT signal-to-noise ratio SPDT single-pole, double-throw
SNR signal-to-hoise ratio SPDT single-pole, double-throw SPG spring SR split ring
SNR signal-to-noise ratio SPDT single-pole, double-throw SPG spring SR split ring SPST single-pole,
SNR signal-to-hoise ratio SPDT single-pole, double-throw SPG spring SR split ring SPST single-pole, circle-throw
SNR signal-to-hoise ratio SPDT single-pole, double-throw SPG spring SR split ring SPST single-pole, circle-throw
SNR signal-to-hoise ratio SPDT single-pole, double-throw SPG spring SR split ring SPST single-pole, circle-throw
SNR signal-to-hoise ratio SPDT single-pole, double-throw SPG spring SR split ring SPST single-pole, circle-throw
SNR signal-to-hoise ratio SPDT single-pole, double-throw SPG spring SR split ring SPST single-pole, circle-throw
SNR

TD time delay
TEDM towning
TERM terminal
TFT thin-film transistor
TGL toggle
THRU thread THRU through TI titanium TOL tolerance TRIM trianistor
THRU through
WY Altenium
II nianium
TOL tolerance
TRIM trimmer
TSTR transistor
TTL . , transistor-transistor
logic
TV television
TV television
TVI television interference TWT . traveling wave tube
TWT traveling wave tube
U micro (10 ⁶) (used
in parts list)
parts list)
UHF ultrahigh frequency
UNREG, unregulated
V volt
VA voltampere
Vie voite voite pere
Vac volts, ac VAR variable
VAR
VCO voitage-controlled
oscillator
Vdc, volts, dc VDCW. volts, dc, working
VDCW, volte de working
(used in parts list)
(used in parts inst)
V(F) volts, filtered
VFO variable-frequency
oscillator
VHF very-high fre-
quency
Vpk volts, peak
Vp-p volts, peak-to-peak
vp-p voits, peak-to-peak
Vrms volts, rms VSWR voltage standing
VSWR voltage standing
wave ratio
VTO voltage-tuned
oscillator
VTVM vacuum-tube
voltmeter
V(X) volts, switched
W
W/ with
WIV working inverse
voltage
WW wirewound
W/O without
W/O without YIG yttrium-iron-garnet
Zo characteristic
impedance

NOTE

All abbreviations in the parts list will be in upper-case.

MULTIPLIERS

Abbreviation	Profix	Multiple
Т	tera	1012
G	giga	109
м	mega	106
k	kilo	103
da	deka	10
đ	deci	10-1
c	centi	10^{-2}
Π.	milli	10-3
μ	micro	10-6
л	nano	10-9
p	pico	10-12
f	femto	10-15
a	atto	10-18

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Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
AT1	00608-610 5040-0232	0	1	ATTENUATOR PLUNGER ASSEMBLY WINDOW-DIAL ATTENUATOR	28490 29480	00608-610 3040-0232
C1 C2 C3 C4 C5	0140-0071 0140-0071 0140-0025 0140-0085 0160-0016	1 3 7 0	2 1 1	NOT ABBIGNED Capacitor-fxd 5600pf +-1% 500vdc Mica Capacitor-fxd 5600pf +-1% 500vdc Mica Capacitor-fxd 470pf +-3% 500vdc Mica Capacitor-fxd .22uf +-10% 400vdc	72136 72136 72136 28480	DH3DE562F0500W1CR DH3DE562F0500W1CR DH2DE471J0560W1CR 0160-0018
C6 C7-	0140-0020	2	3	CAPACITOR-FXD 1800PF +~10% S00VDC MICA	72136	DM30E182K0500WV1CR
C13 C14 C15	0150-0019 0150-0019	0	5	NOT ASSIGNED Capacitor-Fothru 1000pf 20% 500v Cer Capacitor-Fothru 1000pf 20% 500v Cer	72982 72982	327-005-X500-102H 327-005-X500-102H
C16 C17	0150-0028	1	1	CAPACITOR-STDOFF 100PF 10% 300V CER Tuning Capacitor	28480	0130-0058
C18	0133~0091	•	4	THIS CAPACITOR IS NUT REPLACEABLE Capacitur-V trmr-PSTN .3-3PF 500V (Hodel 608E)	28480	0133-0001
C17 C20 C21 C22 C23	0150~0014 0150~0010 0150-0010 0150-0089 0133~0002	5 1 1 8 5	1 2 1	CAPACITOR-FXD 5000PF +100-0X 500VDC CER CAPACITOR-FXD 47PF +-SX 500VDC CER 0→30 CAPACITOR-FXD 47PF +-SX 500VDC CER 0→30 CAPACITOR-FXD 10PF +-SPF 500VDC CER CAPACITOR-V TRMR-P6TN 1-12PF 500V	28480 04222 04222 04222 28480	0150-0014 51-27NPO-47-5 81-27NPO-47-5 C1-1NPO-10-1/2 0133-0002
C24 C25	0150-0015	6	1	CAPACITOR-FXD 2.2PF +-1DZ 500VOC TI DIDX NSR - Part of V6	28480	0150-0615
C26 C27 C28	0150-0012	3	ÿ	CAPACITOR-FXD .01UF +-20% 1KVDC CER Not Assigned Capacitor-FdThru 1000pf 20% 500v CER	56289 72982	C023A102J103HS38 327-003-X5U0-102H
C29 C30 C31 C32	0160~2236 0150-0019 0133-0001	B 0	1	NOT ASSIGNED Capacitor-FXD 1PF +25PF 500UDC CER Capacitor-Fothru 1000PF 20% 500U CER Capacitor-U TRMR-PSTN .5-3PF 500U (Model 400F ONLY)	28490 72982 29480	0160-2236 327~005-X5U0~102H 0133-0001
C33 C34 C35 C36 C37	6080-82 0150-0012 0150-0012 0160-0013	8 10 10	1	CAPACITOR-FXD 60PF 10% 600VDCW CAPACITOR-FXD .01UF +-20% 1KVDC CER CAPACITOR-FXD .01UF +-20% 1KVDC CER CAPACITOR-FXD .1UF +-10% 4060VDC NGR - PART OF ATTEN. PROBE AGGEMBLY	28480 56289 56289 28480	6080-82 C023A1023103MS38 C023A1023103MS38 0160-0013
C38 C39 C40 C41 C42	0140-0069 0160-0013 0180-0020 0150-0012 0150-0012	910 43 6	2 2	CAPACIYOR-FOTHRU 350FF 10% 500V HICA Capacitor-FXD (10+ +-)04 40000 Capacitor-FXD 800F+109-20% 450VDC AL Capacitor-FXD 800F+109-20% 450VDC AL Capacitor-FXD 800F+109-20% 450VDC AL	72982 28480 28480 56289 28480	665~034~01A4~551X 0160-0013 0160-0020 C023A102J10JHS38 0100-0020
C43 C44 C43 C45A C45A	0150-0012 0150-0012 0180-0019 0180-0019	NGGG	1 3	CAPACITOR-FXD .01UF +-20% 1XVDC CER CAPACITOR-FXD .01UF +-20% 1XVDC CER CAPACITOR-FXD 45UF+50-10% 450VDE AL CAPACITOR-MLTSECT 10/10UF 450V THD MTG NOT USED: NSR - PART OF C46A	56289 56289 20480 28480	C023A102J103HS38 C023A102J103HS38 0180-0019 0180-0018
C47 C48 C49 C50 C31	0130-0012 0140-0020 0130-0012	u n u		NOT ASSIGNED NOT ASSIGNED CAPACITOR-FXD .01UF →~20% 1KVDC CER CAPACITOR-FXD 1800PF →-10% S00VDC MICA CAPACITOR-FXD .01UF +~20% 1KVDC CER	56289 72136 56289	C023A102J103H538 DH30E182K0500WUSCR C023A102J103H538
C52A C52B	0180-0018	2		CAPACITOR-MLTSECT 10/1085 450V THD MTG NSR - Part of C52A	28480	0180-0018
C53 C54 C55	0140-0020 0180-0018	2		CAPACITOR-FXD 1000PF +-10% 500VDC MICA NOT A881GNED Capacitor-MLTGECT 10/10UF 450V THD NTG	72136 28460	DM30E182K8500WV1CR 0180-8818
C56 C57 C58 C59 C60	0160~0013 0150-0012 0160-0013 0133-0001 0150-0008	53547	3	CAPACITOR-FXD .1UF +-10X 400VDC CAPACITOR-FXD .01UF +-20X 1KVDC CER CAPACITOR-FXD .1UF +-10X 400VDC CAPACITOR-V TRMR-PSTN .5-3PF 500VD CAPACITOR-FXD 5FF +5PF 500VDC CER	28480 56287 28480 28480 04222	0160-0013 C023A102J103KS38 0160-0013 0133-0001 C1-1N330-5-1/2
C61 C62	0120-0094	4	з	NOT AGGIENED Capacitor-FXD 1000F+75-10% 230DC AL	56289	3001076025002
C63- C67 C69	608D-59H	1	1	NOT ASSIGNED Special trimmer;fine freq, adjust	28480	40 8D-59H
C69 C70 C71 C72 C73	0150~0006 0160-4056 0150-0037 ' 0150-0032	5227	1 1 1 3	NSR - PART OF ATTEN. PROBE ASSEMBLY CAPACITOR-FXD 110PF +-20% 3000DC CER CAPACITOR-FXD 75PF +-1% 5000DC CER 0+-30 CAPACITOR-FXD 100PF +-1% 5000DC CER CAPACITOR-FXD 10PF +-1PF 5000DC CER	04222 28480 28480 28480 28480	81-19NPO-110-20 0160~4856 0150~8037 0150~0032

Table 6-3. Replaceable Parts

0150-0019 0150-0057 0150-0074 0150-0074 0180-0074 0180-0074 0180-0105 0180-0105 0180-0119 0140-0119 0140-0146 0140-0146	945 33498 4	1 1 3 2	CAPACITOR-FDTHRU 1000PF 20X 500V CER CAPACITOR-FXD 16PF +-3X 500VDC CER 0+-30 CAPACITOR-FXD .1UF +80-20X 50VDC CER NOT ASSIGNED CAPACITOR-FXD .05UF +80-20X 100VDC CER CAPACITOR-FXD .05UF +80-20X 100VDC CER	72982 28480 28480	327-005-X500-102X 0150-0057 0150-0121
0150-0096 0180-0197 0180-0105 0180-0105 0180-0105 0180-0119 0180-0119 0140-0146 0180-0094	3 4 8 8		CAPACITUR-FXD .05UF +80-20% 100VDC CER		
0150-0096 0180-0197 0180-0105 0180-0105 0180-0105 0180-0119 0180-0119 0140-0146 0180-0094	3 4 8 8				
0140-0119 0140-0146 0180-0094			CAPACITOR-FXD 1000F+75-102 25VDC AL CAPACITOR-FXD 50UF+100-10X 25VDC AL CAPACITOR-FXD 50UF+100-10X 25VDC AL SPDL CAPACITOR-FXD 50UF+100-102 25VDC AL SPDL	28480 28480 56289 28480 28480	0150~0096 0150-0096 3001076025002 0180-0105 0180-0105
0149-0176	4 1 4 9	3 1 2	CAPACITOR-FXD 10F+75-10% 25VDC AL CAPACITOR-FXD 10F+75-10% 25VDC AL CAPACITOR-FXD 3400FF >-5% 500VDC MICA CAPACITOR-FXD 1000F+75-10% 25VDC AL CAPACITOR-FXD 1000F +-2% 300VDC MICA	56287 56287 72136 56289 72136	30D103G0258A2 30D105G025BA2 DM20F342J05004UJCR 30D107G025DD2 DM13F101G0300WJCR
0160-0168 0140-0176 0160-0168 0150-0396 0180-0117	1 9 1 3 2	2	CAPACITOR-FXD .1UF +-10% 200VDC POLYE CAPACITOR-FXD 100PF +-2% 300VDC MICA CAPACITOR-FXD .1UF +-10% 200VDC POLYE CAPACITOR-FXD .05UF +80-20% 100VDC CER CAPACITOR-FXD 2.7UF+-10% 35VDC TA	29490 72136 28480 28480 36289	0160-0168 DH15F10160300WV1CR D160-0168 0150-0096 1500275X903582
8180-0059 0180-1819	1 3	1 1	CAPACITOR-FXD 10UF+75-10% 25VDC AL CAPACITOR-FXD 100UF+75-10% 50VDC AL NDT ASSTRUED	36289 56289	30D106G025B02 30D107G050DH2
0150-0043 0150-0023	2 6	1	CAPACITOR-FXD 8.2PF +-5X 500VDC TI DIOX CAPACITOR-FXD 2000PF +-20X 1KVDC CER	28480 72982	0150-0045 808-45U0-202M1KV
0150-6023 0150-0023 0150-0023 0150-0023 0180-0119 0140-0069	66647		CAPACITOR-FXD 2000PF +-20% 1KVDC CER CAPACITOR-FXD 2000PF +-20% 1KVDC CER CAPACITOR-FXD 2000PF +-20% 1KVDC CER CAPACITOR-FXD 1UF+75-10% 2SVDC AL CAPACITOR-FXD 1UF+75-10% 2S00V MICA	72982 72982 72982 56289 72982	808-45U0-20281KV 808-45U0-20281KV 808-45U0-20281KV 30D105C025BA2 666-034-01A4-551K
0150-0089	4	2	CAPACITOR-FXD 4.7PF +-,25PF 500VDC CER	28480	0150-0089
0150~0089	4		CAPACITOR-FXD 4.7PF +- 25PF 500VDC CER	28485	0150-0089
0160-2295	7	1	(MDDEL 608F ONLY) Capacitor-Fdthru 100pf 20% 300v cer (Model 608F Only)	28493	0160-2293
0180-0089 0160-2234	7	1 2	(MODEL 600F ONLY) CAPACITOR-FXD 10UF+50-102 150VDC AL CAPACITOR-FXD .51PF +25PF 500VDC CER (MODEL 600F ONLY)VALUE SELECTED IN TEST	56299 28480	30D106F150DD2 0160-2234
0160-2235	7	1	(MODEL & GARF.DNLY)VALUE SELECTED IN TEST CAPACITOR-FXD .75PF +25PF 500VDC CER	28490	0160-2235
0160-2234	6		CAPACITOR-FXD .51PF +~.25PF SUDVDC CER (MODEL 608F ONLY)VALUE SELECTED IN TEST	29460	0160-2234
1901-0518 1900-0017	6 0	1	DIDDE-SM SIG SCHOTTKY DIDDE-MICROWAVE 1N21B	28 480 1 <i>7</i> 54 0	1901-0518 1n218
1910-0002	4	1	NOT ABBIGNED DIDDE-GE INJBB 100V SOMA DO-7	08257	1N309
1701-0029 1901-0029 1901-0029	6 6	4	DIODE-PWR RECT 600V 750MA DD-29 DIODE-PWR RECT 600V 750MA DO-29 DIODE-PWR RECT 600V 750MA DO-29 DIODE-PWR RECT 600V 750MA DO-29 DIODE-PWR RECT 400V 750MA DO-29	28480 28480 28480 28480 28480 28480	1901-0029 1901-0029 1901-0029 1901-0029 1901-0029 1901-0028
1901-0028 1901-0028 1901-0028 1901-0028 1901-0026 1901-0026	22223	6	DIODE-PWR REGT 400V 750MA DO-29 DIODE-PWR REGT 400V 750MA DO-29 DIODE-PWR REGT 400V 750MA DO-29 DIODE-PWR REGT 200V 750MA DO-29 DIODE-PWR REGT 200V 750MA DO-29	28480 28490 28490 28480 28480 28480	1901-0028 1901-0028 1901-0028 1901-0026 1901-0026
1901-0026 1901-0026 1982-3129 1901-0025 1901-0038	33823	1 5 1	DIODE-PWR RECT 2007 750MA DO-29 DIODE-PWR RECT 2007 750MA DO-29 DIODE-ZNR 7.50 2% DO-33 PD=.4W TC=+.03% DIODE-GEN PRP 1007 200MA DO-7 DIODE-SWITCHING 007 200MA 2NS DO-35	28490 28480 28480 29480 28490	1901-0026 1901-0026 1902-3129 1901-0023 1901-0023
1901-0025	2		DIDDE-GEN PRP 100V 200MA DO-7	28480	1901-0025
1901-0025 1902-0227 1901-0025	2 8 2	1	NOT ASBIENED Diode-gen PRP 100V 200MA DO-7 Diode-znr 22.6V 3X PD=1.5W TC=+.09X Diode-gen PRP 100V 200MA DO-7	28490 28480 28480	1901-0025 1902-0227 1901-0025
1902-0049 1901-0026 1901-0026 0122-0031	2337	1	DIQDE-ZNR 6.19V 5% DQ-35 PD=.4W DIQDE-PWR RECT 2000 750MA DD-29 DIQDE-PWR RECT 2000 750MA DQ-29 DIQDE-PWR RECT 2000 750MA DQ-29 DIQDE-VUC 15PF 20% C.1/C4-HIN=2.3 DQ-14 (MQDEL 600F ONLY)	28480 28490 26480 12498	1902-0049 1901-0026 1901-0026 VA126
	0140-0146 0140-0176 0140-0176 0140-0176 0160-0148 0140-0176 0160-0148 0150-0059 0160-017 8100-0059 0160-1819 0150-0023 0150-0023 0150-0023 0150-0023 0150-0023 0150-0023 0150-0023 0150-0023 0150-0023 0150-0023 0150-0089 0150-0089 0150-0089 0150-2234 0160-2235 0160-2235 0160-2234 0160-2235 0160-2234 1901-0518 1901-0518 1901-0029 1901-0029 1901-0028 1901-0028 1901-0028 1901-0028 1901-0028 1901-0028 1901-0028 1901-0028 1901-0028 1901-0028 1901-0028 1901-0025 1901-0025 1901-0025 1901-0025 1901-0025 1901-0025 1901-0025 1901-0025	0140-0146 1 0140-0176 9 0140-0176 9 0140-0176 9 0140-0176 9 0150-0168 1 0150-0176 9 0150-0176 9 0150-0176 9 0150-0176 9 0150-0076 3 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0023 6 0150-0025 7 0160-2235 7 0160-2235 7 0160-2234 6 1901-025 5 1901-027 6 1901-028 5 1901-028 5 1901-028 3 1901-02	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	0:14-01164 1 CAPACITOR-FXD 3000F35 3000F35 3000C MICA 0:140-0174 9 2 CAPACITOR-FXD 100PF +-10X 2000C AL 0:140-0176 9 2 CAPACITOR-FXD 100PF +-2X 300VDC MICA 0:140-0176 9 2 CAPACITOR-FXD 100PF +-2X 300VDC MICA 0:140-0176 9 2 CAPACITOR-FXD 100PF +-2X 300VDC MICA 0:160-0168 1 CAPACITOR-FXD 100FF +-2X 300VDC TA 0:160-0169 1 CAPACITOR-FXD 30UF +0-2X 100VDC CER 0:160-0169 1 CAPACITOR-FXD 30UF +7-2X 10VDC CER 0:160-0163 1 CAPACITOR-FXD 300FF +-2X 10VDC CER 0:150-0023 6 CAPACITOR-FXD 300FF +-3X 10VDC CER 0:150-0023 6 CAPACITOR-FXD 300FF +-3X 10VDC CER 0:150-0023 6 CAPACITOR-FXD 300FF +-3X 10VDC CER 0:150-0023 6 CAPACITOR-FXD 10FFXD 10FF 10X 2500F 10X 100 CER 0:150-0023 6 CAPACITOR-FXD 10FFXD 10FF 10X 2500VD CER 0:150-0023 6 CAPACITOR-FXD 10FF 10X 2500VD CER 0:150-0024 1 CAPACITOR-FXD 10FF 10X 2500VD CER 0:150-0089	1:40-01:44 1 CAPACITOR-FX0 3400F -22 35000 72:136 0:80-0094 4 CAPACITOR-FX0 100FF 72:136 52:29 0:40-0176 9 2 CAPACITOR-FX0 100FF 72:136 72:136 0:40-0176 9 2 CAPACITOR-FX0 100FF 72:136 72:136 0:40-0176 7 CAPACITOR-FX0 100FF -22 3000C MICA 72:136 0:40-0176 7 1 CAPACITOR-FX0 10FF -22 3000C MICA 72:136 0:40-0176 7 1 CAPACITOR-FX0 10FF 72:02 2000C MICA 52:299 0:80-0157 1 CAPACITOR-FX0 10FF 72:03 2000C AL 52:299 0:80-0158 1 CAPACITOR-FX0 10FF 72:03 2000C AL 52:299 0:80-0153 1 CAPACITOR-FX0 20:00F +-203 1WUDC CER 72:982 0:50-0123 6 CAPACITOR-FX0 20:00F +-203 1WUDC CER 72:982 0:50-0123 6 CAPACITOR-FX0 10:075-103 20:00 C CER 72:982 0:50-0124 CAPACITOR-FX0 10:075-103 20:00 C CER 72:982 0:5

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Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
CR32	0122-0031	7		DIDDE-VVC 15PF 202 C.1/C4-HIN=2.3 D0-14	12498	VA126
CR33	1901-0025	2		(MODEL 608F DNLY) Didde-gen frp 100v 200ma DO-7	28480	1901-0025
D51	2140-0007	7	۱	LANP-INCAND 12VDC 8MA S-6-BULB Thermal resistor rating of Lamp 12V 8MA.	08806	6A/S6-12V
250	1450-0013	1		(LAHP MAY ALSO BE MARKED 8A/SS 12V) Lampholber Scr-b-Cand-Skt Sldr-Lug-term NOT Assigned	28486	1450-0013
D83	2140-0009	9	2	LAMP-INCAND 47 6.3VDC 150MA T-3-1/4-BULB (DIAL LIGHT)	1F556	47
DS4	1430-0009 2140-0009	5 9	1	LAMP SOCKET MINTR-BAY-SKT SLDR-LUG-TERM Lamp-Incand 47 6.3VDC 150MA T-3-1/4-Bulb	29490 1 F356	1450-0009 47
	\$450~0009	5		(POWER LIGHT) LAMP SOCKET MINTR-BAY-SKT SLDR-LUG-TERM	29480	1450-0009
F1	2110-0014	3	1	FUSE 44 250V TD 1.25X.25 UL	75915	313504
F1	2110-0006	з	1	(FOR 1159 OPERATION) FUSE 2A 2509 TD 1.252.25 UL (FOR 2309 OPERATION)	71400	MDX-2
F2				NOT ABSIGNED		
F3	2110-0004	1	1	FUSE ,25A 250V NTD 1.25X.25 UL	28490	2110-0004
FL1 FL2 FL3	00609-624 6084-27D	9	1 4	FILTER-VARICAF (MODEL 608F OHLY) Filter-d Green NDT Assigned	28480 28486	00608-624 6088-27d
FL4 FLS	606A-27A 608A-27D	3 9	2	FILTER-A RED Filter-d green	28481 28486	608A-27A 608A-27D
FL6 FL7	608D-27C	3	د د	FILTER-RF (MOD, BLUE) Filter-a Red	28480 28480	608D-27C 608A-27A
FLO	609D-27B	1	1	FILTER-RF (WHITE)	28460	609D-27B
FL9 FL10	6080-69M 608A-27D	8 9	1	CHOKE-RF Filter-d green	28480 28480	608d-60m 60 Ba-27d
FL11	608A-27D	9		FILTER-D GREEN	28480	6084-270
15 11	1250-0001	3	3	NOT ASSIGNED Connector-RF bnc fem SGL-Hole-FR 30-ohm	28480	1250-0001
12	1251-0071	9	1	(AK/PULSE KOD, INPUT) CDNNECTOR-TEL JACK 2-CKT ,25-SHK-DIA (XTAL CAL. OUTPUT)	29480	1251-0071
J4	1250-0099	9	1	CONNECTOR-RF N FEK	28460	1250-0099
	5021-0810	1	1	(RE OUTPUT) CONNECTOR-RECEPTACIE JACK CO-AXIAL TYPE	28480	3021-0810
32	1250-0053	5	1	NSR - PART OF US Cap-coax to fit f-bnc non-shtg 2.5-ch	28480	1230-0053
J 6	1250-0001	3		CONNECTOR-RF BNC FEM SGL-HOLE-FR 50-0HM	28480	1250-0001
37	1250-0001	3		(MODEL 608F ONLY) FREQ. CONTROL INPUT Connector-RF BNC FEH SGL-HOLE-FR 50-0HM	26460	1250-0001
	00408-403	1	1	(MODEL 608F ONLY) FREQ. ANALOG OUTPUT Turret dscillator Assembly (Model 608e only)	28480	00608-603
	00608-623	5	1	OSCILLATOR TURRET ASSEMBLY (Nodel 600F only)	29490	10609-623
L1 L2 L3 L4 L5				NSR - PART OF DSC. TURRET ASSEMBLY NSR - PART OF OSC. TURRET ASSEMBLY NSR - PART OF OSC. TURRET ASSEMBLY NSR - PART OF DSC. TURRET ASSEMBLY NSR - PART OF OSC. TURRET ASSEMBLY		
L6 L7	9140-0808 9140-0035	5 8	1 1	COIL-VAR 10.5UH-20UH Coil-Var 10.5UH-20UH	28480 28490	9143-0008 9140-0035
L8- L10				NDT ARSIGNED		
L11	608D-60K 608C-60A DD60B-626	4 8 8	1 1 3	CHOKE-RF Turret Abbenbly-Amp (Model 608e only) Amplifier Turret Assy (Model 608f only)	28480 29490 28480	408D-40K 608C-40A 00608-426
L12-						
L17 L18	9140~0036	9		NSR - PART OF AMPLIFIER TURRET ASSEMBLY Coil-Var 4500H-500UH	28484	9140-0036
L19 L20	9140-0096	1	1	NOT ASSIGNED Inductor RF-CH-HLD 10H 10% .166DX,305LG	28480	9140-0096
L21 L22	9140-0026	7	1	NOT ASSIGNED Inductor 6,8uh 10% ,188DX,688LG Q≈60	28480	7140-0026
H1	* 1120-0396	4	1	HETER (Hodel 600e only) rf dutfut	29481	1120-0396
M1	1120-1503	9	1	METER; SOUA	28480	1120-1505
N1 K2	1120-0397	5	1	(MODEL 608F ONLY) RF QUTPUT Meter Percent Modulation	29480	1120-0397

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Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
P1	0120-0015	1	1	CABLE ASSY 18AWG 3-CNDCT BLK-JKT	26490	8120-0015
91 92 93 94 95	1850~0087 1854-0071 1854-0071 1854-0071 1854-0071 1854-0003	7 7 7 7 7 7	132	TRANSISTOR PNP 2N1544 GE TO-3 PD=1064 Transistor NPN SI PD=300m4 FT=200MHZ Transistor NPN SI PD=300N4 FT=200MHZ Transistor NPN SI PD=300N4 FT=200MHZ Transistor NPN SI PD=300N4 FT=200MHZ	28480 29480 28480 28480 28480 04713	1550-0067 1854-0071 1854-0071 1854-0071 28708
96 97 98	1854-0005 1853-0012 1205-0011 1854-0003 1205-0011	7 4 0 5 0	1 1 3	TRANSISTOR NPN 2N208 SI TO-18 PD=360NW TRANSISTOR PNP 2N2904A SI TO-39 PD=600NW HEAT SINK TO-5/TO-39-CS TRANSISTOR NPN 8I TO-39 PD=800MW HEAT SINK TO-5/TO-39-C8	04713 01295 28480 28480 28480 28480	2N708 2N2904A 1205-0011 1854-0003 1205-0011
49 410 G11	1295-0011 1854-0003 1854-0003 1854-003 1854-0039 1295-0011	0 5 5 7 0	1	HEAT SINK TO-5/TO-39-CS TRANSISTOR NPN SI TO-39 PD=800MW TRANSISTOR NPN SI TO-39 PD=800MW TRANSISTOR NPN 2N30538 SI TO-39 PD≠1W HEAT SINK TO-5/TO-39-C5	28489 28480 28480 01928 29480	1205-0011 1854-0003 1854-0003 2N30538 1205-0011
1 - R2 R3 R4 R5	0727-0184 0730-0058 0727-0184	3 7 3	2	NOT ASSIGNED Resistor 20.4K 1% 10 CF TC≠0-500 Resistor 75K 12.10 CF TC≠0-500 Resistor 20.4K 1% 10 CF TC≠0-500	91637 91637 91637	MCS1/2-2842-F DC1-7332-F MCS1/2-2842-F
R6 R7 R9 R10	0730-0050 2100-0036 0690-1221 0690-1541 0690-1521	7 8 2 9 5	1 1 4 2	RESISTOR 75K 12 1W CF TC=0-300 RESISTOR-VAR CONTROL CCP 1K 20% LIN RESISTOR 1.2K 10% 1W CC TC=0+647 RESISTOR 150K 10% 1W CC TC=0+802 RESISTOR 1.5K 10% 1W CC TC=0+647	91637 28480 01121 01121 01121	DC1-7502-F 2100-0036 GB1221 GB1541 GB1521
R11 R12 R13	0690-1051 0690-1541	6 9	6	REGISTOR IM 10% 1W CC TC=0+1000 Registor 150k 10% 1W CC TC=0+882 Nov Assigned	01121 01121	GB1 051 GB1 541
R-14 R 15-	0693-1231	0	ı	RESISTOR 12K 10% 24 CC TC=0+765	01121	HB1231
832 833 834-	0819-0063	٥	1	NOT ASSIGNED Resistor 5% 5% 20% pw TC=0+-260	91637	HL20-02Z-20W-5001-J
R42 R43 R44	0816-0006	7	1	NOT AGSIGNED Resistor 5% 5% 100 PW TC=0+-260	91637	HL12-027-100-5001-J
R 47 R 48 R 49 R 50 R 51 R 51 R 52	0687-2741 0687-5621 0690-1031 0687-1041 0687-1241	67276	1 1 1 2	NOT ASSIGNED RESISTOR 270K 10% .5W CC TC=0+682 RESISTOR 5.6K 10% .5W CC TC=0+647 RESISTOR 10K 10% .5W CC TC=0+765 RESISTOR 100K 10% .5W CC TC=0+882 RESISTOR 120K 10% 10W CC TC=0+882	01121 01121 01121 01121 01121 01121	E\$2741 E\$5621 G\$1031 E\$1041 G\$1241
R53 R54 R55 R56 R37	0690-1051 0690-1241 0693-2731 2100-0159 0693-4711	6 6 7 6 7	1 1	RESISTOR 1M 10% 1W CC TC=0+1000 ° RESISTOR 120K 10% 1W CC TC=0+802 RESISTOR 27K 10% 2W CC TC=0+765 RESISTOR-VAR CONTROL CCP 1M 20% 10CW REBISTOR 470 10% 2W CC TC=0+529	01121 01121 01121 28480 01121	GB1051 GB1241 HB2731 2100-0159 HB4711
R38 R59 R60* R61~	0721-0006 0721-0006 0684-1211	6 7	2 1	RESISTDR 53.3 1% ,1₩ CF TC≈0-500 REGISTOR 53.3 1% .1₩ CF TC≈0-500 RESISTOR 120 10% .25₩ FC TC=-400/+600	91637 91637 01121	DC1/10-53R3-F DC1/10-53R3-F CB1211
ጽሪዓ ጽሪ5 ጽሪ5 ጽሪ7 ጽሪ8 ጽሪ9	0690-1231 0690-1 341 0690-1851 0690-1821 0690-1021 0690-1031	4 7 4 0 5	i 1 3	NOT ABSIGNED RESISTOR 12K 10% 1W CC TC=0+765 RESISTOR 150K 10% 1W CC TC=0+862 RESISTOR 1.08H 10% 1W CC TC=0+1000 RESISTOR 1K 10% 1W CC TC=0+647 RESISTOR 1H 10% 1W CC TC=0+1000	01121 01121 01121 01121 01121 01121	GB1231 GB1541 GB1851 GB1021 GB1021 GB1051
R70 R71 R72 R73 R74	0690-1541 2140-0157 0690-2241 0690-2731 0693-6831	9 4 8 1 6	2 3 1 1	RESIBTOR 150K 10% 1W CC TC=0+802 RESISTOR-VAR CONTROL CCP 50K 20% LIN REBISTOR 220K 10% 1W CC TC=0+892 RESISTOR 27K 10% 1W CC TC=0+765 REBISTOR 66K 10% 2W CC TC=0+765	01121 29480 01121 01121 01121	CB1541 2100-0157 GB2241 GB2731 HB6631
R73 R76 R77 R78 R79	0690-3331 0690-1051 0690-1021 0690-1051 0690-2241	7 6 0 6 8	1	RESISTOR 33K 10% 1₩ CC TC=0+765 REBISTOR 1H 10% 1₩ CC TC=0+1000 RESIGTOR 1K 10% 1₩ CC TC=0+647 REGISTOR 1H 10% 1₩ CC TC=0+882 REBISTOR 220K 10% 1₩ CC TC=0+882	01121 01121 01121 01121 01121 01121	GB3331 GB1051 GB1021 GB1051 GB2241
880 881 852-	2100-0157 0690-2241	4		RESISTOR-VAR CONTROL CCP SOK 202 LIN RESISTOR 220K 10% 1W CC TC=0+882	28480 01121	2100-0157 GB2241
R89 R89	0687-4741	0	2	NOT ASSIGNED Resistor 470k 10% .50 CC TC=0+892	01121	EB4741
R89 R90 R91 R92 R93	0687-5601 0690-8201 0690-8201 0693-2231 0690-8231	8 2 2 2 2 2	2 2 4 2	REBISTOR 56 10% .5% CC TC=0+412 RESIGTOR 82 10% 1W CC TC=0+412 RESISTOR 82 10% 1W CC TC=0+412 RESISTOR 22% 10% 2W CC TC=0+765 RESISTOR 82% 10% 1W CC TC=0+765	D1121 01121 01121 01121 01121 01121	E85601 G98201 G88201 HB2231 G88231

See introduction to this section for ordering information *Indicates factory selected value

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Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
R74 R95 R94 R97 R98	0487-4741 0487-5601 0490-1511 0493-2231 0690-8231	0 3 2 2 8	1	RESISTOR 470X 10% .5W CC TC=0+B82 RESISTOR 36 10% .5W CC TC=0+412 RESISTOR 150 10% IW CC TC=0+529 RESISTOR 22% 10% 2W CC TC=0+765 RESISTOR 82% 10% IW CC TC=0+765	01121 01121 01121 01121 01121 01121 01121	EB4741 E95601 GB1511 HB2231 GB0231
R99 R100 R101 R102 R103	0490-1821 0490-1041 0690-1051 0493-2231	9 8 6 2	1 1	RESISTOR 1.8K 10% 14 CC TC=0+647 RESISTOR 10M 10% 14 CC TC=0+1059 RESISTOR 1M 10% 14 CC TC=0+1000 RESISTOR 22K 10% 24 CC TC=0+765 NOT ASSIGNED	D1121 01121 01121 01121 01121	GB1821 GB1061 GB1951 HB2231
R104 R105 R106 R107 R108 R108 R113	2100-0156 0690-2211 2100-0156 0693-2231	NUNG	2 1	RESISTOR-VAR CONTROL CCP 10K 20% LIN RESISTOR-220 10% 10 CC TC=0+529 RESISTOR-VAR CONTROL CCP 10K 20% LIN RESISTOR 22K 10% 20 CC TC=0+765 NOT ASSIGNED	20480 01121 20400 01121	2100-0136 GB2211 2100-0136 HB2231
R114 R115 R115-	0690-5631	6		NDR - FART OF FL9 Resistor 56K 10% 16 CC TC=0+765	01121	695631
R117 R118	0690-1021	0		NDT ABBIGNED Resistor 1K 10% 1W CC TC=0+647	01121	GB1021
119 120-	0690-1321	5		RESISTOR 1.5K 10% 14 CC TC=0+647	D3121	GB1521
8121 8121 8122 8123	0687-1021 0687-2731	3	1 1	NOT ASSIGNED Rebistor 1% 10% .54 CC tC=0+647 Rebistor 27% 10% .54 CC tC≖0+765	01121 01121	EB1021 EB2731
R124 R125	0818-0027 0819-0022	6 3	1	REGISTOR 60 5% 400 PM TC=0+-260 RESISTOR 50 5% 200 PM TC=0+-260	91637 35434	HL35-08Z-40W-60R-J CA20
₹126- ₹128 ₹129	0813-0029	8	1	NOT AGSIGNED Resistor 1 3% 3W PW TC=0+-50	28460	0813-0029
R130- R179 R200 R201 R202	07 37 -0429 0757-0439 2100-0092	24.6	2 1 1	NOT ASSIGNED Resistor 1.82k 1x .125% F TC=0+-100 Resistor 6.81k 1x .125% F TC=0+-100 Resistor-Var Control CP 10k 20x Lin	24546 24546 28480	C4-1/8-T0-1821-F C4-1/8-T0-4811~F 2100-0092
R 203 R 204 R 205 R 205 R 206 R 207	0757-0854 2100-0151 0757-0442 0757-0442 0757-0442 0757-0280	79993	2 1 4 3	RESISTOR 56.2K 1% .5W F TC=0+-100 RESIBTOR-VAR CONTROL CP 500 20% LIN RESIBTOR 10K 1% .125W F TC=0+-100 RESISTOP 1K '% 125W F TC=0+-100 RESISTOP 1K '% 125W F TC=0+-100	28490 26460 24546 24546 24546 24546	0757-0834 2100-0151 C4-1/8-T0-1002-F C4-1/8-T0-1002-F C4-1/8-T0-1001-F
R208 R209 R210 R211 R212	0737-0465 2100-0281 2100-0235 0757-0477 2100-1619	6 5905	2 1 1 1 2	RESISTOR 100K 12 .125W F TC=0+-100 RESISTOR-TRMR 100 202 WW TOP-ADJ 1-TRN RESISTOR-VAR CONTROL CCP 5K 202 LIN RESISTOR 332K 12 .125W F TC=0+-100 RESISTOR 32K 12 .125W F TC=0+-100 RESISTOR-VAR DUAL 5K-202-CC 50K-202-CC	24548 28480 28480 19701 28480	C4-1/8-TD-1003-F 2100-0281 2100-0235 KF4C1/8-T0-3323-F 2100-1619
R213 R214	0757-0464 0757-0407	5	1 2	RESISTOR 90.9K 1% .125W F TC=0+-100 REGISTOR 200 1% .125W F TC=0+-100 (Hodel 608e only)	24546 24546	C4-1/8-TC-9092-F C4-1/8-T0-201-F
R214	0757-0413		1	REBISTOR 392 1% ,125₩ F TC≠0+-100 (MODEL 608F ONLY)	24546	C4-1/8-T0-392R-F
R215 R216 R217 R218 R219	0757-0419 0757-0441 0757-0430 0757-0405 0757-0281	08544	1 1 2 2 2	RESISTOR 681 1% .125₩ F TC≈0+-100 RESISTOR 0.25K 1% .125₩ F TC=0+-100 RESISTOR 2.21K 1% .125₩ F TC=0+-100 RESISTOR 162 1% .125₩ F TC=0+-100 RESISTOR 2.74K 1% .125₩ F TC=0+-100	24546 24546 24546 24546 24546 24546	C4-1/0-T0-&81R-F C4-1/0-T0-8231-F C4-1/0-T0-2211-F C4-1/0-T0-162R-F C4-1/0-T0-162R-F C4-1/8-T0-2741-F
R220 R221 R222 R223 R223 R224	0737-0437 0757-0200 0757-0281 0757-0280 0757-0280 0757-0260	27433	2 1	RESISTOR 4.95K 1% .125W F TC=0+-100 RESISTOR 5.62K 1% .125W F TC=0+-100 RESISTOR 2.74K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100	24546 24546 24546 24546 24546 24546	C4-1/0-T0-4731-F C4-1/0-T0-3621-F C4-1/0-T0-2741-F C4-1/0-T0-101-F C4-1/0-T0-1001-F
R223 R226 R227	0737-0430 2140-1619	5 5		RESIGTOR 2,21K 1% .125W F TC=0+-1D0 Resistor-Var Dual 5x-20%-CC 50K-20%-CC NDT Abbigned	24545 29480	C4-1/8-T0-2211-F 2100-1619
R228 R229	0757-0852 0757-0429	2	1	RESISTOR 47.5K 1% .5W F TC=0+-100 RESISTOR 1.82X 1% .125W F TC=0+-100	28480 24546	0757-0852 C4-1/8-T0-1821-F
R230 R231 R232 R233 R234	0757-0811 0484-1001 0684-1001 0757-0444	6331	1 2 1	RESISTOR 392 1% ,5₩ F TC=0+-100 RESISTOR 10 10% ,25₩ FC TC=~400/+500 RESISTOR 10 10% ,25₩ FC TC=-400/+500 RESISTOR 12.1% ;% ,125₩ F TC≈0+-100 NOT ASSIGNED	29480 01121 01121 24346	0757-0811 CB1001 CB1001 C4-1/8-T0-1212-F
R235 R236 R237# R238 R238 R239	0687-6841 0757-0434 0698-4477 0757-0290 0757-0854	59257	1 1 1 1	RESISTOR 680K 10% .5W CC TC=0+882 RESISTOR 3.65K 1% .125W F TC=0+-100 RESISTOR 10.5K 1% .125W F TC=0+-100 RESISTOR 6.19K 1% .125W F TC=0+-100 RESISTOR 56.2K 1% .5W F TC=0+-100	01121 24546 24546 19701 28480	EB6841 C4-1/8-T6-3651-F C4-1/8-T0-1052-F HF4C1/8-T0-6191-F 0957-0854

See introduction to this section for ordering information *Indicates factory selected value

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
R240	0757-0346	s	1	RESISTOR 10 12 .1254 F TC=0+-100	24546	C4-1/8-T0-1080-F
R241 8242	0757-0410	1	2	RESISTOR 301 12 .125W F TC=0+-100	24546	C4-1/8-T0-301R-F
R243	0684-3901 0757-0411	6 2	1	RESISTOR 39 10% .25₩ FC TC=-400/+500 RESISTOR 332 1% .125₩ F TC=0+-100	01121 24546	CB3901 C4-1/B-TD-332R-F
-R244	0698-3430	5	i	REGISTOR 21.5 12 .125W F TC=0+-100	03868	PHE55-1/0-10-21R5-F
R245 R246	0684-1311 0757-0197	0 1	1 2	REGISTOR 150 10% .25₩ FC TL⇒-400/+600 RESISTOR 1,5K 1% .5₩ F TC=0+-100	01121 28480	CB1511 0757-0197
R247	0757-0197	1	- {	RESISTOR 1.5K 1% ,5W F TC=0+~100	29400	0757-0197
R240 R249	0757-0841 0757-0801	2	1	RESISTOR 12.1K 12 .5W F TC=0+-100 RESISTOR 150 12 .5W F TC=0+-100	28486 28480	0757-0841 0757-0801
R250	0757-0276	,	1	RESISTOR 61.9 1% .125W F TC=0+-100	24546	C4-1/8-T0-6192-F
R251	0757-0410	1		REBISTOR 301 1% .125W F TC=0+-100 (MODEL 608E ONLY)	24546	C4-1/8-T0-3018-F
R251	0757-0407	^		RESISTOR 200 1% .125W F TC=0+-100 (Model 508F only)	24546	C4-1/8-T0-201-F
R232	0757~0394	0	1	RESISTOR 51.1 1% .125W F TC=0+-100	24546	C4-1/8-T0-31R1-F
R253 R254	0761-0021	2	1	RESISTOR 1K 5% 1W MD TC=0+-200 Resistor 750 5% 1W MD TC=0+-200	28480 28480	0761-0021 0761-0038
R255	0816-0032	9	i	RESISTOR 4K 5% 100 PW TC=0+-260	91637	HL12-02Z-104-4001-J
R256	0812-0050	3	1	REBISTOR 3K 5% 5W PW TC=0+~20	28480	0B12-0050
R257 R258	0727-0002 0690-1841	4	1	RE91STOR 3 1% .54 CF TC=0-500 Resistor 100K 10% 14 CC TC=0+892	28480 01121	0727-0002 681841
R259	4912.447	_		NOT ASSIGNED	10.00	A612-0078
R260 R261	0812-0038 0757-0457	7	1 2	RESIGTOR 0.66K 37 5W PW TC=0+-20 Resistor 47.5K 17 .125W F TC=0+-10D	28460 24546	0812-0038 24-1/8-10-4752-F
R262	8757-0465	6		RESISTOR 100K 1% .125W & TE=0+-100	24546	C4-1/8-T0-1003-F
R263	0757-0457	6	1	RESISTOR 47,5K 1% ,125% F TC=0+-100	24546	C4-1/8-T0-4752-F
R264	0757-0442	9		RESISTOR 10K 1% ,125W F TC=0+-100 (HODEL 600F ONLY)	24546	C4-1/8-T0-)002-F
R265	0737-0442	9		RESISTOR 10K 1% ,125W F TC=0+-100 (Model 40BF DNLY)	24546	C4-1/8-T0-1002-F
R266	0757-0437	2		RESISTOR 4.75K 12 .125W F TC=0+-100 (Model 600F ONLY)	24546	C4-1/8~T0-4751-F
R267	0698~3415	6	۱	RESISTOR 19.6K 1% ,5W F TC=0-106 (MDDEL 608F ONLY)	28480	0698-3415
R268	0757-0833	2	1	RESISTOR 5.11K 12 .5W F TC=0+-100 (MODEL 609F DNLY)	26480	0757-0833
R269	2100-1750	5	۱	RESISTUR-VAR CONTROL CCP 4K 10% S-LIN (Model 608F DNLY)	28480	2100-1750
R270 R271	0737~0405 0757-0638	4 7	1	RESISTOR 182 12 ,125W F TC=0+-100 RESISTOR 9.09K 12 .5W F TC=0+-100 (MODEL 60BF ONLY)	24546 28480	C4-1/8-10-162R-F 0757-0838
R272	0816-0002	з	1	RESISTOR 3% 5% 104 PW TC=0+-260	91637	HL12-02Z-10W-3001-J
RF1	9170-0029	3	1	CORE-SHIELDING BEAD	28480	9170-0029
51				NOT ABSIGNED		
52	3101-0033	8	1	SWITCH-SL DPDT STD .5A 125VAC/DC	28480	3101-0033
83 84	3101-0003 3101-0012	2	1	SWITCH-TGL BASIC DAST 3A 250VAC/DC SWITCH-TGL BASIC DADT 10A 250VAC	04009 28490	91024-GB 3101-0012
				(XTAL CAL)		
65 56	3102-0006	7	1	NOT ASSIGNED Switch-sens opdt submin 1a Jovdc	28480	3102-0006
S7	0608-60047	ĺ	i	SWITCH-HOD SELECTOR ASSEMBLY (INCLUDES C203 AND R262)	28460	06608-60647
τı	9100-2299	4	1	TRANSFORMER-POWER 115/2304) FREQ NOT	28480	9100-2289
12 T3	9100-0342 9120-0018	6 5	1 1	TRANSFORMER-POWER 115/230V 50-1000HZ TRANSFORMER-AUDIO PRI IMPI 15KOHMS; SEC	28480 28460	9100-0342 9120-0018
٧١				NOT ASBIGNED		-
V2A V2B	1932-0066	6	1	TUBE-ELECTRON 12AU7 TRIODE-DUAL NSR - PART OF V2A	33123 33173	12AU7 12AU7
V3- V5				NOT AGSIGNED		
V6	1921-0001	6		TUBE-ELECTRON 4042 TRIODE	28480	1921-0001
	1200~0010	9	2	SOCKET-TUBE 2-CONT	29489	1200-0010
	6080-59A-4 6080-59A-3	97	1	SOCKET-OSCILLATOR GRID	29490 28490	608D-39A-4 608D-39A-3
	612A-38A	2	2	BOCKET-OSCILLATOR TUBE WRENCH-TUBE	29480	612A-38A
47 V6	1921-0002	2	1	NOT ASSIGNED Tube-Electron 4043 Triode	28459	1921-0002
va	1200-0010	9		SDCKET-TUBE 2~CONT	28480	1200-0010
	609D-S8A 612A-38A	6	۱	CONTACT-AMP GRID	28460 28461	6080-58A 612A-38A
118				TUBE-ELECTRON SUBA TRIDDE-PENTODE	0192B	6UBA
V9 V1 0A	1933-0013 1932-0027	5	2	TUBE-ELECTRON 12AT7 TRIODE-DUAL	24972	12AT7WA
VIOB	1208-0008	5	1	NOT USED Socket-Tube 9-Cant	28480	120 0-00 08
911	1932-0027	9		TUBE-ELECTRON 12AT7 TRIODE-DUAL	24972	12AT7¥A

See introduction to this section for ordering information *Indicates factory selected value

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Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
V12 V13A V13B	1723-0021 1200-0009 1932-0010 1280-0005	4 6 0	2 6 2 1	TUBE-ELECTRON 6AU6A/B425A PENTODE BOCKET-TUBE 7-CONT SLDR-EYE TUBE-ELECTRON 6080 TRIODE-DUAL HGR - PART OF VI3A GOOVERTUBE & CONT OF D. D. EVE	33173 28480 01928	64U6A 1200-0009 6080
V14 V15 V16A V16B	1923-0003 1923-0021 1200-0009 1940-0001 1200-0009 1932-0010 1200-0005	2 4 6 6 6 0 2	1	SOCKET-TUBE 8-CONT SLDR-EYE TUBE-ELECTRON 6AU6A/8425A PENTODE SOCKET-TUBE 7-CONT SLDR-EYE TUBE-ELECTRON 3651A DIODE-V RCLTR SOCKET-TUBE 7-CONT SLDR-EYE TUBE-ELECTRON 6980 TRIDDE-DUAL NSR - PART OF VI6A SOCKET-TUBE 9-CONT SLDR-EYE	28480 33173 28480 01929 28480 01928 28480 28480	1200-0005 6AU6A 1200-0007 5651A 1200-0007 6086
417 V18 V19	1923-0017 1200-0009 1923-0017 1200-0009	E 6 6 6	2	NOT ASSIGNED TUBE-ELECTRON 6AH6 PENTODE Socket-Tube 7-Cont SLDR-Eye TUDE-ELECTRON 6AH6 PENTODE Socket-Tube 7-Cont SLDR-Eye	33173 28480 33173 28480	1200-0005 6AH6 1260-0009 6AH6 1200-6D09
V20 V21A V218	1730-0013 1200-0007 1932-0029 1200-0008	9 6 1 5	1 1	TUBE-ELECTRON 6ALS DIDDE-DUAL Socket-Tube 7-Cont 3LDR-Eye Tube-Electron 12AU7 TRIDDE-DUAL NSR - Part of V21A Socket-Tube 9-Cont	33173 28480 003J 28480	6AL5 1280-0009 12AV7 1200-0008
V22	1921~0036 1200-0019	79	1 1	TUBE-ELECTRON ECOSIO TRIDDE SOCKET-TUBE 9-CONT	28480 28480	1921-8036 1200-0019
41 XF1	04608-60037 1400-0090 2110-0470 2110-0465 2110-0467	9 9 5 8 0	1 1 1	CABLE-UNCAL OUTPUT ASSEMBLY FUGEHOLDER COMPONENT FOR USE ON FUGEHOLDER BODY EXTR PST; BAYONET; TND FUGEHOLDER CAP EXTR PST; BAYONET; 20A FUGEHOLDER COMPONENT HEX NUT: 1/2-20 NOTE	28480 28480 75915 28480 28480	00608-60037 1400-0090 345003-010 2110-0465 2110-0467
				IF ANY PART OF THE OLD FUSEHOLDER (1400-0004)NEEDS REPLACING, ALL FOUR PARTS OF THE NEW FUSEHOLDER MUST BE ORDERED. THE OLD FUSEHOLDER CAN BE IDENTIFIED BY A STRAIGHT SOLDER LUG TO WHICH THE WHITE-BLACK-GRAY WIRE ATTACHES.		
VI/ #	1000 0000			ON THE NEW FUSEHOLDER THE SOLDER LUG IS At a Right angle to the body.		
XY1 Y1	1200-0028	9	1	SOCKET-XIAL 2-CONT Crystal-quartz 5.000 MHz	28480 28480	1200-0028 8410~0009
				MISCELLANEOUS PARTS		
	6090-40A 00609-610 00608-60044 00608-60043 00608-60045	4 0 9 0	1 2 1 1	ATTENUATOR DIAL ASSEMBLY ATTENUATOR PLUNGER ASSEMBLY BOARD-RESISTOR ASSEMBLY BOARD-RESISTOR ASSEMBLY BOARD-RECTIFIER ASSEMBLY	28481 28480 28480 28480 28480	&08C-40A 00608-610 00608-60044 00608-60045 00608-60045
	00608-60041 00609-604 00509-613 00608-614 00608-614	52349	1 1 1 1	ATTENUATOR TUBE ASSEMBLY ASSEMBLY-RESISTOR RECTIFIER BOARD BOARD-RESISTOR ASSEMBLY BOARD-RESISTOR ASSEMBLY ALC BOARD ASSEMBLY (MODEL 608E ONLY)	28480 28480 29480 28480 28480 28480	0C608-60041 00608-604 0C608-613 0C608-614 DC608-614 DC608-601
	00608-00024 00608-623 00608-623 00608-602 00608-622	877 704	1 2 1 1	WASHER-GROUNDING BOARD-VARICAP ASSEMBLY BDARD-VARICAP ASSEMBLY (MODEL 600F ONLY) BUFFER ASSEMBLY (MODEL 600F ONLY) BUFFER ASSEMBLY (MODEL 600F ONLY)	28490 28480 28480 28480 28480 28480	00608-00024 00608-625 00608-625 00608-625 00608-622 00608-622
	508D-448 508D-44R 508D-15J 608D-15R 05508-505	5 5 5 7 7 7 7	1 1 1 1	CABINET ASSEMBLY CABINET ASSEMBLY-RACK CABLE-CRYSTAL CALIBRATOR CABLE-CRYSTAL CALIBRATOR ADJUST CABLE-POWER HARNESS	28480 28480 28480 28480 28480 28480	609D-445 609D-448 609D-16J 609D-16R 00608-606
	00508-607 00608-611 00608-612 00409-60037 508D-400	51290	1 1 2 1	CABLE-PANEL HARNESS CABLE-LEVELLER INPUT CABLE-LEVELLER INPUT CABLE-LEVELLER OUTPUT ASSEMBLY DIAL-FREQUENCY VERNIER	28480 28480 28480 28480 28480 28480	00608-607 00608-611 00608-612 00608-612 50608-60037 508D-40C
	00608-600 60668-620 6370-0025 0370-0026	60 /2 ≪ ID	1 1 1	GENERATOR ASSEMBLY (MODEL 60BE ONLY) GENERATOR ABSEMBLY (MODEL 60BF ONLY) KNOB RNDIBLX:FOR .250 SHFT:.750D (XTAL CAL GAIN) KNOB RND:BLK:FOR .250 SHFT:1 ARC:.750D	28490 28490 28480 28480	00609-600 00608-620 0370-0025 0370-0026

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
	0370-0028 0370-0029	7	1	(FINE FREQUENCY) KNOB RND;BLK:FOR ,230 SHFT:1.00D (MOD METER AND RF OUYPUT) KNOB RND:BLK:FOR ,250 SHFT:1 AR0:1.00D	28480 28480	0370-0028 0370-0029
	0370~0035	6	1	(AMPLIFIER TRIMMER) KNOB &KRTD:BAR:BLK:FOR .250SHT:1.000D (NDD SWITCH)	28490	0370~0035
	0370-0049	2	1	KNOB SKRTDIBAR:BLK:FOR .3755HT:1,000D (FREQUENCY RANGE) KNOB RNDIBLK:.375D:CRANK SPINNER	28480 28480	0370-004 5 0370-0050
	608D-8300 612A-38A	7 2	1	WINDOW-FREQ. DIAL Wrench-Tube	28480 28488	600D-83C0 612A-30A
						-

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Mfr Code	Manufacturer Name	Address	Zip Code
	Manufacturer Name	Address	Zip Code
		X	

Table 6-4. Code List of Manufactures

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SECTION VII SCHEMATIC DIAGRAMS COMPONENT LOCATION INFORMATION

7-1. INTRODUCTION.

7-2. Schematic presentations in this manual show electrical circuit operation and are not intended to serve as wiring diagrams. Figure 7-1 shows an example of the schematic illustrations. Figure 7-2 lists notes which apply to the schematic diagrams.

7-3. Circuit board assemblies are shown in part on different pages. To find a specific instrument component, refer to the "REFERENCE DESIGNA-TIONS" box which appears on each schematic diagram.

7-4. Component procurement information and specific component descriptions are included in Section VI. Refer to page 6-1 for information on how to order parts.

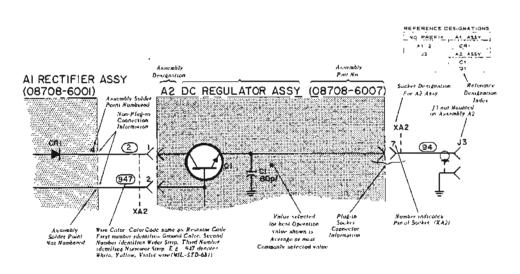
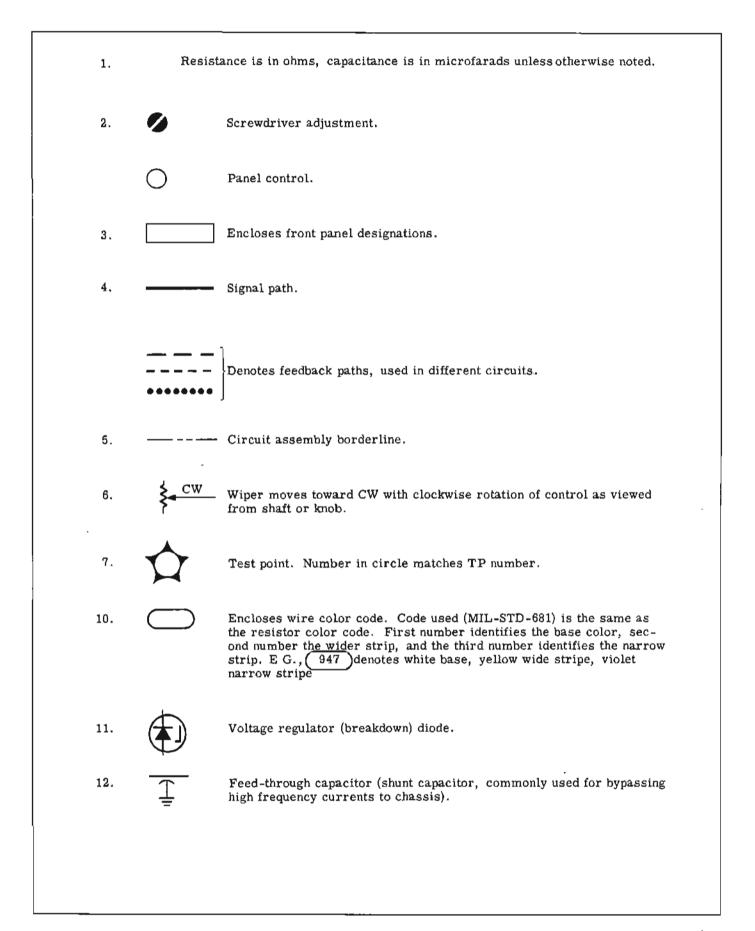


Figure 7-1. Schematic Information Illustration

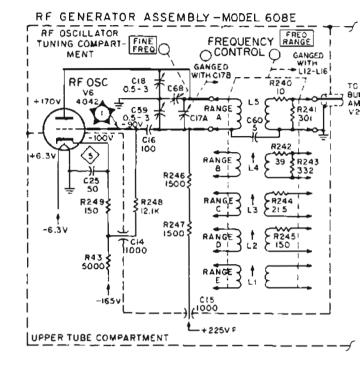




NOTES

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- INSTALLED IN MODEL 608F ONLY
- 2 VALUE SELECTED IN TEST-NOMINAL VALUE SHOWN
- 3 R214 IS 200 OHMS IN MODEL GOBE AND 392 OHMS IN MODEL 6000
- A 251 IS 301 OHMS IN MODEL GOBE AND 200 OHMS IN MODEL 6084
- 5 C25 IS NSR PART OF VE
- 6 RILA IS NSR . PART OF FLS
- (7) C37 AND C69 ARE NSR PART OF AT I
- (B) LIT IS NSR : PART OF BOLOMETER
- 9 INDICATED VOLTAGES-IMC PULSE SET RF OUTPUT MODE



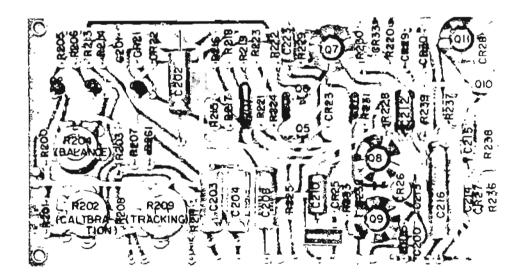
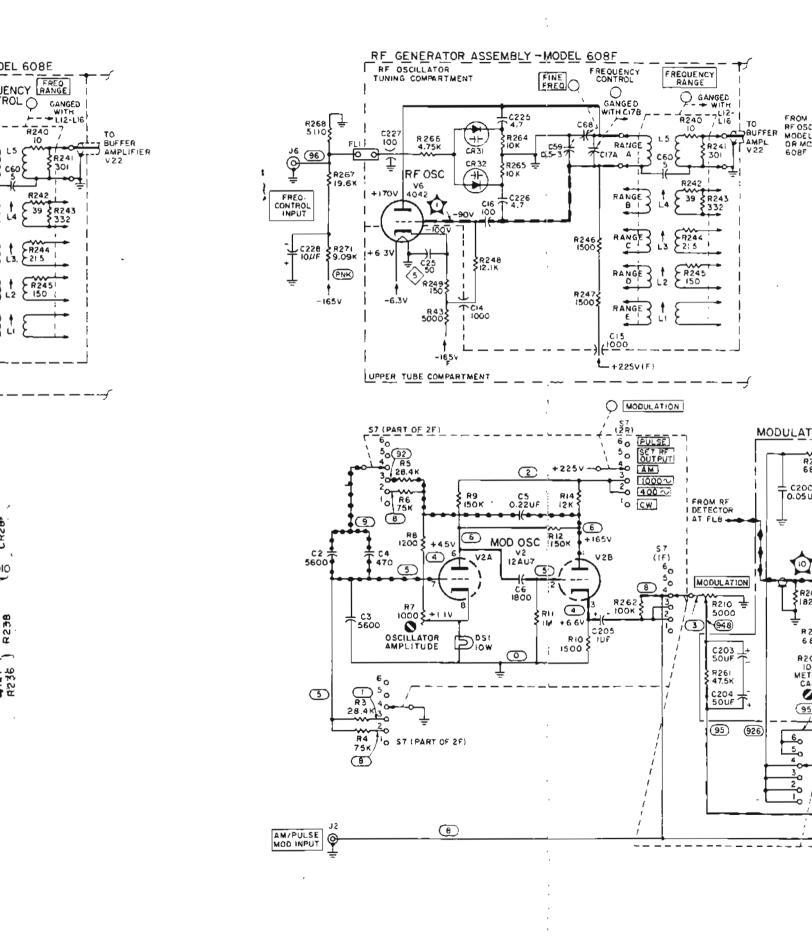
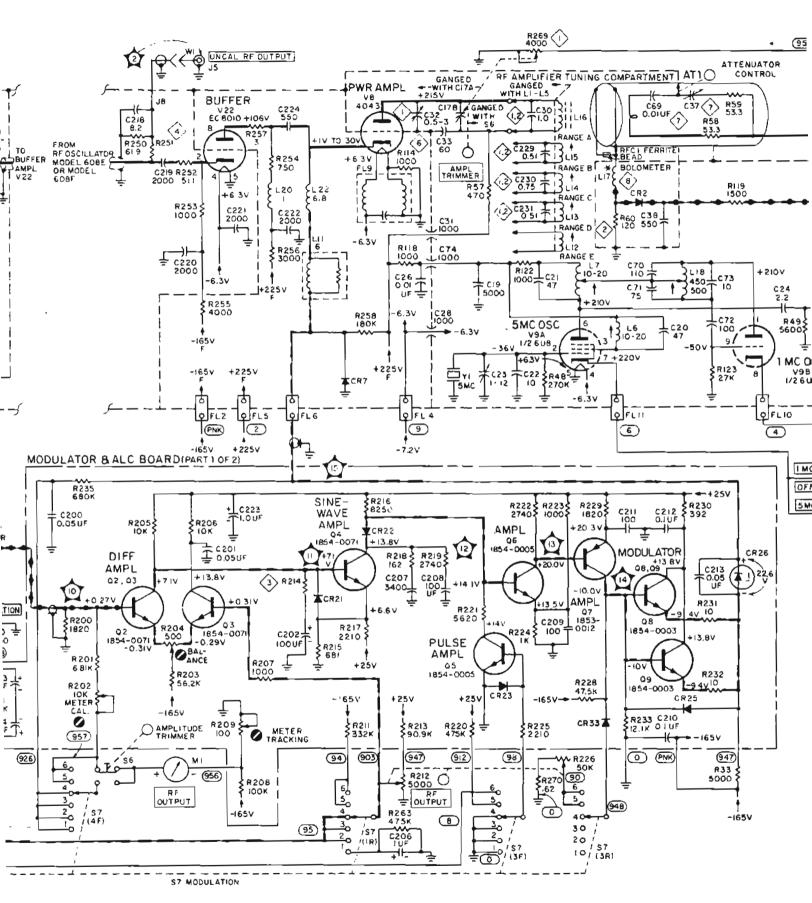


Figure 7-3. Component Identification. Modulation and ALC Board

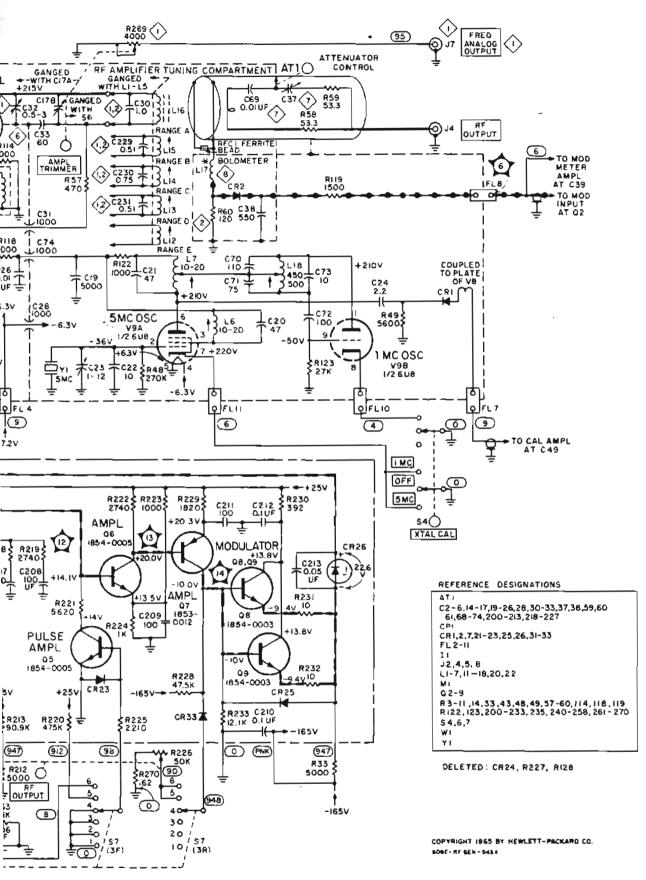


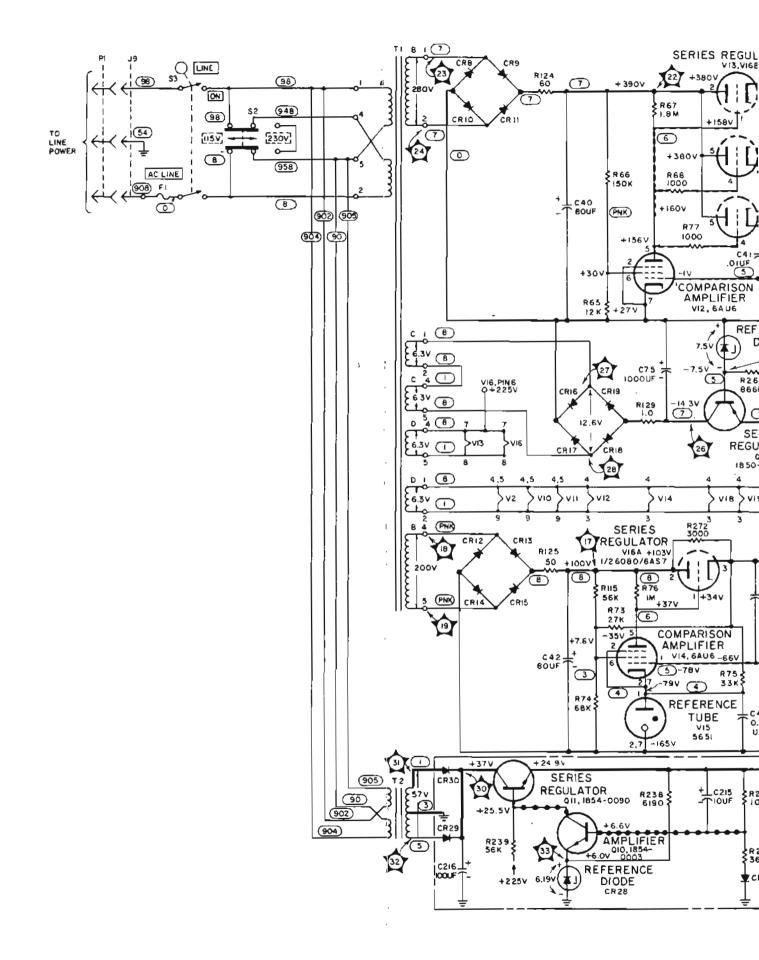
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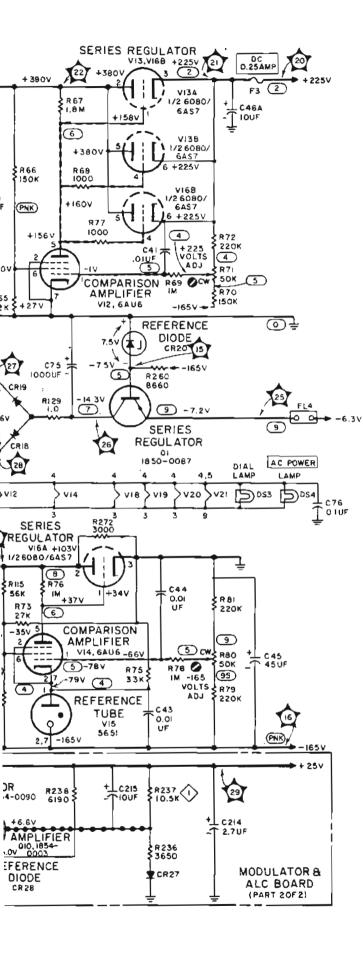
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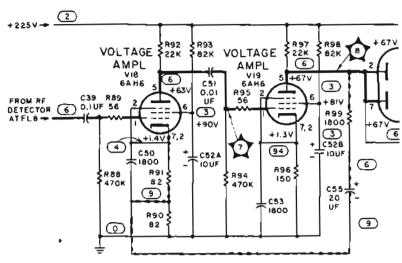
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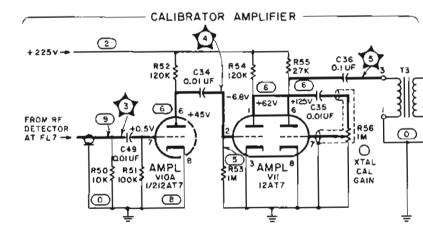






---- MODULATION METER AMPLIFIER -



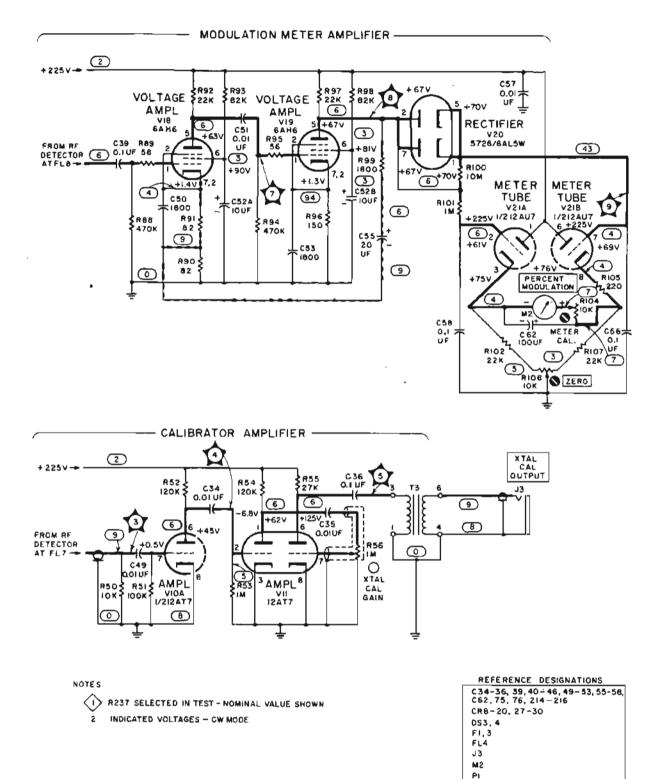


NOTES (1) R237 SELECTED IN TEST - NOMINAL VALUE SHOWN 2 INDICATED VOLTAGES - CW MODE

Figu

Section VII

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R50-56, 65-81, 88-102, 104-107, 115, R124, 125, 129, 236-239, 260, 272

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V2, 10-14, 16, 18-21

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APPENDIX I

BACKDATING

MANUAL CHANGES

Model 608E

VHF SIGNAL GENERATOR

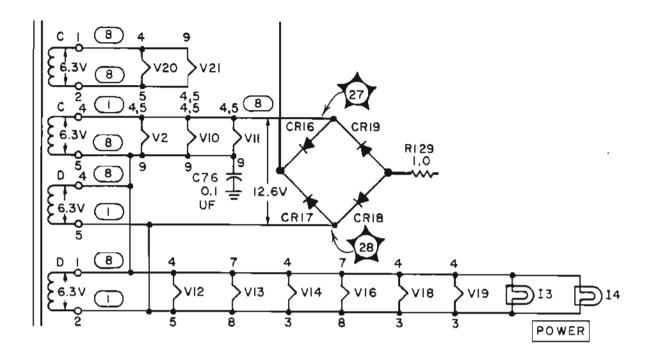
Make all backdating corrections in this manual according to the changes below.

SERIAL PREFIX OR NUMBER	MAKE MANUAL CHANGES	SERIAL PREFIX OR NUMBER	MAKE MANUAL CHANGES
539- below 00120	A through M	833-	I through M
543-	B through M	833- above 01671	J, through M
610-, 637-	C through M	833- above 01821	K, L, M
637- above 00320	D through M	833- above 02121	L,M
710-	E through M	833-above 02720	M
742-	F through M		
742- above 00970	G through M		
826-	H through M		

CHANGE A	Figure 7-5 and Parts List: Add (In series with emitter of Q1): R128, HP Stock No.0812-0033, Fixed WW 7 ohm 3% 5W. CR20, change HP Stock No. to 1902-0037.
CHANGE B	Parts List: Change Q7 to HP Stock No. 1850-0103, 2N2190 under Q7 change HP Stock No. 1205-0011 to read 1205-0012. (Change included in instruments 710-00500 and below, Recommended change for all older models). Delete under R237 descriptions: "Factory selected part - typical value given."
CHANGE C	 Figure 7-5: Delete asterisk (*) after R237 denoting factory selection; typical value given. Figure 7-5 and Parts List: Change T1 HP Stock No. from 9100-2289 to 9100-0047. Change schematic to show filament wiring of T1 as follows:

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CHANGE D	Parts List: Add the following: Right-angle connector, HP Stock No. 1250-0076.
	BNC connector J8, HP Stock No. 1250-0118. Change the HP Stock No. of cable assembly W1 from 00608-60037 to 00608-605
CHANGE E	Table 1-1: Change second sentence of External Pulse Modulation specification to read, "From 220 MHz to 480 MHz, combined rise and decay time less than 2 μ sec.
	Paragraph 5-27, step h: Change test limit for "Combined pulse rise and decay times" to read, $\dots \dots $
CHANGE F	Figure 7-5 and Parts List: Change Resistor R237 typical value from 10.5K, HP Stock No. 0698-4477 to 11K, HP Stock No. 0757-0443. Delete the asterisk indicating typical value given.
CHANGE G	Figure 7-4 and Parts List: Change MOD OSCILLATOR resistor R12 from 150K to 100K, 1W, 10%, HP Stock No. 0690-1041.
	Change schematic to show R12 connected between plate of V2A (pin 6) and cathode of V2A (pin 8).
	Change Tube V2 HP Stock No. to 1932-0043.
	Parts List: Change F1 to: 2110-0013, 3.2 amp, slow-blow, for 115V operation 2110-0005, 1.6 amp, slow-blow, for 230V operation

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CHANGE H	 Table 1-1; Change drift specification to read; "Less than 50 parts in 10⁶ per 10 minute period after one hour warmup. Less than 10 minutes to restabilize after changing frequency." Parts List:
	Change V2 from HP Part No. 1932-0029 to HP Part No. 1932-0066.
CHANGE I	Figure 7-4 and Parts List: Change Resistor R7 from 1K, HP Part No. 2100-0036 to 2K, HP Part No. 2100-0010.
	Change Resistor R8 from 1.2K, HP Part No. 0690-1221 to 680 ohms, HP Part No. 0690-6811.
CHANGE J	Figure 7-4 and Parts List: Delete Ferrite Bead RFC1, HP Part No. 9170-0029.
CHANGE K	Parts List: Change Diode CR1 from HP Part No. 1901-0518 to HP Part No. 1901-0011.
CHANGE L	Figure 7-5 and Parts List: Change Capacitor C216 from 50 WVDC, HP Part No. 0180-1819 to 100 MF 40V, HP Part No. 0180-0138.
CHANGE M	Figure 7-3: Change ALC Board pictorial to show soldered connections around the edges, instead of plug-in pins.

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APPENDIX II

BACKDATING

MANUAL CHANGES

Model 608F

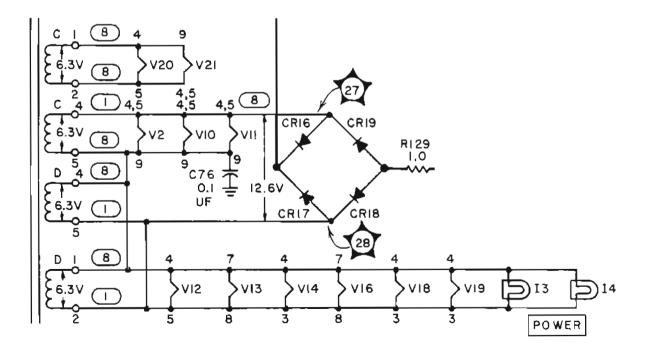
VHF SIGNAL GENERATOR

Make all backdating corrections in this manual according to the changes below.

SERIAL PREFIX OR NUMBER	MAKE MANUAL CHANGES	SERIAL PREFIX OR NUMBER	MAKE MANUAL CHANGES
530-	A through N	827-	I through N
608-	B through N	832-	J through N
610-, 637-	C through N	832- above 01101	K, through N
637- above 00200	D through N	832- above 01151	L, M,N
710-	E through N	832- above 01251	M,N
742-	F through N	832-above 01501	N
742- above 00850	G through N		
824-	H through N		

CHANGE AFigure 7-5 and Parts List:
Delete Resistor R129, 1 ohm, HP Stock No. 0813-0029.CHANGE BFigure 7-5 and Parts List:

Change T1 from HP Stock No. 9100-2289 to 9100-0047. Change schematic to show filament wiring of T1 as follows:



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CHANGE C	Parts List: Add the following: Right-angle connector, HP Stock No. 1250-0076; BNC connector J8, HP Stock No. 1250-0118. Change the HP Stock No. of cable assembly W1 from 00608-60037 to 00608-605.
CHANGE D	Parts List: Change Q7 to HP Stock No. 1850-0103, 2N2190. Under Q7 change HP Stock No. 1205-0011 to read 1205-0012. (Change included in instruments 710-00500 and below. Recommended change for all older models.)
	Delete under R237 description: "Factory selected part - typical value given."
	Figure 7-5: Delete asterisk (*) after R237 to denoting factory selection; typical value given.
	Table 1-1 (instrument serials below 710-00550): Change Auxiliary RF Output specification to read:
	"Fixed level CW signal from RF Oscillator (min amplitude 180 mv rms into 50 ohms) provided at front panel BNC female connector for use with HP 8708A Synchronizer or other external equipment (e.g. frequency counter)."
	Paragraph 5-22, step <u>b</u> (instrument serials below 710-00550): Change to read, "Power meter should indicate -1.8 dBm minimum."
CHANGE E .	Table 1-1: Change second sentence of External Pulse Modulation specification to read:
	"Above 220 MHz, combined rise and decay time less than 2 μ sec."
	Paragraph 5-27 step h: Change test limit for "Combined pulse rise and decay times" to read:
	"2 $\mu \sec \max$."
CHANGE F	Figure 7-5 and Parts List: Change Resistor R237 typical value 10.5K, HP Stock No. 0698-4477 to 11K, HP Stock No. 0757-0443. Retain the asterisk to indicate typical value given.
CHANGE G	Parts List: Change Meter, M1, HP Stock No. from 1120-1505 to 1120-0396.
CHANGE H	Figure 7-4 and Parts List: Change Resistor R12 from 150K, HP Stock No. 0690-1541 to 100K, HP Stock No. 0690-1041. Change Tube V2 HP Stock No. 1932-0043. Change schematic to show R12 connected from V2A pin 6 to V2A pin 8.
CHANGE I	Table 1-1: Change drift specification to read:
	"Less than 50 parts in 10^6 per 10 min period after one hour warmup. Less than 10 min to restabilize after changing frequency. Stability when used with 8708A Synchronizer: $5 \times 10^{-8}/\text{min}$; $2 \times 10^{-7}/10$ min; $2 \times 10^{-6}/\text{day}$; $2 \times 10^{-7}/^{\circ}C$ (0° to 55°C); $2 \times 10^{-7}/10\%$ line voltage change."

	Parts List: Change F1 to: 2100-0014, 3.2 amp, slow blow, for 115V operation. 2100-0005, 1.6 amp, slow blow, for 250V operation. Change V2 from HP Part No. 1932-0029 to HP Part No. 1932-0066.
CHANGE J	Figure 7-4 and Parts List: Change Resistor R7 from 1K, HP Stock No. 2100-0036 to 2K, HP Stock No. 2100-0010.
	Change Resistor R8 from 1.2K, HP Stock No. 0690-1221 to 680 ohms, HP Stock No. 0690-6811.
CHANGE K	Figure 7-4 and Parts List: Delete Ferrite Bead RFC1, HP Stock No. 9170-0029.
CHANGE L	Parts List: Change Diode CR1 from HP Stock No. 1901-0518 to HP Stock No. 1910-0011.
CHANGE M	Figure 7-5 and Parts List: Change Capacitor C216 from 50 WVDC, HP Stock No. 0180-1819 to 100 MF 40V, HP Stock No. 0180-0138.
CHANGE N	Figure 7-3: Change ALC Board pictorial to show soldered connections around the edges, instead of plug-in pins.

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